

# DLPC230S-Q1 Automotive DMD Controller for the DLP553x-Q1 Chipset

## 1 Features

- Qualified for automotive applications
- AEC-Q100 qualified with the following results:
  - Device temperature grade 2: –40°C to +105°C ambient operating temperature
  - Device HBM ESD classification level 2
  - Device CDM ESD classification level C4B
- **Functional Safety Quality-Managed**
  - Documentation available to aid ISO 26262 functional safety system design up to ASIL-B
- DMD display controller supporting:
  - DLP5530S-Q1 automotive interior display chipset
- Video processing
  - Scales input image to match DMD resolution
  - Bezel adjustment up ±50% vertical image position and ±10% horizontal reducing the need for mechanical alignment (HUD)
  - Support for pixel doubling or quadrupling to allow low resolution video input
  - Gamma correction
- Embedded processor with error correction (ECC)
  - On-chip diagnostic and self-test capability
  - System diagnostics including temperature monitoring, device interface monitoring, and photodiode monitoring
  - Integrated management of smooth dimming
  - Configurable GPIO
- No external RAM required, internal SRAM for image processing
- 600-MHz sub-LVDS DMD interface for low power and emission
- Spread spectrum clocking for reduced EMI
- Video input interface
  - Single OpenLDI (FPD-Link I) port up to 110 MHz
  - 24-bit RGB parallel interface up to 110 MHz
- Configurable host control interface
  - Serial peripheral interface (SPI) 10 MHz
  - I<sup>2</sup>C (400 kHz)
  - Host IRQ signal to provide real-time feedback for critical system errors
- Interface to TPS99000S-Q1 system management and illumination controller

## 2 Applications

- **Wide field of view and augmented reality head-up display (HUD)**
- Digital cluster, navigation, and infotainment windshield displays

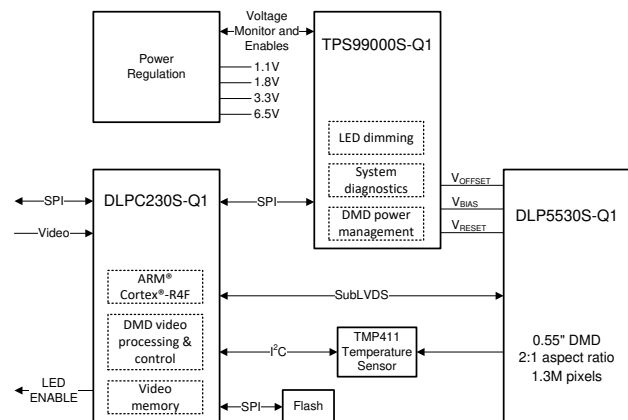
## 3 Description

The DLPC230S-Q1 DMD Display Controller for automotive applications is part of the DLP5530S-Q1 chipset which is used for interior display applications with a functional safety requirement (such as augmented reality HUDs and windshield clusters). The chipset also includes a 0.55" DMD and the TPS99000S-Q1 System Management and Illumination controller. The DLPC230S-Q1 integrates an embedded processor with error code correction (SECDED ECC), enabling host control and real-time feedback, on-chip diagnostics, and system monitoring functions. On-chip SRAM is included to remove the need for external DRAM. Combined with the TPS99000S-Q1, the DLPC230S-Q1 supports high dynamic range dimming of over 5000:1 for HUD applications. Sub-LVDS 600-MHz DMD interface allows high DMD refresh rates to generate seamless and brilliant digital images, while simultaneously reducing radiated emissions.

### Device Information

PART NUMBER	PACKAGE <sup>(1)</sup>	BODY SIZE (NOM)
DLPC230S-Q1	BGA (324)	23.00 mm × 23.00 mm

- (1) For all available packages, see the orderable addendum at the end of the data sheet.



**DLP553x-Q1 TI DLP® Chipset System Block Diagram**



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## 4 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

Changes from Revision * (August 2020) to Revision A (October 2020)	Page
• Changed device status from <i>Advance Information</i> to <i>Production Data</i> .....	1

## 5 Pin Configuration and Functions

	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	17	18	19	20	21	22
A	GND18A_LVDS	DMD_HS1_WDATA0_N	DMD_HS1_WDATA1_N	DMD_HS1_WDATA2_N	DMD_HS1_WDATA3_N	DMD_HS1_CLK_N	DMD_HS1_WDATA4_N	DMD_HS1_WDATA5_N	DMD_HS1_WDATA6_N	DMD_HS1_WDATA7_N	DMD_LSO_WDATA_N	DMD_LSO_CLK_N	DMD_HSO_WDATA7_N	DMD_HSO_WDATA6_N	DMD_HSO_WDATA5_N	DMD_HSO_WDATA4_N	DMD_HSO_WDATA3_N	DMD_HSO_WDATA2_N	DMD_HSO_WDATA1_N	DMD_HSO_WDATA0_N	DMD_HSO_WDATA0_N	GND18A_LVDS
B	VCC18A_LVDS	DMD_HS1_WDATA0_P	DMD_HS1_WDATA1_P	DMD_HS1_WDATA2_P	DMD_HS1_WDATA3_P	DMD_HS1_CLK_P	DMD_HS1_WDATA4_P	DMD_HS1_WDATA5_P	DMD_HS1_WDATA6_P	DMD_HS1_WDATA7_P	DMD_LSO_WDATA_P	DMD_LSO_CLK_P	DMD_HSO_WDATA7_P	DMD_HSO_WDATA6_P	DMD_HSO_WDATA5_P	DMD_HSO_WDATA4_P	DMD_HSO_WDATA3_P	DMD_HSO_WDATA2_P	DMD_HSO_WDATA1_P	DMD_HSO_WDATA0_P	DMD_HSO_WDATA0_P	VCC18A_LVDS
C	VCC18A_LVDS	GND18A_LVDS	GND18A_LVDS	GND18A_LVDS	GND18A_LVDS	GND18A_LVDS	GND18A_LVDS	DMD_LS1_RDATA	DMD_LS0_RDATA	DMD_LS0_WDATA	DMD_LS0_CLK	GND11AD_PLD	GND11AD_PLM	GNDIOLA_COSC	VCC310_OSC	GND18A_LVDS	GND18A_LVDS	GND18A_LVDS	GND18A_LVDS	GND18A_LVDS	GND18A_LVDS	VCC18A_LVDS
D	PMIC_SPI_DOUT	VCC18A_LVDS	VCC18A_LVDS	VCC18A_LVDS	VCC18A_LVDS	RPI_1	VCC18A_LVDS	GND18A_LVDS	RPI_LS	VCC18IO	DMD_DE_N_ARSTZ	VCC11AD_PLD	VCC11AD_PLM	PLL_REFCLK_O	PLL_REFCLK_I	OSC_BYPASS	RPI_0	VCC18A_LVDS	VCC18A_LVDS	VCC18A_LVDS	VCC18A_LVDS	GPIO_31
E	PMIC_SPI_CLK	PMIC_SPI_CSZ_0	PMIC_PA_RKZ	VCC11A_DDI_1														VCC11A_DDI_0	VCC18A_LVDS	GPIO_30	GPIO_29	
F	PMIC_SPI_DIN	PMIC_LED_SEL_0	RESETZ	VCC11A_DDI_1														VCC11A_DDI_0	GPIO_28	GPIO_27	GPIO_26	
G	PMIC_LED_SEL_1	PMIC_LED_SEL_2	PMIC_INT_Z	VCC														VCC310_2	JTAGTDO_3	JTAGTMS_1	JTAGTCK	
H	PMIC_LED_SEL_3	PMIC_AD_3_CLK	HWTEST_EN	VCC310_MVGP														VCC	JTAGTDO_2	VSYN	HSYN	
J	PMIC_AD_3_MOSI	PMIC_AD_3_MISO	GPIO_00	VCC310				GND	GND	VCC	GND	GND	GND					VCC	JTAGTDO_1	PDATA_2_3	PDATA_2_2	
K	GPIO_01	GPIO_02	GPIO_03	VCC310				GND	GND	GND	GND	GND	GND					VCC310_1_NTF	JTAGTDI	PDATA_2_1	PDATA_2_0	
L	GPIO_04	GPIO_05	GPIO_06	VCC				GND	GND	GND	GND	GND	GND					VCC310_1_NTF	JTAGTRST_Z	PDATA_1_9	PDATA_1_8	
M	GPIO_07	GPIO_08	GPIO_09	VCC310				GND	GND	GND	GND	GND	GND					VCC310_1_NTF	JTAGTMS_3	PDATA_1_7	PDATA_1_6	
N	GPIO_10	GPIO_11	GPIO_12	VCC310				GND	GND	GND	GND	GND	GND					VCC	JTAGTMS_2	PDATA_1_4	PDATA_1_5	
P	GPIO_13	GPIO_14	GPIO_15	VCC310				GND	GND	GND	GND	GND	GND					VCC	HOST_IIC_SDA	DATEN	PDATA_1_3	
R	GPIO_16	GPIO_17	GPIO_18	HOST_IF_SEL														VCC310_1_NTF	HOST_IIC_SCL	PDATA_1_1	PCLK	
T	GPIO_19	GPIO_20	GPIO_21	VCC														VCC310_1_NTF	HOST_IRQ	PDATA_9	PDATA_1_2	
U	GPIO_22	GPIO_23	GPIO_24	VCC														VCC	HOST_SPI_DOUT	PDATA_7	PDATA_1_0	
V	HOST_SPI_MODE	FLSH_SPI_DIO_0	GPIO_25	VCC310_FLSH														VCC	HOST_SPI_DIN	PDATA_5	PDATA_8	
W	FLSH_SPI_CLK	FLSH_SPI_DIO_1	FLSH_SPI_DIO_3	VCC310	VCC310	VCC	EFUSE_VDDQ	VCC	VCC33A_LVDS	VCC	VCC11A_LVDS	VCC11A_LVDS	VCC33A_LVDS	GND33A_LVDS	VCC33A_LVDS	VCC	VCC11A_LVDS	VCC11A_LVDS	VCC33A_LVDS	HOST_SPI_CSZ	PDATA_3	PDATA_6
Y	FLSH_SPI_CSZ	FLSH_SPI_DIO_2	GND	TSTPT_0	TSTPT_2	TSTPT_4	TSTPT_6	EFUSE_PROB_R33	VCC33A_LVDS	GND11A_LVDS	GND11A_LVDS	GND11A_LVDS	VCC33A_LVDS	GND33A_LVDS	VCC33A_LVDS	GND11A_LVDS	GND11A_LVDS	GND11A_LVDS	VCC33A_LVDS	HOST_SPI_CLK	PDATA_2	PDATA_4
AA	GND	GND	RTTPUB_ENZ	TSTPT_1	TSTPT_3	TSTPT_5	TSTPT_7	GND33A_LVDS	L1_DATA0_N	L1_DATA1_N	L1_CLK_N	L1_DATA2_N	L1_DATA3_N	GND33A_LVDS	L2_DATA0_N	L2_DATA1_N	L2_CLK_N	L2_DATA2_N	L2_DATA3_N	GND33A_LVDS	PDATA_0	PDATA_1
AB	GND	GND	CRCZ_CH_KSM_SEL	MSTR_SCL	MSTR_SDA	ETM_TRA_CECLK	ETM_TRA_CECLK	GND33A_LVDS	L1_DATA0_P	L1_DATA1_P	L1_CLK_P	L1_DATA2_P	L1_DATA3_P	GND33A_LVDS	L2_DATA0_P	L2_DATA1_P	L2_CLK_P	L2_DATA2_P	L2_DATA3_P	GND33A_LVDS	GND33A_LVDS	GND

Note that there is one VCC power ball located in the thermal ball array.

Figure 5-1. ZDQ Package 324-Pin BGA Top View

## Pin Functions – Board Level Test, Debug, and Initialization

NAME	PIN		I/O <sup>(1)</sup>	DESCRIPTION
	NUMBER	NUMBER		
RESETZ	F3		I <sub>7</sub>	Active low power-on reset for the DLPC230S-Q1. A low-to-high transition starts self-configuration and initialization of the ASIC. (‘0’ = Reset, ‘1’ = Normal Operation) All ASIC power and input clocks must be stable before this reset is de-asserted high. The signals listed below should be forced low by external pull-down, and will then be driven low as the power supplies stabilize with RESETZ asserted. <i>PMIC_LEDSEL_0, PMIC_LEDSEL_1, PMIC_LEDSEL_2, PMIC_LEDSEL_3, DMD_DEN_ARSTZ, PMIC_AD3_CLK, and PMIC_AD3_MOSI</i> All other bi-directional and output signals will be tri-stated while reset is asserted. External pull-ups or pull-downs must be added where necessary to protect external devices that would typically be driven by the ASIC to prevent device malfunction. This pin includes hysteresis. Specific timing requirements for this signal are shown in <a href="#">Section 6.12</a> .
PMIC_PARKZ	E3		I <sub>7</sub>	DMD Park Control (‘0’ = Park, ‘1’ = Un-Park) The TI TPS99000S-Q1 device is used to control this signal. As part of this function, it monitors power to the DLPC230S-Q1 watching for an imminent power loss condition, upon which it will drive the PMIC_PARKZ signal accordingly. The specific timing requirements for this signal are shown in <a href="#">Section 6.12</a> .

PIN		I/O <sup>(1)</sup>	DESCRIPTION
NAME	NUMBER		
HOST_IF_SEL	R4	B <sub>13,14</sub>	<p>Selects which input interface port will be used for Host Command and Control. The port that is not selected as the Host Command and Control port will be available as a Diagnostic Processor monitoring port. (‘0’ = Host SPI, ‘1’ = Host I<sup>2</sup>C)</p> <p>This pin includes a weak internal pull-down. If a pull-up is used to obtain a ‘1’ value, the pull-up value must be ≤ 8 kΩ.</p> <p>Tri-stated while RESETZ is asserted low, and is sampled as a host directive approximately 1.5 μs after RESETZ is de-asserted. It may be driven as an output for TI debug use after sampling.</p>
HOST_SPI_MODE	V1	B <sub>13,14</sub>	<p>Selects the SPI mode (clock phase and polarity) that will be used with the HOST SPI interface. This value is applicable regardless of whether the Host SPI interface is used for Host Command and Control, or for the Diagnostic Processor monitoring port. (‘0’ = SPI Mode 0 or 3, ‘1’ = SPI Mode 1 or 2)</p> <p>This pin includes a weak internal pull-down. If a pull-up is used to obtain a ‘1’ value, the pull-up value must be ≤ 8 kΩ.</p> <p>Tri-stated while RESETZ is asserted low, and is sampled as a host directive approximately 1.5 μs after RESETZ is de-asserted. It may be driven as an output for TI debug use after sampling.</p>
RTPUB_ENZ	AA3	B <sub>13,14</sub>	TI internal use. Must be left unconnected. Includes a weak pull-down.
CRCZ_CHKSUM_SEL	AB3	B <sub>13,14</sub>	<p>Selects whether the Host will use 8-bit CRC or Checksum on the Host Command and Control interface. This value is only applicable for the Host Command and Control interface. The value for the Diagnostic Processor monitoring port will be specified in Flash. (‘0’ = 8-bit CRC, ‘1’ = 8-bit Checksum)</p> <p>This pin includes a weak internal pull-down. If a pull-up is used to obtain a ‘1’ value, the pull-up value must be ≤ 8 kΩ.</p> <p>Tri-stated while RESETZ is asserted low, and is sampled as a host directive approximately 1.5 μs after RESETZ is de-asserted. It may be driven as an output for TI debug use after sampling.</p>
ETM_TRACECLK	AB6	O <sub>13</sub>	TI internal use. Must be left unconnected. (Clock for Trace Debug)
ETM_TRACECTL	AB7	O <sub>13</sub>	TI internal use. Must be left unconnected. (Control for Trace Debug)
TSTPT_0	Y4	B <sub>13,14</sub>	<p>Test pin 0 / STAY-IN-BOOT: Selects whether the system should stay in the Boot Application, or proceed with the normal load of the Main Application. (‘0’ = Load Main Application, ‘1’ = Stay in Boot Application)</p> <p>This pin includes a weak internal pull-down. If a pull-up is being used to obtain a ‘1’ value, the pull-up value must be ≤ 8 kΩ.</p> <p>Tri-stated while RESETZ is asserted low, and is sampled as a host directive approximately 1.5 μs after RESETZ is de-asserted. It may be driven as an output for debug use after sampling as described in <a href="#">Section 8.3.11</a>.</p>
TSTPT_1	AA4	B <sub>13,14</sub>	<p>Test pin 1 : This pin must be externally pulled down, left open or unconnected. Includes a weak pull-down. It may be driven as an output for debug use as described in <a href="#">Section 8.3.11</a>.</p>
TSTPT_2	Y5	B <sub>13,14</sub>	<p>Test pin 2 : This pin must be externally pulled down, left open or unconnected. Includes a weak pull-down. It may be driven as an output for debug use as described in <a href="#">Section 8.3.11</a>.</p>
TSTPT_3	AA5	B <sub>13,14</sub>	<p>Test pin 3 : This pin must be externally pulled down, left open or unconnected. Includes a weak pull-down. It may be driven as an output for debug use as described in <a href="#">Section 8.3.11</a>.</p>
TSTPT_4	Y6	B <sub>13,14</sub>	<p>Test pin 4: This pin must be externally pulled down, left open or unconnected. Includes a weak pull-down. It may be driven as an output for debug use as described in <a href="#">Section 8.3.11</a>.</p>

PIN		I/O <sup>(1)</sup>	DESCRIPTION
NAME	NUMBER		
TSTPT_5	AA6	B <sub>13,14</sub>	<p>Test pin 5 / Spread Spectrum Disable: Selects whether spread spectrum flash settings are used or whether spread spectrum clocking will be disabled. (‘0’ = Spread Spectrum Disabled, ‘1’ = Use flash Spread Spectrum settings) This pin includes a weak internal pull-down. If a pull-up is being used to obtain a ‘1’ value, the pull-up value must be ≤ 8 kΩ. This signal is tri-stated while RESETZ is asserted low, and is sampled as a host directive approximately 1.5 μs after RESETZ is de-asserted. It may be driven as an output for debug use after sampling as described in <a href="#">Section 8.3.11</a>.</p>
TSTPT_6	Y7	B <sub>13,14</sub>	<p>Test pin 6: An external pull-up resistor must be used (≤ 8 kΩ since pin includes a weak pull-down). This signal is tri-stated while RESETZ is asserted low, and is sampled as a host directive approximately 1.5 μs after RESETZ is de-asserted. It may be driven as an output for debug use after sampling as described in <a href="#">Section 8.3.11</a>.</p>
TSTPT_7	AA7	B <sub>13,14</sub>	<p>Test pin 7: This pin must be externally pulled down, left open or unconnected. Includes a weak pull-down. It may be driven as an output for debug use as described in <a href="#">Section 8.3.11</a>.</p>
HWTEST_EN	H3	I <sub>14</sub>	<p>Manufacturing test enable signal. This signal must be connected directly to ground on the PCB. Includes weak internal pull-down and hysteresis.</p>
JTAGTCK	G22	I <sub>11</sub>	<p>JTAG Serial Data Clock Includes a weak internal pull-up.</p>
JTAGTMS1	G21	I <sub>11</sub>	<p>JTAG Test Mode Select Includes weak internal pull-up.</p>
JTAGTRSTZ	L20	I <sub>11</sub>	<p>JTAG Reset Includes a weak internal pull-up and Hysteresis. For normal operation, this pin must be pulled to ground through an external 8 kΩ or less resistor. <b>Failure to pull this pin low during normal operation will cause start-up and initialization problems.</b> For JTAG Boundary Scan, this pin must be pulled-up or left disconnected.</p>
JTAGTDI	K20	I <sub>11</sub>	<p>JTAG Serial Data In Includes a weak internal pull-up.</p>
JTAGTDO1	J20	B <sub>10,11</sub>	<p>JTAG Serial Data Out Includes weak internal pull-up.</p>
JTAGTDO2	H20	B <sub>10,11</sub>	<p>This pin must be left open or unconnected. Includes a weak internal pull-up.</p>
JTAGTDO3	G20	B <sub>10,11</sub>	<p>This pin must be left open or unconnected. Includes a weak internal pull-up.</p>
JTAGTMS2	N20	I <sub>11</sub>	<p>This pin must be left open or unconnected. Includes a weak internal pull-up. See <a href="#">Section 8.3.11</a> for important debug access considerations.</p>
JTAGTMS3	M20	I <sub>11</sub>	<p>This pin must be left open or unconnected. Includes a weak internal pull-up. See <a href="#">Section 8.3.11</a> for important debug access considerations.</p>

(1) See [Table 5-1](#) for more information on I/O definitions.

## Pin Functions – Parallel Port Input Data and Control

PIN <sup>(1)</sup>		I/O <sup>(2)</sup>	DESCRIPTION PARALLEL RGB MODE
NAME	NUMBER		
PCLK	R22	I <sub>11</sub>	Pixel clock
VSYNC	H21	I <sub>11</sub>	Vsync <sup>(3)</sup>
HSYNC	H22	I <sub>11</sub>	Hsync <sup>(3)</sup>
DATEN	P21	I <sub>11</sub>	Data Valid
			<b>(TYPICAL RGB 888)</b>
PDATA_0	AA21	I <sub>11</sub>	Blue (bit weight 1)
PDATA_1	AA22		Blue (bit weight 2)
PDATA_2	Y21		Blue (bit weight 4)
PDATA_3	W21		Blue (bit weight 8)
PDATA_4	Y22		Blue (bit weight 16)
PDATA_5	V21		Blue (bit weight 32)
PDATA_6	W22		Blue (bit weight 64)
PDATA_7	U21		Blue (bit weight 128)
			<b>(TYPICAL RGB 888)</b>
PDATA_8	V22	I <sub>11</sub>	Green (bit weight 1)
PDATA_9	T21		Green (bit weight 2)
PDATA_10	U22		Green (bit weight 4)
PDATA_11	R21		Green (bit weight 8)
PDATA_12	T22		Green (bit weight 16)
PDATA_13	P22		Green (bit weight 32)
PDATA_14	N21		Green (bit weight 64)
PDATA_15	N22		Green (bit weight 128)
			<b>(TYPICAL RGB 888)</b>
PDATA_16	M22	I <sub>11</sub>	Red (bit weight 1)
PDATA_17	M21		Red (bit weight 2)
PDATA_18	L22		Red (bit weight 4)
PDATA_19	L21		Red (bit weight 8)
PDATA_20	K22		Red (bit weight 16)
PDATA_21	K21		Red (bit weight 32)
PDATA_22	J22		Red (bit weight 64)
PDATA_23	J21		Red (bit weight 128)

- (1) Unused inputs should be grounded or pulled down to ground through an external resistor ( $\leq 10\text{ k}\Omega$ ).
- (2) See [Table 5-1](#) for more information on I/O definitions.
- (3) VSYNC and HSYNC polarity are software programmable.

## Pin Functions – OpenLDI Ports Input Data and Control

PIN <sup>(1)</sup> <sup>(2)</sup>		I/O <sup>(3)</sup>	DESCRIPTION
NAME	NUMBER		
L1_CLK_P L1_CLK_N	AB11 AA11	I <sub>18</sub>	OpenLDI (FPD Link I) Port 1 Clock Lane
L1_DATA0_P L1_DATA0_N L1_DATA1_P L1_DATA1_N L1_DATA2_P L1_DATA2_N L1_DATA3_P L1_DATA3_N	AB9 AA9 AB10 AA10 AB12 AA12 AB13 AA13	I <sub>18</sub>	OpenLDI (FPD Link I) Port 1 Data Lanes: Intra-port data lane swapping can be done on a product configuration basis to support board considerations.
L2_CLK_P L2_CLK_N	AB17 AA17	I <sub>18</sub>	OpenLDI (FPD Link I) Port 2 Clock Lane
L2_DATA0_P L2_DATA0_N L2_DATA1_P L2_DATA1_N L2_DATA2_P L2_DATA2_N L2_DATA3_P L2_DATA3_N	AB15 AA15 AB16 AA16 AB18 AA18 AB19 AA19	I <sub>18</sub>	OpenLDI (FPD Link I) Port 2 Data Lanes: Intra-port data lane swapping can be done on a product configuration basis to support board considerations.

- (1) The system only supports the operational use of one port. As two ports are available, the host can select which port they wish to be active (to optimize board routing as an example).
- (2) The inputs for any un-used port(s) should be left unconnected, and will be powered down by the system.
- (3) See [Table 5-1](#) for more information on I/O definitions.

## Pin Functions – DMD Reset and Bias Control Interfaces

PIN <sup>(1) (2)</sup>		I/O <sup>(3)</sup>	DESCRIPTION
NAME	NUMBER		
DMD_DEN_ARSTZ	D11	O <sub>1</sub>	DMD driver enable signal ('1' = Enabled, '0' = Reset) This signal will be driven low after the DMD is parked and before power is removed from the DMD. If the 1.8-V power to the DLPC230S-Q1 is independent of the 1.8-V power to the DMD, then an external pull-down resistor ( $\leq 2.2 \text{ k}\Omega$ ) must be used to hold the signal low in the event DLPC230S-Q1 power is inactive while DMD power is applied.
DMD_LS0_CLK	C11	O <sub>2</sub>	TI internal use. Must be left unconnected.
DMD_LS0_WDATA	C10	O <sub>2</sub>	TI internal use. Must be left unconnected.
DMD_LS0_RDATA	C9	I <sub>3</sub>	DMD, low-speed single-ended serial read data
DMD_LS1_RDATA	C8	I <sub>3</sub>	DMD, low-speed single-ended serial read data (Training data response for second port of DMD)
DMD_LS0_CLK_P DMD_LS0_CLK_N	B12 A12	O <sub>4</sub>	DMD low-speed differential interface clock
DMD_LS0_WDATA_P DMD_LS0_WDATA_N	B11 A11	O <sub>4</sub>	DMD low-speed differential interface write data

- (1) The low-speed write control interface to the DMD is differential.
- (2) All control interface reads will make use of the single-ended low-speed signals. The read data will be clocked by the write clock.
- (3) See [Table 5-1](#) for more information on I/O definitions.

## Pin Functions – DMD Sub-LVDS Interfaces

PIN		I/O <sup>(1)</sup>	DESCRIPTION
NAME	NUMBER		
DMD_HS0_CLK_P DMD_HS0_CLK_N	B17 A17	O <sub>4</sub>	DMD high-speed interface, Port 0 Clock Lane.
DMD_HS0_WDATA0_P DMD_HS0_WDATA0_N DMD_HS0_WDATA1_P DMD_HS0_WDATA1_N DMD_HS0_WDATA2_P DMD_HS0_WDATA2_N DMD_HS0_WDATA3_P DMD_HS0_WDATA3_N DMD_HS0_WDATA4_P DMD_HS0_WDATA4_N DMD_HS0_WDATA5_P DMD_HS0_WDATA5_N DMD_HS0_WDATA6_P DMD_HS0_WDATA6_N DMD_HS0_WDATA7_P DMD_HS0_WDATA7_N	B21 A21 B20 A20 B19 A19 B18 A18 B16 A16 B15 A15 B14 A14 B13 A13	O <sub>4</sub>	DMD high-speed interface, Port 0 Data Lanes: The true numbering and application of the DMD_HS_DATA pins are software configuration dependent as discussed in <a href="#">Section 8.3.3</a> .
DMD_HS1_CLK_P DMD_HS1_CLK_N	B6 A6	O <sub>4</sub>	DMD high-speed interface, Port 1 Clock Lane.



PIN		I/O <sup>(1)</sup>	DESCRIPTION
NAME	NUMBER		
DMD_HS1_WDATA0_P	B2	O <sub>4</sub>	DMD high-speed interface, Port 1 Data Lanes: The true numbering and application of the DMD_HS_DATA pins are software configuration dependent as discussed in <a href="#">Section 8.3.3</a> .
DMD_HS1_WDATA0_N	A2		
DMD_HS1_WDATA1_P	B3		
DMD_HS1_WDATA1_N	A3		
DMD_HS1_WDATA2_P	B4		
DMD_HS1_WDATA2_N	A4		
DMD_HS1_WDATA3_P	B5		
DMD_HS1_WDATA3_N	A5		
DMD_HS1_WDATA4_P	B7		
DMD_HS1_WDATA4_N	A7		
DMD_HS1_WDATA5_P	B8		
DMD_HS1_WDATA5_N	A8		
DMD_HS1_WDATA6_P	B9		
DMD_HS1_WDATA6_N	A9		
DMD_HS1_WDATA7_P	B10		
DMD_HS1_WDATA7_N	A10		

(1) See [Table 5-1](#) for more information on I/O definitions.

## Pin Functions – Peripheral Interfaces

PIN		I/O <sup>(1)</sup>	DESCRIPTION
NAME	NUMBER		
HOST_IRQ <sup>(2)</sup>	T20	O <sub>10</sub>	Host interrupt (output active HIGH) This signal is used to indicate that the DLPC230S-Q1 has detected a serious error for which the ASIC has initiated an Emergency Shutdown. This is discussed further in <a href="#">Section 7.1</a> . The DLPC230S-Q1 tri-states this output during reset. An external pull-down ( $\leq 10\text{ k}\Omega$ ) is required to drive this signal to its inactive state.
HOST_IIC_SCL	R20	B <sub>12</sub>	I <sup>2</sup> C Port (Slave), Host Command and Control to ASIC, SCL (bidirectional, open-drain): An external pull-up is required.
HOST_IIC_SDA	P20	B <sub>12</sub>	I <sup>2</sup> C Port (Slave), Host Command and Control to ASIC, SDA. (bidirectional, open-drain): An external pull-up is required.
HOST_SPI_CLK	Y20	I <sub>11</sub>	SPI Port (Slave), Host Command and Control to ASIC, clock
HOST_SPI_CSZ	W20	I <sub>11</sub>	SPI Port (Slave), Host Command and Control to ASIC, chip select (active low input) An external pull-up resistor ( $\leq 2.2\text{ k}\Omega$ ) is required to avoid a floating chip select input to the ASIC
HOST_SPI_DIN	V20	I <sub>11</sub>	SPI Port (Slave), Host Command and Control to ASIC, receive data in
HOST_SPI_DOUT	U20	O <sub>10</sub>	SPI Port (Slave), Host Command and Control to ASIC, transmit data out
FLSH_SPI_CSZ	Y1	O <sub>8</sub>	SPI Port (Master), Control Interface to Flash device, chip select (active low output) An external pullup resistor ( $\leq 10\text{ k}\Omega$ ) is required to avoid a floating chip select input to the Flash
FLSH_SPI_CLK	W1	O <sub>8</sub>	SPI Port (Master), Control Interface to Flash device, clock
FLSH_SPI_DIO_0	V2	B <sub>8,9</sub>	SPI Port (Master), Control Interface to Flash device, transmit and receive data An external pullup resistor ( $\leq 10\text{ k}\Omega$ ) is required
FLSH_SPI_DIO_1	W2	B <sub>8,9</sub>	SPI Port (Master), Control Interface to Flash device, transmit and receive data An external pullup resistor ( $\leq 10\text{ k}\Omega$ ) is required
FLSH_SPI_DIO_2	Y2	B <sub>8,9</sub>	SPI Port (Master), Control Interface to Flash device, transmit and receive data An external pullup resistor ( $\leq 3.3\text{ k}\Omega$ ) is required
FLSH_SPI_DIO_3	W3	B <sub>8,9</sub>	SPI Port (Master), Control Interface to Flash device, transmit and receive data An external pullup resistor ( $\leq 3.3\text{ k}\Omega$ ) is required
PMIC_INTZ <sup>(2)</sup>	G3	I <sub>7</sub>	TPS99000S-Q1 interrupt (input with hysteresis) The ASIC provides a weak internal pull-up
PMIC_SPI_CLK	E1	O <sub>6</sub>	SPI Port (Master), General Control Interface to TPS99000S-Q1, clock
PMIC_SPI_CSZ0	E2	O <sub>6</sub>	SPI Port (Master), General Control Interface to TPS99000S-Q1, chip select 0 (active low output) An external pullup resistor ( $\leq 10\text{ k}\Omega$ ) must be used to avoid floating chip select inputs to the external SPI device during ASIC reset assertion.
PMIC_SPI_DIN	F1	I <sub>7</sub>	SPI Port (Master), General Control Interface to TPS99000S-Q1, receive data in
PMIC_SPI_DOUT	D1	O <sub>6</sub>	SPI Port (Master), General Control Interface to TPS99000S-Q1, transmit data out
PMIC_AD3_CLK	H2	O <sub>20</sub>	Sequencer Clock / TPS99000S-Q1 primary system clock An external pull-down resistor ( $\leq 10\text{ k}\Omega$ ) must be used to avoid uncontrolled behavior during ASIC reset assertion.
PMIC_AD3_MISO	J2	I <sub>14</sub>	Measurement control interface to TPS99000S-Q1, receive data in
PMIC_AD3_MOSI	J1	O <sub>20</sub>	Measurement control interface to TPS99000S-Q1, transmit data out An external pull-down resistor ( $\leq 10\text{ k}\Omega$ ) must be used to avoid uncontrolled behavior during ASIC reset assertion.
PMIC_LEDSEL_0	F2	O <sub>6</sub>	LED Control Interface to TPS99000S-Q1 An external pull-down resistor ( $\leq 10\text{ k}\Omega$ ) must be used to avoid uncontrolled illumination during ASIC reset assertion.
PMIC_LEDSEL_1	G1	O <sub>6</sub>	LED Control Interface to TPS99000S-Q1 An external pull-down resistor ( $\leq 10\text{ k}\Omega$ ) must be used to avoid uncontrolled illumination during ASIC reset assertion.
PMIC_LEDSEL_2	G2	O <sub>6</sub>	LED Control Interface to TPS99000S-Q1 An external pull-down resistor ( $\leq 10\text{ k}\Omega$ ) must be used to avoid uncontrolled illumination during ASIC reset assertion.

PIN		I/O <sup>(1)</sup>	DESCRIPTION
NAME	NUMBER		
PMIC_LEDSEL_3	H1	O <sub>6</sub>	LED Control Interface to TPS9900S-Q1 An external pull-down resistor ( $\leq 10\text{ k}\Omega$ ) must be used to avoid uncontrolled illumination during ASIC reset assertion.
MSTR_SDA	AB5	B <sub>15</sub>	I <sup>2</sup> C Port (Master), SDA. (bidirectional, open-drain) An external pull-up is required. Typical use of the Master I <sup>2</sup> C port is communication with temperature sensing devices and an optional EEPROM. The Master I <sup>2</sup> C I/Os are powered by VCC3IO (3.3 V only).
MSTR_SCL	AB4	B <sub>15</sub>	I <sup>2</sup> C Port (Master), SCL. (bidirectional, open-drain) An external pull-up is required. Typical use of the Master I <sup>2</sup> C port is communication with temperature sensing devices and an optional EEPROM. The Master I <sup>2</sup> C I/Os are powered by VCC3IO (3.3 V only).

- (1) See [Table 5-1](#) for more information on I/O definitions.  
 (2) For more information about usage, see [Section 7.1](#).

## Pin Functions – GPIO Peripheral Interface

PIN <sup>(1) (3)</sup>		I/O <sup>(2)</sup>	DESCRIPTION
NAME	NUMBER		
GPIO_31	D22	B <sub>20,14</sub>	General purpose I/O 31
GPIO_30	E21	B <sub>20,14</sub>	General purpose I/O 30
GPIO_29	E22	B <sub>20,14</sub>	General purpose I/O 29
GPIO_28	F20	B <sub>20,14</sub>	General purpose I/O 28
GPIO_27	F21	B <sub>20,14</sub>	General purpose I/O 27
GPIO_26	F22	B <sub>20,14</sub>	General purpose I/O 26
GPIO_25	V3	B <sub>20,14</sub>	General purpose I/O 25
GPIO_24	U3	B <sub>20,14</sub>	General purpose I/O 24
GPIO_23	U2	B <sub>20,14</sub>	General purpose I/O 23
GPIO_22	U1	B <sub>20,14</sub>	General purpose I/O 22
GPIO_21	T3	B <sub>20,14</sub>	General purpose I/O 21
GPIO_20	T2	B <sub>20,14</sub>	General purpose I/O 20
GPIO_19	T1	B <sub>20,14</sub>	General purpose I/O 19
GPIO_18	R3	B <sub>20,14</sub>	General purpose I/O 18
GPIO_17	R2	B <sub>20,14</sub>	General purpose I/O 17
GPIO_16	R1	B <sub>20,14</sub>	General purpose I/O 16
GPIO_15	P3	B <sub>20,14</sub>	General purpose I/O 15
GPIO_14	P2	B <sub>20,14</sub>	General purpose I/O 14
GPIO_13	P1	B <sub>20,14</sub>	General purpose I/O 13
GPIO_12	N3	B <sub>20,14</sub>	General purpose I/O 12
GPIO_11	N2	B <sub>20,14</sub>	General purpose I/O 11
GPIO_10	N1	B <sub>20,14</sub>	General purpose I/O 10
GPIO_09	M3	B <sub>20,14</sub>	General purpose I/O 09
GPIO_08	M2	B <sub>20,14</sub>	General purpose I/O 08
GPIO_07	M1	B <sub>20,14</sub>	General purpose I/O 07
GPIO_06	L3	B <sub>20,14</sub>	General purpose I/O 06
GPIO_05	L2	B <sub>20,14</sub>	General purpose I/O 05
GPIO_04	L1	B <sub>20,14</sub>	General purpose I/O 04
GPIO_03	K3	B <sub>20,14</sub>	General purpose I/O 03
GPIO_02	K2	B <sub>20,14</sub>	General purpose I/O 02
GPIO_01	K1	B <sub>20,14</sub>	General purpose I/O 01
GPIO_00	J3	B <sub>20,14</sub>	General purpose I/O 00

- (1) Some GPIO signals are reserved for specific purposes. These signals vary per product configuration. These product allocations are discussed further in [Section 8.3.7](#). All GPIO that are available for Host use must be configured as an input, a standard output, or an open-drain output. This is set in the flash configuration or by command using the Host command interface. The reset default for all GPIO is as an input signal. An external pull-up ( $\leq 10\text{ k}\Omega$ ) is required for each signal configured as open-drain.
- (2) See [Table 5-1](#) for more information on I/O definitions.
- (3) All GPIO include hysteresis.

## Pin Functions – Clock and PLL Support

PIN		I/O <sup>(1)</sup>	DESCRIPTION
NAME	NUMBER		
PLL_REFCLK_I	D15	I <sub>17</sub>	Reference clock crystal input. <i>If an external oscillator is used in place of a crystal, this pin should be left unconnected (floating with no added capacitive load).</i>
PLL_REFCLK_O	D14	B <sub>16,17</sub>	Reference clock crystal return. <i>If an external oscillator is used in place of a crystal, this pin must be used for the oscillator input.</i>
OSC_BYPASS	D16	I <sub>19</sub>	Selects whether an external crystal or external oscillator will be used to drive the internal PLL. (‘0’ = Crystal, ‘1’ = Oscillator) This pin includes a weak internal pull-down. If a pull-up is being used to obtain a ‘1’ value, the pull-up value must be ≤ 8 kΩ.

(1) See [Table 5-1](#) for more information on I/O definitions.

## Pin Functions – Power and Ground

PIN		I/O <sup>(1)</sup>	DESCRIPTION
NAME	NUMBER		
VCC18A_LVDS	B1, B22, C1, C22, D2, D3, D4, D5, D7, D18, D19, D20, D21, E20	PWR	1.8-V Power for the differential High-Speed and Low-Speed DMD Interfaces
GND18A_LVDS	A1, A22, C2, C3, C4, C5, C6, C7, C16, C17, C18, C19, C20, C21, D8	RTN	1.8-V GND for the differential High-Speed and Low-Speed DMD Interfaces
VCC18IO	D10	PWR	1.8-V Power for 1.8-V IO
VCC3IO_MVGP	H4	PWR	3.3-V Power for TPS99000S-Q1 Interfaces
VCC3IO_FLSH	V4	PWR	3.3-V Power for the Serial Flash Interface
VCC3IO_INTF	K19, L19, M19, R19, T19	PWR	3.3-V Power for the Parallel Data, JTAG, and Host Command Interfaces
VCC3IO_COSC	C15	PWR	3.3-V I/O Power for the Crystal Oscillator
GNDIOLA_COSC	C14	RTN	3.3-V I/O GND for the Crystal Oscillator
VCC3IO	J4, K4, M4, N4, P4, W4, W5, G19	PWR	3.3-V I/O Power for all "other" I/O (such as GPIO, TSTPT, PMIC_AD3)
VCC33A_LVDS	W9, W13, W15, W19, Y9, Y13, Y15, Y19	PWR	3.3-V I/O Power for the OpenLDI Interface
GND33A_LVDS	W14, Y14, AA8, AA14, AA20, AB8, AB14, AB20, AB21	RTN	3.3-V I/O GND for the OpenLDI Interface
VCC11AD_PLLM	D13	PWR	1.1-V Analog/Digital Power for MCG (Master Clock Generator) PLL
GND11AD_PLLM	C13	RTN	1.1-V Analog/Digital GND for MCG (Master Clock Generator) PLL
VCC11AD_PLLD	D12	PWR	1.1-V Analog/Digital Power for DCG (DMD Clock Generator) PLL
GND11AD_PLLD	C12	RTN	1.1-V Analog/Digital GND for DCG (DMD Clock Generator) PLL
VCC11A_DDI_0	E19, F19	PWR	1.1-V Filtered Core Power - External Filter Group A (HS DMD Interface 0)
VCC11A_DDI_1	E4, F4	PWR	1.1-V Filtered Core Power - External Filter Group B (HS DMD Interface 1)
VCC11A_LVDS	W11, W12, W17, W18	PWR	1.1-V Filtered Core Power - External Filter Group C (OpenLDI Interface)
VCCK	G4, H19, (J11), J19, L4, N19, P19, T4, U4, U19, V19, W6, W8, W10, W16	PWR	1.1-V Core Power (Ball numbers in parenthesis are also used as thermal ball and are located within the package center region)

PIN		I/O <sup>(1)</sup>	DESCRIPTION
NAME	NUMBER		
GND	(J9, J10, J12, J13, J14, K9, K10, K11, K12, K13, K14, L9, L10, L11, L12, L13, L14, M9, M10, M11, M12, M13, M14, N9, N10, N11, N12, N13, N14, P9, P10, P11, P12, P13, P14), Y3, AA1, AA2, AB1, AB2, AB22, Y10, Y11, Y12, Y16, Y17, Y18	RTN	1.1-V Core GND (Ball numbers in parenthesis are also used as thermal ball and are located within the package center region)
EFUSE_VDDQ	W7		Manufacturing use only. Must be tied to ground.
EFUSE_POR33	Y8		Manufacturing use only. Must be tied to ground.
RPI_0	D17	I <sub>5</sub>	Bandgap Reference for sub-LVDS drivers (Supports DMD_HS0_xxxx). Requires a resistor (1% Tolerance) to GND18A_LVDS - Value specified in <a href="#">Table 11-4</a> .
RPI_1	D6	I <sub>5</sub>	Bandgap Reference for sub-LVDS drivers (Supports DMD_HS1_xxxx). Requires a resistor (1% Tolerance) to GND18A_LVDS - Value specified in <a href="#">Table 11-4</a> .
RPI_LS	D9	I <sub>5</sub>	Bandgap References for sub-LVDS drivers (Supports DMD_LS0_xxxx differential bus signals). Requires a resistor (1% Tolerance) to GND18A_LVDS - Value specified in <a href="#">Table 11-4</a> .

(1) See [Table 5-1](#) for more information on I/O definitions.

**Table 5-1. I/O Type Subscript Definition**

I/O		SUPPLY REFERENCE	ESD STRUCTURE
SUBSCRIPT	DESCRIPTION		
1	1.8-V LVCMOS Input	VCC18IO	ESD diode to GND and supply rail
2	1.8-V LVCMOS Output	VCC18IO	ESD diode to GND and supply rail
3	1.8-V LVCMOS Input	VCC18IO	ESD diode to GND and supply rail
4	1.8-V sub-LVDS Output	VCC18A_LVDS	ESD diode to GND and supply rail
5	1.8-V sub-LVDS Input	VCC18A_LVDS	ESD diode to GND and supply rail
6	3.3-V LVCMOS Output	VCC3IO_MVGP	ESD diode to GND and supply rail
7	3.3-V LVCMOS Input	VCC3IO_MVGP	ESD diode to GND and supply rail
8	3.3-V LVCMOS Output	VCC3IO_FLSH	ESD diode to GND and supply rail
9	3.3-V LVCMOS Input	VCC3IO_FLSH	ESD diode to GND and supply rail
10	3.3-V LVCMOS Output	VCC3IO_INTF	ESD diode to GND and supply rail
11	3.3-V LVCMOS Input	VCC3IO_INTF	ESD diode to GND and supply rail
12	3.3-V I <sup>2</sup> C I/O	VCC3IO_INTF	ESD diode to GND and supply rail
13	3.3-V LVCMOS Output	VCC3IO	ESD diode to GND and supply rail
14	3.3-V LVCMOS Input	VCC3IO	ESD diode to GND and supply rail
15	3.3-V I <sup>2</sup> C I/O with 3-mA drive	VCC3IO	ESD diode to GND and supply rail
16	3.3-V LVCMOS Output	VCC3IO_OSC	ESD diode to GND and supply rail
17	3.3-V LVCMOS Input	VCC3IO_OSC	ESD diode to GND and supply rail
18	3.3-V LVDS Input	VCC33A_LVDS	ESD diode to GND and supply rail
19	3.3-V LVCMOS Input	VCC3IO_OSC	ESD diode to GND and supply rail
20	3.3-V LVCMOS Output	VCC3IO	ESD diode to GND and supply rail
<b>TYPE</b>			
I	Input		N/A
O	Output		
B	Bidirectional		
PWR	Power		
RTN	Ground return		

**Table 5-2. Internal Pull-up and Pull-down Characteristics**

INTERNAL PULL-UP AND PULL-DOWN RESISTOR CHARACTERISTICS <sup>(1)</sup> <sup>(2)</sup>	VCCIO	MIN	MAX	UNIT
Weak pull-up resistance	3.3 V	40	190	kΩ
Weak pull-down resistance	3.3 V	30	190	kΩ

- (1) The resistance is dependent on the supply voltage level applied to the I/O.  
(2) An external 8-kΩ or less pull-up or pull-down (if needed) will work for any voltage condition to correctly override any associated internal pull-ups or pull-downs.

## 6 Specifications

### 6.1 Absolute Maximum Ratings

over operating free-air temperature (unless otherwise noted)<sup>(1)</sup>

		MIN	MAX	UNIT
<b>SUPPLY VOLTAGE<sup>(2)</sup></b>				
$V_{(VCC)}$ (Core)		-0.5	1.5	V
$V_{(VCC11A\_DDIx)}$ (Core)		-0.5	1.5	V
$V_{(VCC11A\_LVDS)}$ (Core)		-0.5	1.5	V
$V_{(VCC11AD\_PLLM)}$ (Core)		-0.5	1.5	V
$V_{(VCC11AD\_PLLD)}$ (Core)		-0.5	1.5	V
$V_{(VCC18A\_LVDS)}$		-0.5	2.5	V
$V_{(VCC18IO)}$		-0.5	2.5	V
$V_{(VCC3IO\_MVGP)}$		-0.5	4.6	V
$V_{(VCC3IO\_INF)}$		-0.5	4.6	V
$V_{(VCC3IO\_FLSH)}$		-0.5	4.6	V
$V_{(VCC3IO\_OSC)}$		-0.5	4.6	V
$V_{(VCC3IO)}$		-0.5	4.6	V
$V_{(VCC33A\_LVDS)}$		-0.5	4.6	V
<b>GENERAL</b>				
$T_J$	Operating junction temperature	-40	125	°C
$T_C$	Operating case temperature	-40	124 <sup>(3)</sup>	°C
$I_{lat}$	Latch-up	-100	100	mA
$T_{stg}$	Storage temperature range	-40	150	°C

(1) Stresses beyond those listed under [Section 6.1](#) may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under [Section 6.3](#). Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

(2) All voltage values are with respect to GND.

(3) Value calculated using package parameters defined in [Section 6.4](#).

### 6.2 ESD Ratings

			VALUE	UNIT	
$V_{(ESD)}$	Electrostatic discharge	Human-body model (HBM), per AEC Q100-002 <sup>(1)</sup>	±2000	V	
		Charged-device model (CDM), per AEC Q100-011	All pins (except corner pins)		±500
			Corner pins (A1, A22, AB0, and AB22) only		±750

(1) AEC Q100-002 indicates that HBM stressing shall be in accordance with the ANSI/ESDA/JEDEC JS-001 specification.



## 6.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

			MIN	NOM	MAX	UNIT
V <sub>(VCCK)</sub>	Core power 1.1 V (main 1.1 V)	±5% tolerance	1.045	1.1	1.155	V
V <sub>(VCC11A_DDL_0)</sub>	Core power 1.1 V (External Filter Group A - HS DMD Interface 0)	±8.18% tolerance <sup>(1)</sup>	1.01	1.1	1.19	V
V <sub>(VCC11A_DDL_1)</sub>	Core power 1.1 V (External Filter Group B - HS DMD Interface 1)	±8.18% tolerance <sup>(1)</sup>	1.01	1.1	1.19	V
V <sub>(VCC11A_LVDS)</sub>	Core power 1.1 V (External Filter Group C - OpenLDI Interface)	±8.18% tolerance <sup>(1)</sup>	1.01	1.1	1.19	V
V <sub>(VCC11AD_PLLM)</sub>	MCG PLL 1.1-V power (Analog/Digital)	±8.18% tolerance <sup>(1)</sup>	1.01	1.1	1.19	V
V <sub>(VCC11AD_PLLD)</sub>	DCG PLL 1.1-V power (Analog/Digital)	±8.18% tolerance <sup>(1)</sup>	1.01	1.1	1.19	V
V <sub>(VCC18IO)</sub>	1.8-V I/O power (Supports DMD Single-Ended LS interface I/O)	±8.3% tolerance	1.65	1.8	1.95	V
V <sub>(VCC18A_LVDS)</sub>	1.8-V I/O power (Supports High-Speed and Low-Speed differential DMD interfaces)	±8.3% tolerance	1.65	1.8	1.95	V
V <sub>(VCC3IO_MVGP)</sub>	3/3-V I/O power (Supports TPS99000S-Q1: SPI, interrupt, park, RESETZ, and LEDSEL interfaces)	±8.5% tolerance	3.02	3.3	3.58	V
V <sub>(VCC3IO_FLSH)</sub>	3/3-V I/O power (Supports serial flash interface)	±8.5% tolerance	3.02	3.3	3.58	V
V <sub>(VCC3IO_INTF)</sub>	3.3-V I/O power (Supports: host command (SPI and I <sup>2</sup> C), parallel data interface, HOST_IRQ, and JTAG)	±8.5% tolerance	3.02	3.3	3.58	V
V <sub>(VCC3IO_OSC)</sub>	3.3-V I/O power (Supports Oscillator)	±8.5% tolerance	3.02	3.3	3.58	V
V <sub>(VCC33A_LVDS)</sub>	3.3-V I/O power (Supports OpenLDI interface)	±8.5% tolerance	3.02	3.3	3.58	V
V <sub>(VCC3IO)</sub>	3.3-V I/O power (Supports all remaining I/O including: GPIO, PMIC_AD3, TSTPT, ETM_TRACE, et cetera)	±8.5% tolerance	3.02	3.3	3.58	V
T <sub>J</sub>	Operating junction temperature		–40		125	°C
T <sub>C</sub>	Operating case temperature		–40		124	°C
T <sub>A</sub>	Operating ambient temperature <sup>(2)</sup>		–40		105	°C

(1) These I/O supply ranges are wider to facilitate additional external filtering.

(2) Operating ambient temperature is dependent on system thermal design. Operating case temperature may not exceed its specified range across ambient temperature conditions.

## 6.4 Thermal Information

THERMAL METRIC <sup>(1)</sup>		DLPC230S-Q1	UNIT
		ZDQ (BGA)	
		324 PINS	
Ψ <sub>JT</sub> <sup>(2)</sup>	Temperature variance from junction to package top center temperature, per unit power dissipation	0.77	°C/W

(1) For more information about traditional and new thermal metrics, see the *Semiconductor and IC Package Thermal Metrics* application report, [SPRA953](#).

(2) (1.22 W) × (0.77°C/W) ≈ 1.00°C temperature difference.

## 6.5 Electrical Characteristics

over operating free-air temperature range (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP <sup>(1)</sup>	MAX <sup>(2)</sup>	UNIT
<b>TOTAL</b>						
$I_{(VCC11)}$	1.1-V total current			201	743.9	mA
$I_{(VCC18)}$	1.8-V total current			71	122.9	mA
$I_{(VCC33)}$	3.3-V total current			28.1	30.1	mA
<b>ESTIMATED CURRENT PER SUPPLY<sup>(3)</sup></b>						
$I_{(VCCK)}$	1.1-V Core current			131.5	667.5	mA
$I_{(VCC11A\_DDI\_0)}$	1.1-V Core current (Filtered)	At 600-MHz data rate		15.8	17.4	mA
$I_{(VCC11A\_DDI\_1)}$	1.1-V Core current (Filtered)	At 600-MHz data rate		15.8	17.4	mA
$I_{(VCC11A\_LVDS)}$	1.1-V Core current (Filtered)	OpenLDI Interface, single port, 5 lanes active		22.5	24.8	mA
$I_{(VCC11AD\_PLL)}$	1.1-V Core current (MCG PLL)			7.7	8.4	mA
$I_{(VCC11AD\_PLLD)}$	1.1-V Core current (DCG PLL)			7.7	8.4	mA
$I_{(VCC18A\_LVDS)}$	1.8-V I/O current (Both 8-bit ports - DMD HS differential Interface)	At 600-MHz data rate		63.3	106.6	mA
$I_{(VCC18A\_LVDS)}$	1.8-V I/O current (DMD LS differential Interface)	At 120-MHz data rate		5.2	8.7	mA
$I_{(VCC18IO)}$	1.8-V I/O current (DMD LS single-ended interfaces, DMD reset)			2.5	7.6	mA
$I_{(VCC3IO\_MVGP)}$	3.3-V I/O current (TPS99000S-Q1 SPI, TPS99000S-Q1 Reset, PMIC_PARKZ, RESETZ)			1.7	1.8	mA
$I_{(VCC3IO\_INTF)}$	3.3-V I/O current (Host SPI, Host I <sup>2</sup> C, Host IRQ, JTAG, Parallel Port)			1.7	1.8	mA
$I_{(VCC3IO\_FLSH)}$	3.3-V I/O current (Serial Flash SPI interface)			5.5	5.9	mA
$I_{(VCC3IO\_OSC)}$	3.3-V I/O current (Crystal/Oscillator)	With 3-k $\Omega$ external series resistor (R <sub>S</sub> )		0.975	1.3	mA
$I_{(VCC3IO)}$	3.3-V I/O current (GPIO, PMIC_AD3, Mstr I <sup>2</sup> C, TSTPT, ETM, and so forth)			12.6	13.5	mA
$I_{(VCC33A\_LVDS)}$	3.3-V I/O current (OpenLDI Interface - each port - 5 lanes active)			6.3	6.8	mA

- (1) Typical-case power measured with PVT condition = nominal process, typical voltage, typical temperature (25°C junction). Input source 1152 × 576 24-bit 60-Hz OpenLDI with RGBW ramp image.
- (2) Worst-case power PVT condition = corner process, high voltage, high temperature (125°C junction). Input source 1152 × 1152 24-bit. 60 Hz OpenLDI with pseudo-random noise image.
- (3) Estimated current per supply was not directly measured. These values are based on an approximate expected current consumption percentage of the total measured current drawn by each voltage rail.

## 6.6 Electrical Characteristics for Fixed Voltage I/O

over operating free-air temperature range (unless otherwise noted)<sup>(1)</sup>

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
V <sub>IH</sub>	High-level input threshold voltage	1.8-V LVCMOS (I/O type 3)			0.7 × VCC18IO	V
		3.3-V LVCMOS (I/O type 7)			2.0	
		3.3-V LVCMOS (I/O type 9)			2.0	
		3.3-V LVCMOS (I/O type 11)			2.0	
		3.3-V I <sup>2</sup> C buffer (I/O type 12)			0.7 × VCC_INTF	
		3.3-V LVCMOS (I/O type 14)			2.0	
		3.3-V LVCMOS (I/O type 16,17)			0.7 × VCC3IO	
		3.3-V LVCMOS (I/O type 19)			2.0	
		3.3-V I <sup>2</sup> C buffer (I/O type 15)			0.7 × VCC3IO	
V <sub>IL</sub>	Low-level input threshold voltage	1.8-V LVCMOS (I/O type 3)			0.3 × VCC18IO	V
		3.3-V LVCMOS (I/O type 7)			0.8	
		3.3-V LVCMOS (I/O type 9)			0.8	
		3.3-V LVCMOS (I/O type 11)			0.8	
		3.3-V I <sup>2</sup> C buffer (I/O type 12)			0.3 × VCC_INTF	
		3.3-V LVCMOS (I/O type 14)			0.8	
		3.3-V LVCMOS (I/O type 16,17)			0.3 × VCC3IO	
		3.3-V LVCMOS (I/O type 19)			0.8	
		3.3-V I <sup>2</sup> C buffer (I/O type 15)			0.3 × VCC3IO	
V <sub>OH</sub>	High-level output voltage	1.8-V LVCMOS (I/O type 1,2)	I <sub>OH</sub> = Max rated		0.75 × VCC18IO	V
		3.3-V LVCMOS (I/O type 6)	I <sub>OH</sub> = Max rated		2.4	
		3.3-V LVCMOS (I/O type 8)	I <sub>OH</sub> = Max rated		2.4	
		3.3-V LVCMOS (I/O type 10)	I <sub>OH</sub> = Max rated		2.4	
		3.3-V I <sup>2</sup> C buffer (I/O type 12)	I <sub>OH</sub> = Max rated		N/A	
		3.3-V LVCMOS (I/O type 13)	I <sub>OH</sub> = Max rated		2.4	
		3.3-V I <sup>2</sup> C buffer (I/O type 15)	I <sub>OH</sub> = Max rated		N/A	
		3.3-V LVCMOS (I/O type 20)	I <sub>OH</sub> = Max rated		2.4	
V <sub>OL</sub>	Low-level output voltage	1.8-V LVCMOS (I/O type 1,2)	I <sub>OL</sub> = Max rated		0.4	V
		3.3-V LVCMOS (I/O type 6)	I <sub>OL</sub> = Max rated		0.4	
		3.3-V LVCMOS (I/O type 8)	I <sub>OL</sub> = Max rated		0.4	
		3.3-V LVCMOS (I/O type 10)	I <sub>OL</sub> = Max rated		0.4	
		3.3-V I <sup>2</sup> C buffer (I/O type 12)	I <sub>OL</sub> = Max rated		0.4	
		3.3-V LVCMOS (I/O type 13)	I <sub>OL</sub> = Max rated		0.4	
		3.3-V I <sup>2</sup> C buffer (I/O type 15)	I <sub>OL</sub> = Max rated		0.4	
		3.3-V LVCMOS (I/O type 20)	I <sub>OL</sub> = Max rated		0.4	
I <sub>OH</sub>	High-level output current	1.8-V LVCMOS (I/O type 1)			6	mA
		1.8-V LVCMOS (I/O type 2)			7.2	
		3.3-V LVCMOS (I/O type 6)			6	
		3.3-V LVCMOS (I/O type 8)			6	
		3.3-V LVCMOS (I/O type 10)			6	
		3.3-V I <sup>2</sup> C buffer (I/O type 12)			N/A	
		3.3-V LVCMOS (I/O type 13)			8	
		3.3-V I <sup>2</sup> C buffer (I/O type 15)			N/A	
		3.3-V LVCMOS (I/O type 20)			6	

over operating free-air temperature range (unless otherwise noted)<sup>(1)</sup>

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
I <sub>OL</sub>	Low-level output current	1.8-V LVCMOS (I/O type 1)		6		mA
		1.8-V LVCMOS (I/O type 2)		7.2		
		3.3-V LVCMOS (I/O type 6)		6		
		3.3-V LVCMOS (I/O type 8)		6		
		3.3-V LVCMOS (I/O type 10)		6		
		3.3-V I <sup>2</sup> C buffer (I/O type 12)		3		
		3.3-V LVCMOS (I/O type 13)		8		
		3.3-V I <sup>2</sup> C buffer (I/O type 15)		3		
		3.3-V LVCMOS (I/O type 20)		6		
I <sub>OZ</sub>	High-impedance leakage current	1.8-V LVCMOS (I/O type 1,2)		±1.0	±10	μA
		3.3-V LVCMOS (I/O type 6)		±1.0	±10	
		3.3-V LVCMOS (I/O type 8)		±1.0	±10	
		3.3-V LVCMOS (I/O type 10)		±1.0	±10	
		3.3-V I <sup>2</sup> C buffer (I/O type 12)			±10	
		3.3-V LVCMOS (I/O type 13)		±1.0	±10	
		3.3-V LVCMOS (I/O type 16)		±1.0		
		3.3-V I <sup>2</sup> C buffer (I/O type 15)			±10	
3.3-V LVCMOS (I/O type 20)		±1.0	±10			

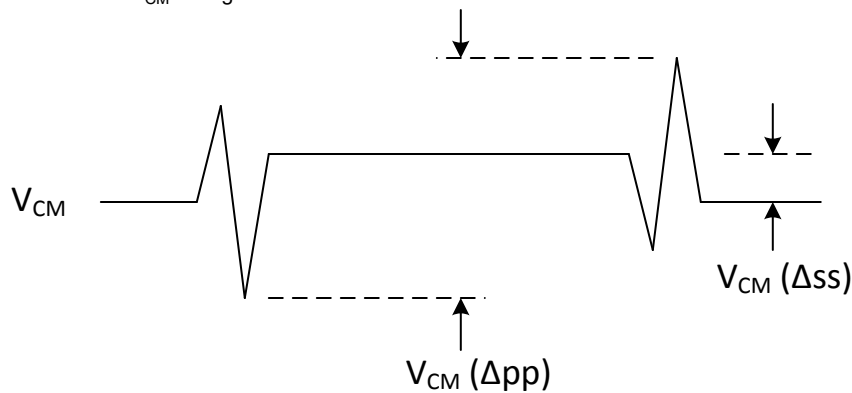
(1) The number inside each parenthesis for the I/O refers to the type defined in [Table 5-1](#).

## 6.7 DMD High-Speed Sub-LVDS Electrical Characteristics

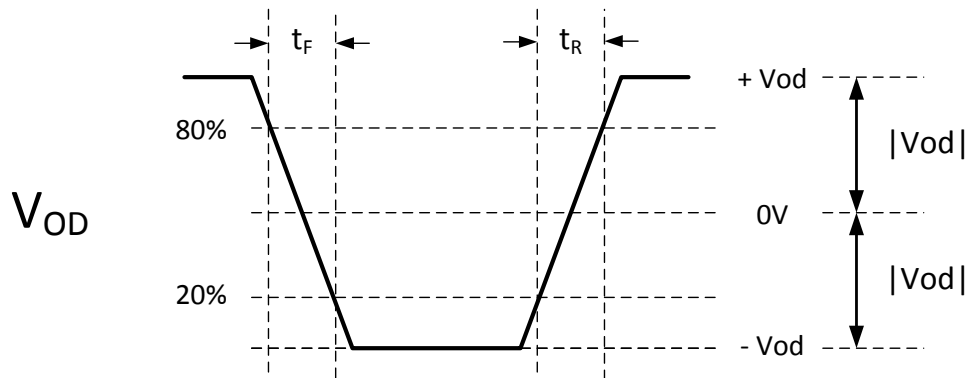
over operating free-air temperature range (unless otherwise noted)

PARAMETER		MIN	NOM	MAX	UNIT	
$V_{CM}$	Steady-state common mode voltage	1.8-V sub-LVDS (I/O type 4,5)	0.8	0.9	1.0	V
$V_{CM} (\Delta pp)^{(1)}$	$V_{CM}$ change peak-to-peak (during switching)	1.8-V sub-LVDS (I/O type 4,5)			75	mV
$V_{CM} (\Delta ss)^{(1)}$	$V_{CM}$ change steady state	1.8-V sub-LVDS (I/O type 4,5)			10	mV
$ V_{OD} ^{(2)}$	Differential output voltage magnitude. $R_{BGR} = 75k\Omega$ .	1.8-V sub-LVDS (I/O type 4,5)	155	200	250	mV
$V_{OD} (\Delta)^{(3)}$	$V_{OD}$ change (between logic states)	1.8-V sub-LVDS (I/O type 4,5)			10	mV
$V_{OH}$	Single-ended output voltage high	1.8-V sub-LVDS (I/O type 4,5)	0.88	1.00	1.125	V
$V_{OL}$	Single-ended output voltage low	1.8-V sub-LVDS (I/O type 4,5)	0.675	0.80	0.925	V
$t_R^{(2)}$	Differential output rise time	1.8-V sub-LVDS (I/O type 4,5)			250	ps
$t_F^{(2)}$	Differential output fall time	1.8-V sub-LVDS (I/O type 4,5)			250	ps
$f_{MAX}$	Max switching rate	1.8-V sub-LVDS (I/O type 4,5)			1200	Mbps
DCout	Output duty cycle	1.8-V sub-LVDS (I/O type 4,5)	45%	50%	55%	
$T_{Xterm}^{(1)}$	Internal differential termination	1.8-V sub-LVDS (I/O type 4,5)	80	100	120	$\Omega$

(1) Definition of  $V_{CM}$  changes:



(2) Note that  $V_{OD}$  is the differential voltage swing measured across a 100- $\Omega$  termination resistance connected directly between the transmitter differential pins.  $|V_{OD}|$  is the magnitude of the peak to peak voltage swing across the P and N output pins. Since  $V_{CM}$  cancels out when measured differentially,  $V_{OD}$  voltage swings relative to 0. Rise and fall times are defined for the differential  $V_{OD}$  signal as follows:



### Differential Output Signal

(Note:  $V_{CM}$  is removed when signals are viewed differentially)

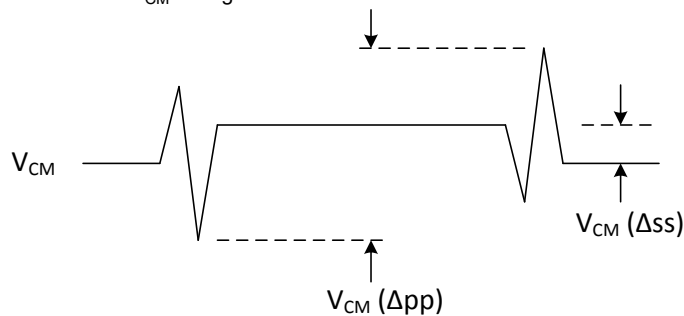
- (3) When TX data input = '1', differential output voltage  $V_{OD1}$  is defined. When TX data input = '0', differential output voltage  $V_{OD0}$  is defined. As such, the steady state magnitude of the difference is:  $|V_{OD}| (\Delta) = ||V_{OD1}| - |V_{OD0}|$ .

## 6.8 DMD Low-Speed Sub-LVDS Electrical Characteristics

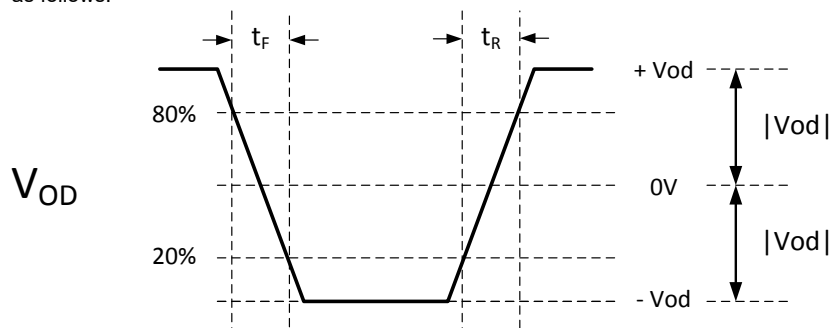
over operating free-air temperature range (unless otherwise noted)

PARAMETER			MIN	NOM	MAX	UNIT
$V_{CM}$	Steady-state common mode voltage	1.8-V sub-LVDS (I/O type 4,5)	0.8	0.9	1.0	V
$V_{CM} (\Delta pp)^{(1)}$	$V_{CM}$ change peak-to-peak (during switching)	1.8-V sub-LVDS (I/O type 4,5)			75	mV
$V_{CM} (\Delta ss)^{(1)}$	$V_{CM}$ change steady state	1.8-V sub-LVDS (I/O type 4,5)	-10		10	mV
$ V_{OD} ^{(2)}$	Differential output voltage magnitude. $R_{BGR} = 75k\Omega$ .	1.8-V sub-LVDS (I/O type 4,5)	155	200	250	mV
$V_{OD} (\Delta)^{(3)}$	$V_{OD}$ change (between logic states)	1.8-V sub-LVDS (I/O type 4,5)	-10		10	mV
$V_{OH}$	Single-ended output voltage high	1.8-V sub-LVDS (I/O type 4,5)	0.88	1.00	1.125	V
$V_{OL}$	Single-ended output voltage low	1.8-V sub-LVDS (I/O type 4,5)	0.675	0.80	0.925	V
$t_R^{(2)}$	Differential output rise time	1.8-V sub-LVDS (I/O type 4,5)			250	ps
$t_F^{(2)}$	Differential output fall time	1.8-V sub-LVDS (I/O type 4,5)			250	ps
$t_{MAX}$	Max switching rate	1.8-V sub-LVDS (I/O type 4,5)			240	Mbps
DCout	Output duty cycle	1.8-V sub-LVDS (I/O type 4,5)	45%	50%	55%	
$TX_{term}$	Internal differential termination	1.8-V sub-LVDS (I/O type 4,5)	80	100	120	$\Omega$

(1) Definition of  $V_{CM}$  changes:



(2) Note that  $V_{OD}$  is the differential voltage swing measured across a 100- $\Omega$  termination resistance connected directly between the transmitter differential pins.  $|V_{OD}|$  is the magnitude of the peak to peak voltage swing across the P and N output pins. Since  $V_{CM}$  cancels out when measured differentially,  $V_{OD}$  voltage swings relative to 0. Rise and fall times are defined for the differential  $V_{OD}$  signal as follows:



### Differential Output Signal

(Note:  $V_{CM}$  is removed when signals are viewed differentially)

(3) When TX data input = '1', differential output voltage  $V_{OD1}$  is defined. When TX data input = '0', differential output voltage  $V_{OD0}$  is defined. As such, the steady state magnitude of the difference is:  $|V_{OD}| (\Delta) = ||V_{OD1}| - |V_{OD0}||$ .

## 6.9 OpenLDI LVDS Electrical Characteristics

over operating free-air temperature range (unless otherwise noted)

PARAMETER			MIN	NOM	MAX	UNIT
$V_{CM}$	Steady-state common mode voltage	3.3-V LVDS (I/O type 18)	0.35	1.2	1.6	V
$ V_{ID} $	Differential Input Voltage	3.3-V LVDS (I/O type 18)	100		700	mV
$R_{Xterm}$	Internal differential termination	3.3-V LVDS (I/O type 18)	90	111	132	$\Omega$

## 6.10 Power Dissipation Characteristics

PARAMETER	VALUE	UNIT	
$P_{MAX}$	Package - Maximum Power	1.22	W

## 6.11 System Oscillators Timing Requirements

			MIN	NOM	MAX	UNIT
$f_{clock}$	Clock frequency, MOSC <sup>(1)</sup>		15.997	16.000	16.003	MHz
$t_c$	Cycle time, MOSC <sup>(1)</sup>		62.488	62.500	62.512	ns
$t_{w(H)}$	Pulse duration <sup>(2)</sup> , MOSC, high	50% to 50% reference points (signal)	40% of $t_c$			
$t_{w(L)}$	Pulse duration <sup>(2)</sup> , MOSC, low	50% to 50% reference points (signal)	40% of $t_c$			
$t_t$	Transition time <sup>(2)</sup> , MOSC, $t_t = t_f / t_r$	20% to 80% reference points (signal)	0.2		2	ns
$t_{jp}$	Long term periodic jitter <sup>(2)</sup> , MOSC (that is the deviation in period from ideal period due solely to high frequency jitter)				100	ps

(1) The MOSC input cannot support spread spectrum clock spreading.

(2) Applies only when driven through an external digital oscillator. This is a 1 sigma RMS value.

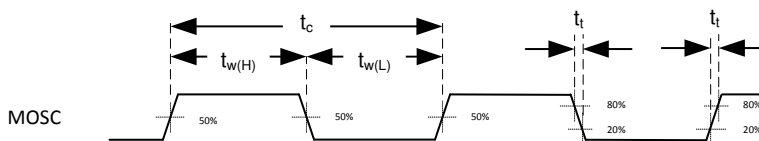


Figure 6-1. System Oscillators

Table 6-1. Crystal / Oscillator Electrical Characteristics

PARAMETER	NOMINAL	UNIT
PLL_REFCLK_I TO GND capacitance	3.5	pF
PLL_REFCLK_O TO GND capacitance	3.45	pF



## 6.12 Power Supply and Reset Timing Requirements

			MIN	MAX	UNIT
<b>TPS99000S-Q1 REQUIREMENTS<sup>(1)</sup></b>					
$t_{ramp}$	Power supply ramp time <sup>(2)</sup>	Power supply ramp to minimum recommended operating voltage	0.5	10	ms
$t_{ps\_aln}$	1.1-V Power Supply Alignment <sup>(3)</sup>	Leading edge for application or removal of power. Each 1.1-V power supply to the DLPC230S-Q1 must be applied simultaneously within this time.		10	$\mu$ s
$t_{rst}$	RESETZ low to Power Supply disable <sup>(4)</sup>	Leading edge for removal of power	1.0		$\mu$ s
$t_{w(L1)}$	Pulse duration, active low, RESETZ <sup>(4)</sup>	95% power to 50% RESETZ reference point At initial application of power#unique_11/ unique_11_Connect_42_DLPS0543246753 8	5.0		ms
$t_{w(L2)}$	Pulse duration, active low, RESETZ	50% to 50% reference points (RESETZ) Subsequent resets after initial application of power	1.0		$\mu$ s
$t_t$	Transition time, RESETZ, $t_t = t_f$ and $t_r$	20% to 80% reference points (signal)		6	$\mu$ s

- (1) The TPS99000S-Q1 controls power supply timing for the DLPC230S-Q1. Refer to the TPS99000S-Q1 data sheet for additional system power timing requirements.
- (2) Power supplies do not need to ramp simultaneously, but each supply must reach its minimum voltage within the maximum ramp time specified.
- (3) The DLPC230S-Q1 does not require specific sequencing or alignment of 1.8-V and 3.3-V supplies. However, the TPS99000S-Q1 enforces sequencing of the 1.1-V, 1.8-V, and 3.3-V voltage rails. The following describes DLPC230S-Q1 behavior when the voltage rails are not brought up simultaneously:
  - VCKK (1.1-V core) Power = On, I/O Power = Off, RESETZ = '0': While this condition exists, additional leakage current may be drawn, and all outputs are unknown (likely to be a weak "low").
  - VCKK (1.1-V core) Power = Off, I/O Power = On, RESETZ = '0': While this condition exists all outputs are tri-stated.
 Neither of these two conditions will impact normal DLPC230S-Q1 reliability.
- (4) RESETZ must be held low if any supply (Core or I/O) is less than its minimum specified on value. For more information on RESETZ, see [Section 5](#).

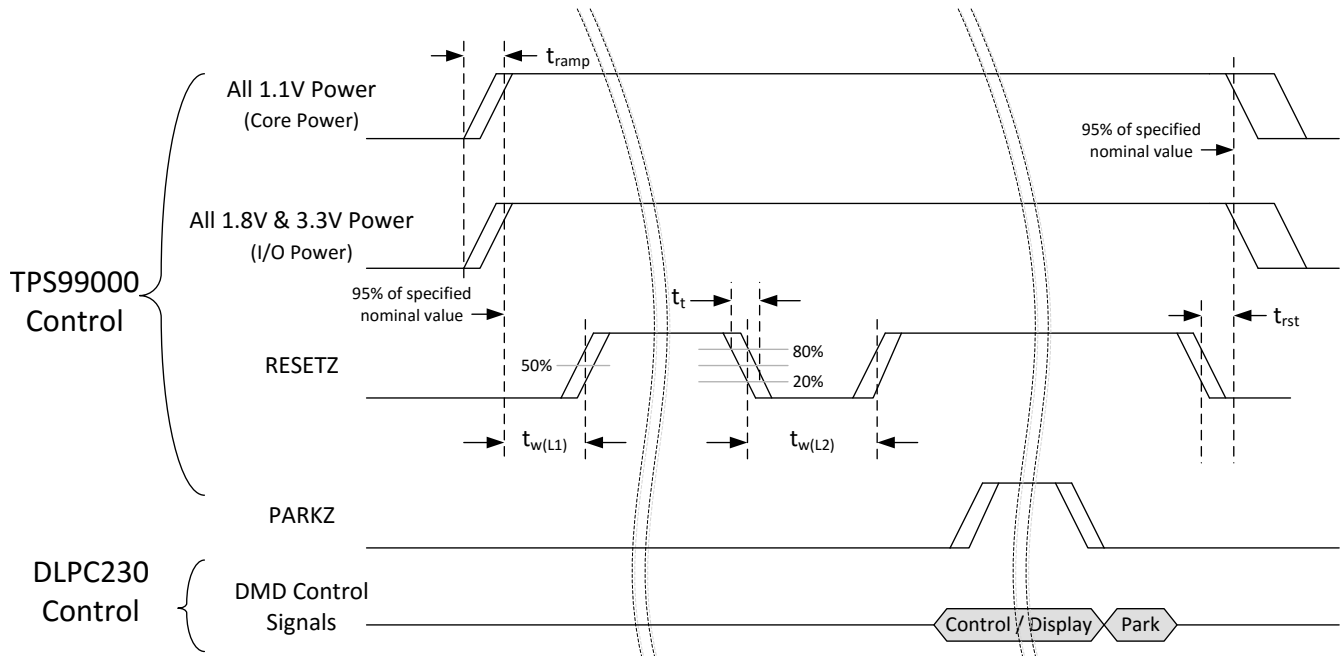


Figure 6-2. Power Supply and RESETZ Timing

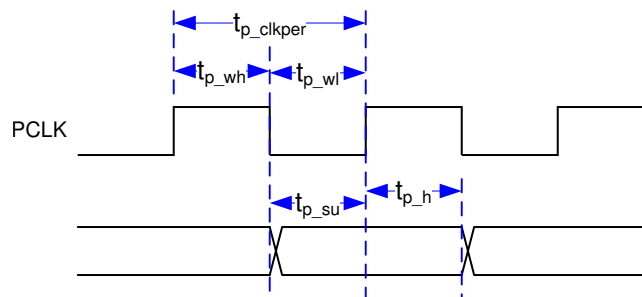
## 6.13 Parallel Interface General Timing Requirements

		MIN	MAX	UNIT	
$f_{\text{clock}}$	Clock frequency, PCLK	12.0	110.0	MHz	
$t_{\text{p\_clkper}}$	Clock period, PCLK	50% reference points	9.091	83.33	ns
$t_{\text{p\_wh}}$	Pulse duration low, PCLK	50% reference points	2.286		ns
$t_{\text{p\_wl}}$	Pulse duration high, PCLK	50% reference points	2.286		ns
$t_{\text{p\_su}}$	Setup time – HSYNC, DATEN, PDATA(23:0) valid before the active edge of PCLK	50% reference points	0.8		ns
$t_{\text{p\_h}}$	Hold time – HSYNC, DATEN, PDATA(23:0) valid after the active edge of PCLK	50% reference points	0.8		ns
$t_{\text{t\_clk}}$	Transition time – PCLK	20% to 80% reference points		6	ns
$t_{\text{t}}$	Transition time – all other signals on this port	20% to 80% reference points		6	ns
$f_{\text{spread}}$	Supported Spread Spectrum range	Percent of $f_{\text{clock}}$ rate	-1%	+1% <sup>(1)</sup>	
$f_{\text{mod}}$	Supported Spread Spectrum Modulation Frequency <sup>(1) (2)</sup>		25	65 <sup>(3)</sup>	kHz
$t_{\text{p\_clkjit}}$	Clock jitter, PCLK		$t_{\text{p\_clkper}} - 5.414$		ps

(1) This value is limited by the maximum clock frequency for  $f_{\text{clock}}$  (that is, if  $f_{\text{clock}} = \text{max clock freq}$ , then  $f_{\text{spread}} \text{ max} = 0\%$ ).

(2) Modulation Waveforms supported: Sine and Triangle.

(3) Spread spectrum modulation tested at a maximum of 35 kHz. Simulated up to 65 kHz.



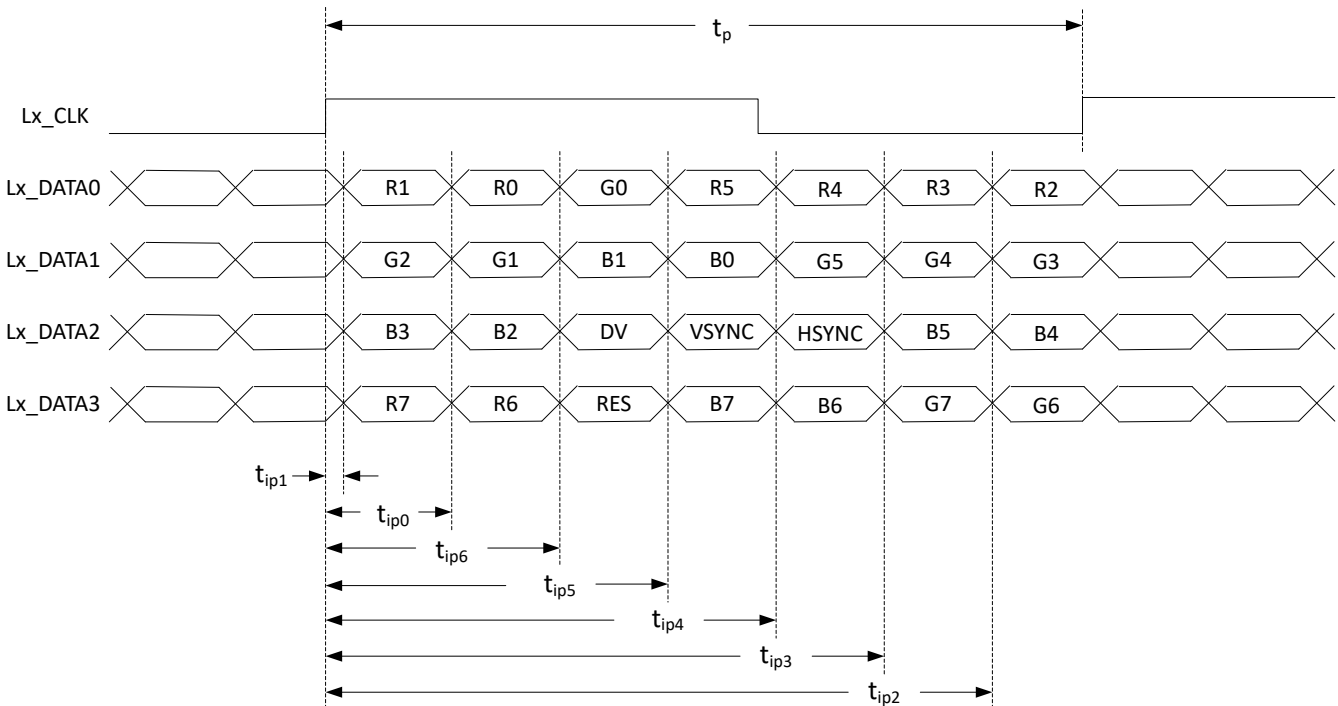
**Figure 6-3. Parallel Interface General Timing**

### 6.14 OpenLDI Interface General Timing Requirements

The DLPC230S-Q1 ASIC input interface supports a subset of the industry standard OpenLDI (FPD-Link I) interface (Open LVDS Display Interface Specification v0.95 - May 13, 1999). Specifically, from the standard, the ASIC supports the 24-bit, Single Pixel Format, using the Unbalanced Operating Mode and Pixel Mapping.

		MIN	NOM	MAX	UNIT
$f_{clock}$	Clock frequency, L1_CLK_P/N, L2_CLK_P/N	20.0		110	MHz
$t_p$	Clock period, PCLK	9.091		50	ns
$t_{skew}$	Skew Margin (between clock and data )	$f_{clock} = 85 \text{ MHz}$ -400 <sup>(5)</sup>	0	400 <sup>(5)</sup>	ps
$t_{skew\_ports}$	Clock to clock skew margin between ports on same ASIC, and between ports on different ASICs			1	clocks
$t_{ip0}$	Input data position 1	$(t_p / 7) - t_{skew}$	$(t_p / 7)$	$(t_p / 7) + t_{skew}$	ps
$t_{ip6}$	Input data position 2	$2 * (t_p / 7) - t_{skew}$	$2 * (t_p / 7)$	$2 * (t_p / 7) + t_{skew}$	ps
$t_{ip5}$	Input data position 3	$3 * (t_p / 7) - t_{skew}$	$3 * (t_p / 7)$	$3 * (t_p / 7) + t_{skew}$	ps
$t_{ip4}$	Input data position 4	$4 * (t_p / 7) - t_{skew}$	$4 * (t_p / 7)$	$4 * (t_p / 7) + t_{skew}$	ps
$t_{ip3}$	Input data position 5	$5 * (t_p / 7) - t_{skew}$	$5 * (t_p / 7)$	$5 * (t_p / 7) + t_{skew}$	ps
$t_{ip2}$	Input data position 6	$6 * (t_p / 7) - t_{skew}$	$6 * (t_p / 7)$	$6 * (t_p / 7) + t_{skew}$	ps
$t_{jitter}$	Input Jitter Tolerance (cycle to cycle, peak to peak)			100	ps
$f_{spread}$	Supported Spread Spectrum range	percent of $f_{clock}$ rate	-1% <sup>(1)</sup>	+1% <sup>(2)</sup>	
$f_{mod}$	Supported Spread Spectrum Modulation Frequency <sup>(3) (4)</sup>	25		65	kHz

- (1) This value is limited by the minimum clock frequency for  $f_{clock}$  (that is, if  $f_{clock} = \text{min clock freq}$ , then  $f_{spread} \text{ max} = 0\%$ ).
- (2) This value is limited by the maximum clock frequency for  $f_{clock}$  (that is, if  $f_{clock} = \text{max clock freq}$ , then  $f_{spread} \text{ max} = 0\%$ ).
- (3) Modulation Waveforms supported: Sine and Triangle.
- (4) Spread spectrum on OpenLDI interfaces was simulated, but not tested.
- (5)  $t_{skew}$  for other  $f_{clock}$  values can be estimated by  $\pm t_{skew} = -7.143 * f_{clock} + 1007.1 - (t_{jitter} - 100)$



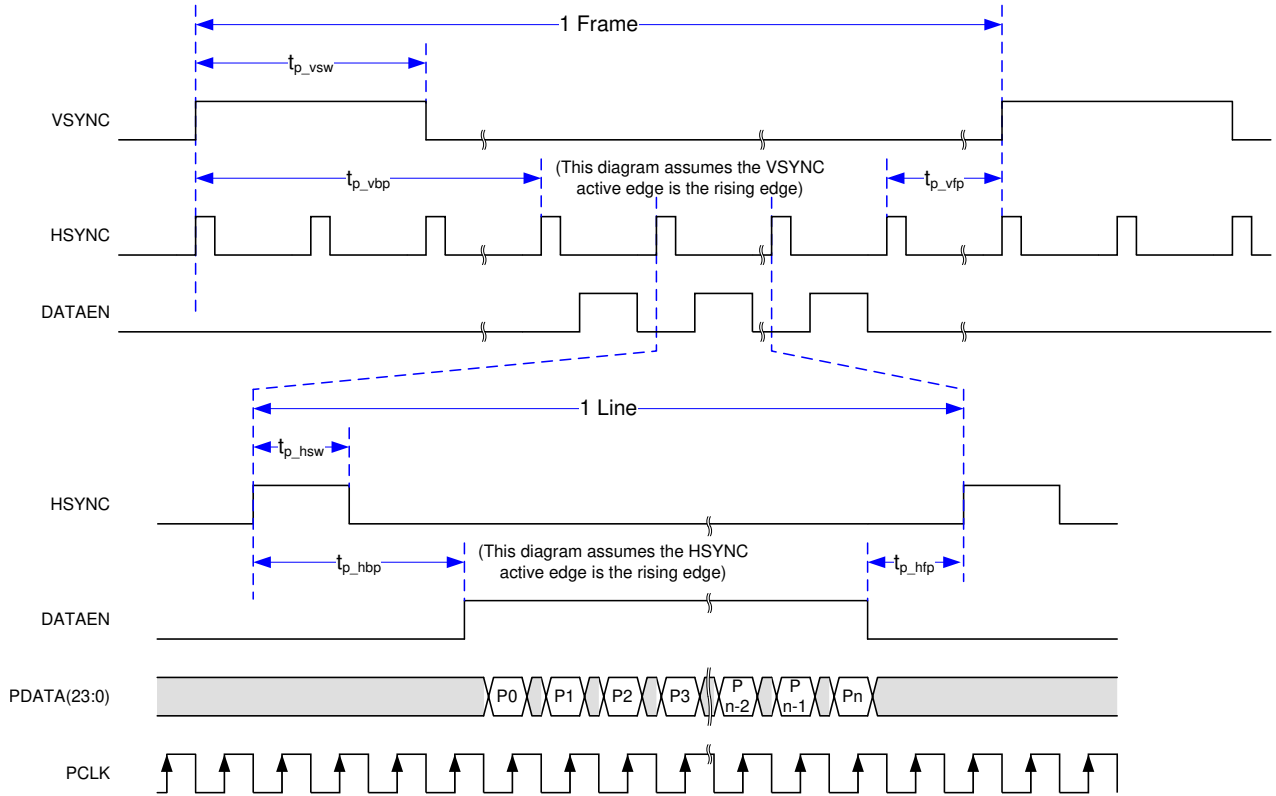
**Figure 6-4. OpenLDI Interface Timing**

## 6.15 Parallel/OpenLDI Interface Frame Timing Requirements

See<sup>(1)</sup>

			MIN	MAX	UNIT
VSYNC	Vertical Sync Rate (for the specified active source resolution)	1152 × 576 See <a href="#">Section 7.2.1</a>	58	61	Hz
VSYNC	Vertical Sync Rate (for the specified active source resolution)	1152 × 1152 See <a href="#">Section 7.2.1</a>	58	61	Hz
VSYNC	Vertical Sync Rate (for the specified active source resolution)	576 × 288 See <a href="#">Section 7.2.1</a>	58	61	Hz
$t_{p\_vsw}$	Pulse duration – VSYNC high	50% reference points	1		lines
$t_{p\_vbp}$	Vertical back porch (VBP) – time from the leading edge of VSYNC to the leading edge HSYNC for the first active line (includes $t_{p\_vsw}$ ).	50% reference points	2		lines
$t_{p\_vfp}$	Vertical front porch (VFP) – time from the leading edge of the HSYNC following the last active line in a frame to the leading edge of VSYNC	50% reference points	1		lines
$t_{p\_tvb}$	Total vertical blanking – time from the leading edge of HSYNC following the last active line of one frame to the leading edge of HSYNC for the first active line in the next frame. (This is equal to the sum of VBP ( $t_{p\_vbp}$ ) + VFP ( $t_{p\_vfp}$ ))	50% reference points	14		lines
$t_{p\_hsw}$	Pulse duration – HSYNC high	50% reference points	8		PCLKs
$t_{p\_hbp}$	Horizontal back porch – time from rising edge of HSYNC to rising edge of DATEN (includes $t_{p\_hsw}$ )	50% reference points	9		PCLKs
$t_{p\_hfp}$	Horizontal front porch – time from falling edge of DATEN to rising edge of HSYNC	50% reference points	8		PCLKs
$t_{p\_thb}$	Total horizontal blanking	50% reference points	64		PCLKs
TPPL	Total Pixels Per Line			8191	Pixels

- (1) While these requirements are not specific to the OpenLDI interface, they are appropriate for any source that drives an OpenLDI transmitter connected to the ASIC OpenLDI interface.



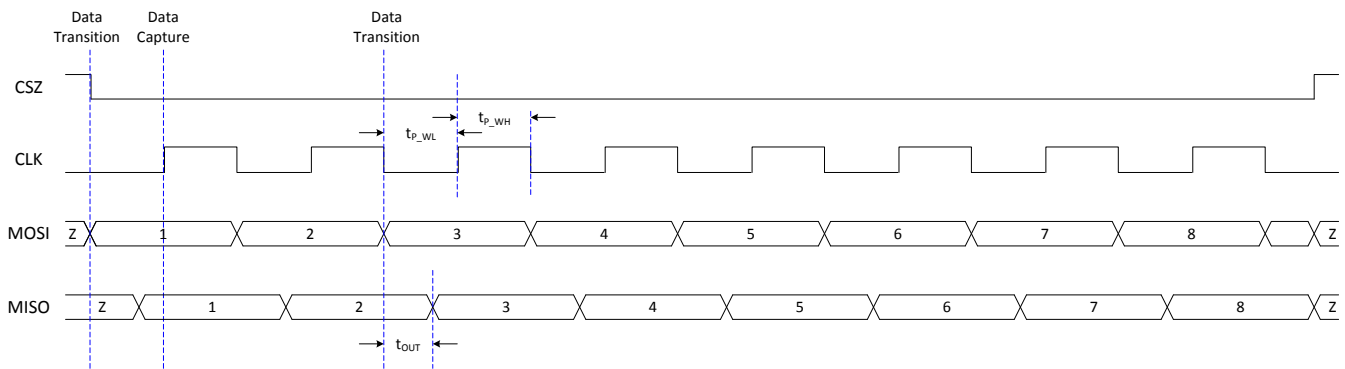
**Figure 6-5. Source Frame Timing**

## 6.16 Host/Diagnostic Port SPI Interface Timing Requirements

The DLPC230S-Q1 ASIC Host/Diagnostic SPI port interface timing requirements are shown below.<sup>(1)</sup>

		MIN	MAX	UNIT
$f_{\text{clock}}$	Clock frequency, HOST_SPI_CLK (50% reference points)		10.00	MHz
$t_{\text{p\_wh}}$	Pulse duration low, HOST_SPI_CLK (50% reference points)	45.0		ns
$t_{\text{p\_wl}}$	Pulse duration high, HOST_SPI_CLK (50% reference points)	45.0		ns
$t_t$	Transition time – all input signals	20% to 80% reference points	6	ns
$t_{\text{p\_su}}$	Setup time – HOST_SPI_DIN valid before HOST_SPI_CLK capture edge (50% reference points)	10.0		ns
$t_{\text{p\_h}}$	Hold time – HOST_SPI_DIN valid after HOST_SPI_CLK capture edge	50% reference points	18.0	ns
$t_{\text{out}}$	Clock-to-Data out - HOST_SPI_DOUT from HOST_SPI_CLK launch edge (50% reference points)	0.0	35.0	ns

- (1) The DLPC230S-Q1 Host/Diagnostic Port SPI interface supports SPI Modes 0, 1, 2, and 3 (that is, both clock polarities and both clock phases). The HOST\_SPI\_MODE input must be set to match the SPI mode being used.



**Figure 6-6. Host/Diagnostic Port SPI Interface Timing (Example: SPI Mode 0 (Clock Polarity = 0, Clock Phase = 0))**

## 6.17 Host/Diagnostic Port I<sup>2</sup>C Interface Timing Requirements

The DLPC230S-Q1 ASIC Host/Diagnostic I<sup>2</sup>C port interface timing requirements are shown below.<sup>(1) (2)</sup>

		MIN	MAX	UNIT
$f_{\text{clock}}$	Clock frequency, HOST_I <sup>2</sup> C_SCL (50% reference points)	Fast-Mode	400	kHz
		Standard Mode	100	
$C_L$	Capacitive Load (for each bus line)		200	pF

- (1) Meets all I<sup>2</sup>C timing per the I<sup>2</sup>C Bus Specification (except for capacitive loading as specified above). For reference see version 2.1 of the Phillips/NXP specification.
- (2) The maximum clock frequency does not account for rise time, nor added capacitance of PCB or external components which may adversely impact this value.

## 6.18 Flash Interface Timing Requirements

The DLPC230S-Q1 ASIC flash memory interface consists of a SPI serial interface. See [Section 8.3.4](#).

(1)			MIN	MAX	UNIT
$f_{\text{clock}}$	Clock frequency, FLSH_SPI_CLK	When VCC3IO_FLSH = 3.3 VDC	9.998	50.01 <sup>(2)</sup>	MHz
$t_{\text{p\_clkper}}$	Clock period, FLSH_SPI_CLK (50% reference points)	When VCC3IO_FLSH = 3.3 VDC	20.0	100	ns
$t_{\text{p\_wh}}$	Pulse duration low, FLSH_SPI_CLK (50% reference points)	When VCC3IO_FLSH = 3.3 VDC	9		ns
$t_{\text{p\_wl}}$	Pulse duration high, FLSH_SPI_CLK (50% reference points)	When VCC3IO_FLSH = 3.3 VDC	9		ns
$t_{\text{t}}$	Transition time – all input signals	20% to 80% reference points		6	ns
$t_{\text{p\_su}}$	Setup time – FLSH_SPI_DIO[3:0] valid before FLSH_SPI_CLK falling edge (50% reference points)	When VCC3IO_FLSH = 3.3 VDC	7.0		ns
$t_{\text{p\_h}}$	Hold time – FLSH_SPI_DIO[3:0] valid after FLSH_SPI_CLK falling edge	50% reference points	0.0		ns
$t_{\text{p\_clqv}}$	FLSH_SPI_DIO[3:0] output delay valid time  (with respect to falling edge of FLSH_SPI_CLK or falling edge of FLSH_SPI_CSZ) (50% reference points)	When VCC3IO_FLSH = 3.3 VDC	-3.0	3.0	ns

- (1) The DLPC230S-Q1 communicates with flash devices using a slight variant of SPI Transfer Mode 0 (that is, clock polarity = 0, clock phase = 0). Instead of capturing MISO data on the clock edge opposite from that used to transmit MOSI data, the DLPC230S-Q1 captures MISO data on the same clock edge used to transmit the next MOSI data. As such, the DLPC230S-Q1 Flash SPI interface requires that MISO data from the flash device remain active until the end of the full clock cycle to allow the last data bit to be captured. This is shown in [Figure 6-8](#).
- (2) The actual maximum clock rate driven from the DLPC230S-Q1 may be slightly less than this value.

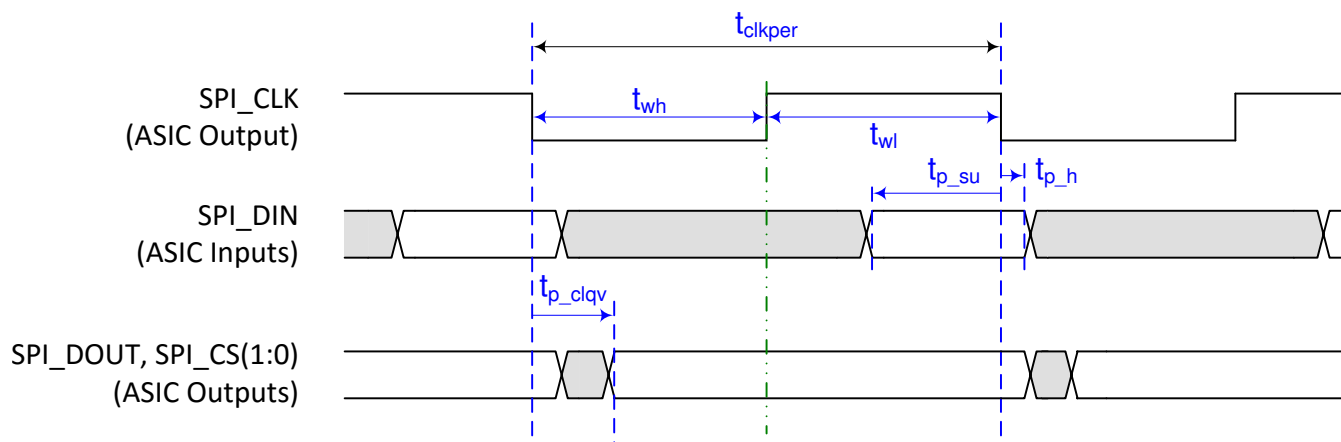
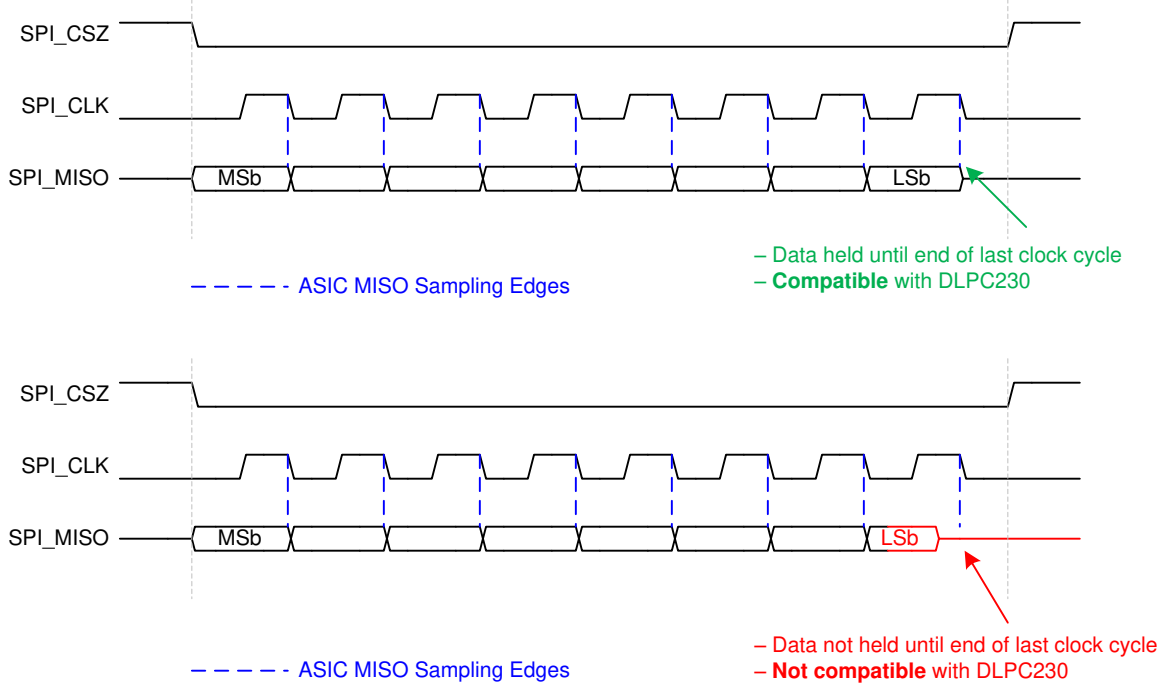


Figure 6-7. Flash Interface Timing



**Figure 6-8. Flash Interface Data Capture Requirements**

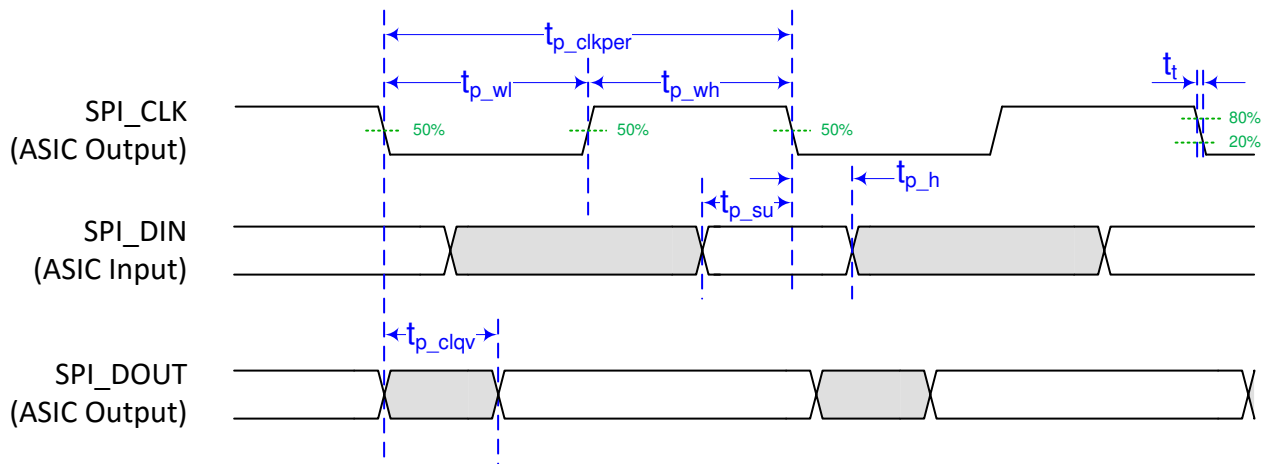


## 6.19 TPS99000S-Q1 SPI Interface Timing Requirements

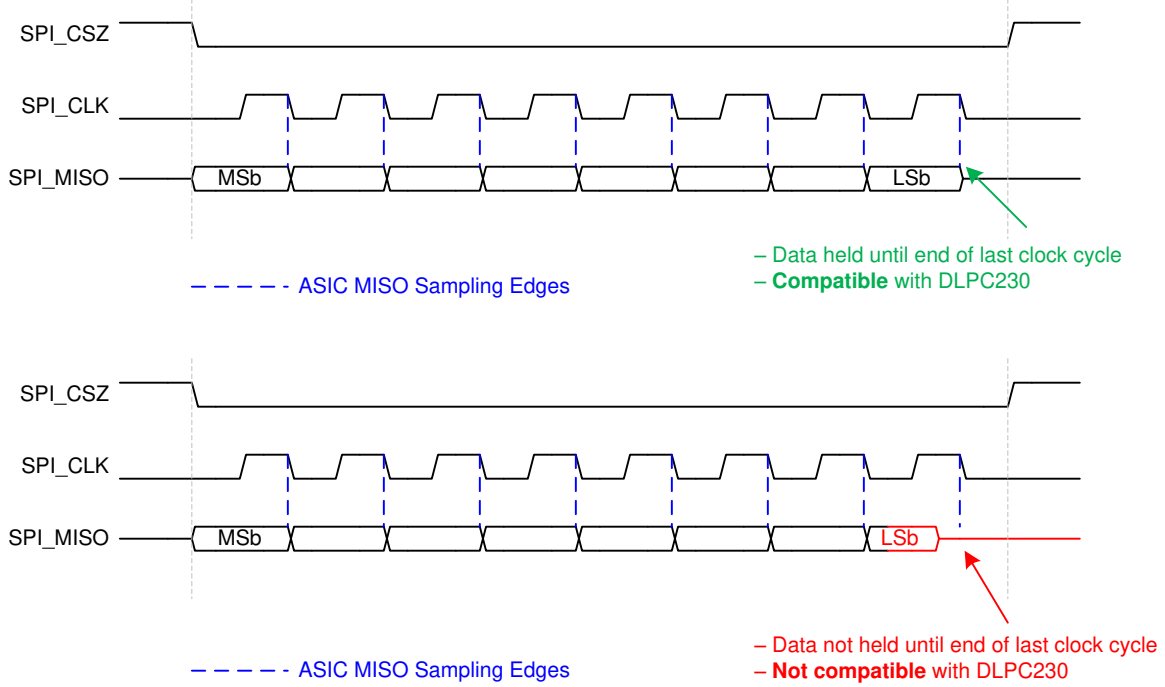
The DLPC230S-Q1 ASIC to TPS99000S-Q1 interface consists of a SPI serial interface.

(1)		MIN	MAX	UNIT
$f_{\text{clock}}$	Clock frequency, PMIC_SPI_CLK	9.998	30.006	MHz
$t_{\text{p\_clkper}}$	Clock period, PMIC_SPI_CLK (50% reference points)	33.3	100	ns
$t_{\text{p\_wh}}$	Pulse duration high, PMIC_SPI_CLK (50% reference points)	11.5		ns
$t_{\text{p\_wl}}$	Pulse duration low, PMIC_SPI_CLK (50% reference points)	11.5		ns
$t_t$	Transition time – all input signals	20% to 80% reference points		6
$t_{\text{p\_su}}$	Setup time – PMIC_SPI_DIN valid before PMIC_SPI_CLK falling edge (50% reference points)	7.0		ns
$t_{\text{p\_h}}$	Hold time – PMIC_SPI_DIN valid after PMIC_SPI_CLK falling edge	50% reference points		0.0
$t_{\text{p\_clqv}}$	PMIC_SPI_DOUT output delay (valid) time  (with respect to falling edge of PMIC_SPI_CLK or falling edge of PMIC_SPI_CSZ0) (50% reference points)	-3.0	3.0	ns

- (1) The DLPC230S-Q1 communicates with the TPS99000S-Q1 using a slight variant of SPI Transfer Mode 0 (that is, clock polarity = 0, clock phase = 0). Instead of capturing MISO data on the clock edge opposite from that used to transmit MOSI data, the DLPC230S-Q1 captures MISO data on the same clock edge used to transmit the next MOSI data. As such, the DLPC230S-Q1 SPI interface to the TPS99000S-Q1 requires that MISO data from the TPS99000S-Q1 remain active until the end of the full clock cycle to allow the last data bit to be captured. This is shown in Figure 6-12.



**Figure 6-9. TPS99000S-Q1 Interface Timing**



**Figure 6-10. TPS99000S-Q1 Interface Data Capture Requirements**

## 6.20 TPS99000S-Q1 AD3 Interface Timing Requirements

The DLPC230S-Q1 ASIC to TPS99000S-Q1 AD3 interface is used to retrieve ADC measurements from the TPS99000S-Q1. The interface is similar to SPI and includes a clock, MOSI, and MISO signal.

(1) (2) (3)		MIN	MAX	UNIT
$f_{\text{clock}}$	Clock frequency, PMIC_AD3_CLK	29.326	30.006	MHz
$t_{\text{p\_clkper}}$	Clock period, PMIC_AD3_CLK (50% reference points)	33.327	34.100	ns
$t_{\text{p\_wh}}$	Pulse duration high, PMIC_AD3_CLK (50% reference points) (Referenced to $t_{\text{p\_clkper}}$ )	40%		
$t_{\text{p\_wl}}$	Pulse duration low, PMIC_AD3_CLK (50% reference points) (Referenced to $t_{\text{p\_clkper}}$ )	40%		
$t_t$	Transition time – all input signals	20% to 80% reference points		6 ns
$t_{\text{p\_su}}$	Setup time – PMIC_AD3_MISO valid before PMIC_AD3_CLK rising edge (50% reference points)	14.5		ns
$t_{\text{p\_h}}$	Hold time – PMIC_AD3_MISO valid after PMIC_AD3_CLK rising edge (50% reference points)	0		ns
$t_{\text{p\_clqv}}$	PMIC_AD3_MOSI output delay (valid) time (with respect to falling edge of PMIC_SPI_CLK) (50% reference points)	-2.0	2.0	ns

- (1) PMIC\_AD3\_MOSI (Master (DLPC230S-Q1) Output / Slave (TPS99000S-Q1) Input) is transmitted on the falling edge of PMIC\_AD3\_CLK.
- (2) PMIC\_AD3\_MISO (Master (DLPC230S-Q1) Input / Slave (TPS99000S-Q1) Output) is captured on the rising edge of PMIC\_AD3\_CLK.
- (3) PMIC\_AD3\_CLK is used as the primary TPS99000S-Q1 system clock in addition to supporting the AD3 interface.

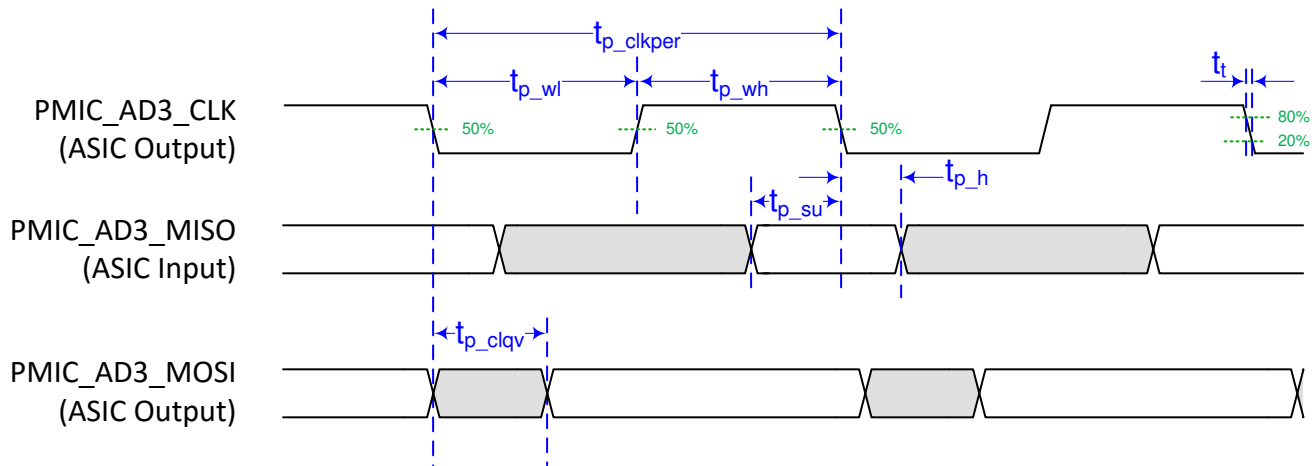
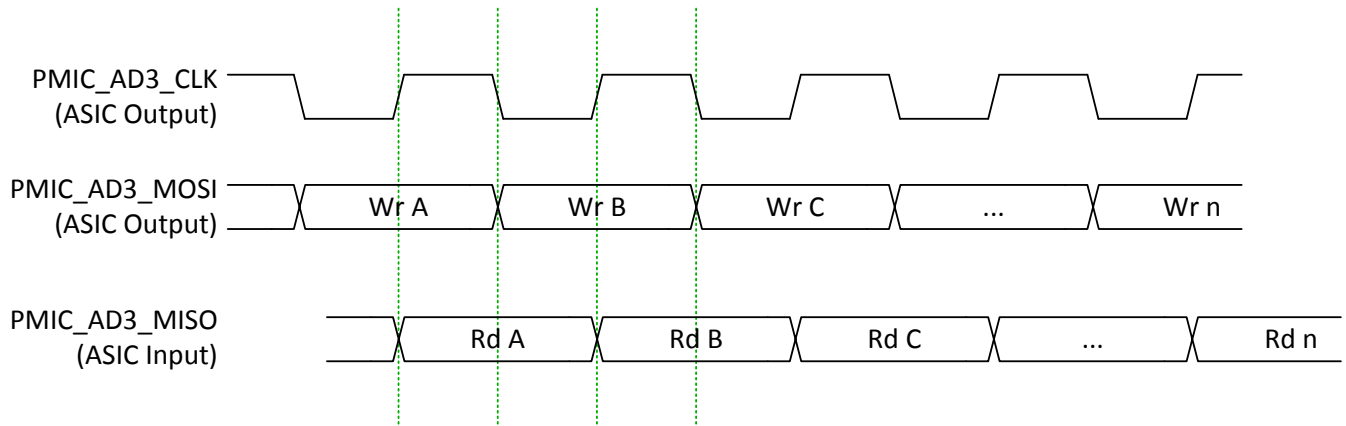


Figure 6-11. TPS99000S-Q1 AD3 Interface Timing



**Figure 6-12. TPS99000S-Q1 AD3 Data Capture and Transition**

## 6.21 Master I<sup>2</sup>C Port Interface Timing Requirements

The DLPC230S-Q1 ASIC Master I<sup>2</sup>C port interface timing requirements are shown below.

(1) (2)		MIN	MAX	UNIT
f <sub>clock</sub>	Clock frequency, MSTR_SCL (50% reference points)	Fast-Mode	400	kHz
		Standard Mode	100	
C <sub>L</sub>	Capacitive Load (for each bus line)		200	pF

- (1) Meets all I<sup>2</sup>C timing per the I<sup>2</sup>C Bus Specification (except for Capacitive Loading as specified above).
- (2) The maximum clock frequency does not account for rise time, nor added capacitance of PCB or external components which may adversely impact this value.

## 6.22 Chipset Component Usage Specification

TI DLP® chipsets include a DMD and one or more controllers. Reliable function and operation of TI DMDs requires that they be used in conjunction with all of the other components in the applicable chipset, including those components that contain or implement TI DMD control technology, such as the DLPC230S-Q1. TI DMD control technology is the TI technology and devices for operating or controlling a DLP® products DMD.

## 7 Parameter Measurement Information

### 7.1 HOST\_IRQ Usage Model

In the DLPC230S-Q1, the Host\_IRQ signal is used to serve as an indication that a serious system error has occurred for which the ASIC has executed an emergency shutdown. The specific error(s) that precipitated the shutdown can be retrieved via the Host Command and Control interface. The actions that are taken by the ASIC for an emergency shutdown are:

- LEDs are disabled
- The DMD is parked and powered-down
- The ASIC operational mode is transitioned to Standby
- The precipitating errors are captured for later review
- The Host\_IRQ signal is set to a high state

To recover from an emergency shutdown, the system will require a full power cycle (De-assertion of PROJ\_ON). The host should be sure to obtain the error history from the ASIC prior to this full reset, as the reset will remove all error history from the system.

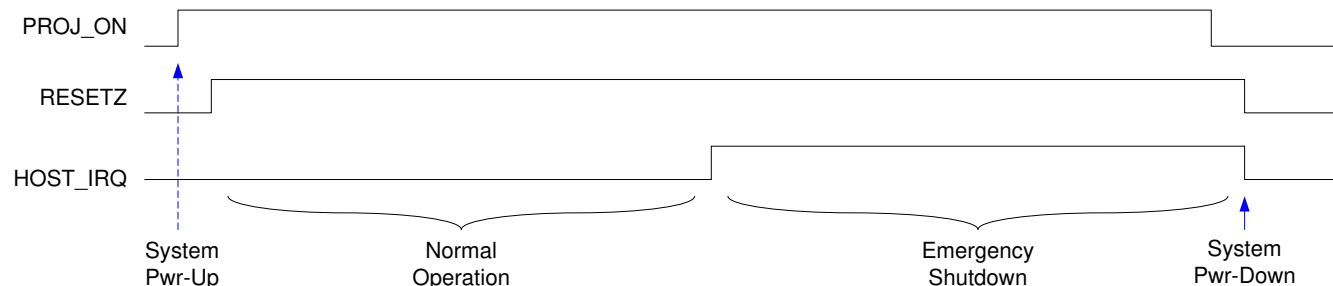


Figure 7-1. Host IRQ Timing

### 7.2 Input Source

The video input source can be configured to accommodate various desired input resolutions. Image processing such as scaling and line replication may be applied in order to achieve the necessary display resolution. The desired input resolution may depend on product configuration.

#### 7.2.1 Supported Input Sources

The supported sources with typical timings are shown in [Table 7-1](#). These typical timing examples do not minimize blanking or pixel clock rate. Refer to [Section 6.15](#) for minimum timing specifications.

Table 7-1. Typical Timing for Supported Source Resolutions

HORIZONTAL RESOLUTION	VERTICAL RESOLUTION	HORIZONTAL BLANKING				VERTICAL BLANKING				VERTICAL RATE (Hz)	PIXEL CLOCK (MHz)
		TOTAL <sup>(1)</sup>	SYNC (PIXEL CLOCKS)	BACK PORCH (PIXEL CLOCKS)	FRONT PORCH (PIXEL CLOCKS)	TOTAL <sup>(1)</sup>	SYNC (LINES)	BACK PORCH (LINES)	FRONT PORCH (LINES)		
576	288	322	8	154	160	181	8	83	90	60	25.270
1152	576	80	8	32	40	25	8	14	3	60	44.426
1152	1152	80	8	32	40	33	8	6	19	60	87.595

- (1) Sync clocks/lines are counted as a part of total blanking in these examples (Total Blanking = sync + back porch + front porch). Note that the specifications in [Section 6.15](#) include sync width as part of back porch (Total Blanking = back porch + front porch).

## 7.2.2 Parallel Interface Supported Data Transfer Formats

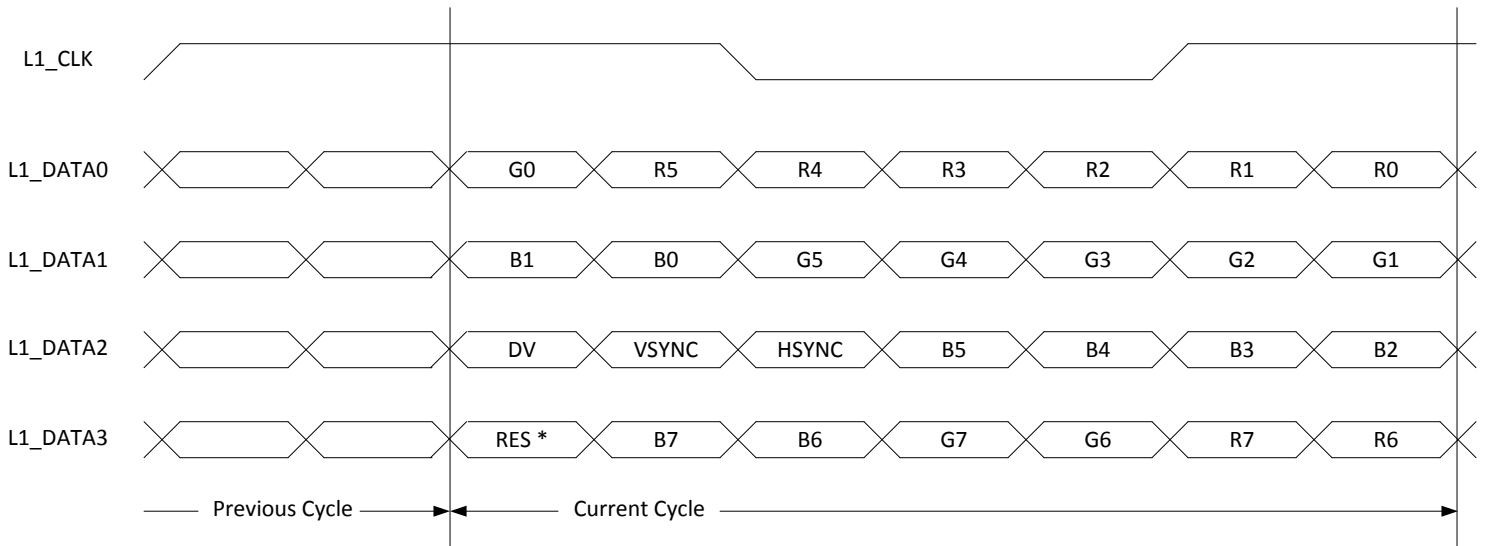
- 24-bit RGB888 on a 24 data wire interface

### 7.2.2.1 OpenLDI Interface Supported Data Transfer Formats

- 1X 24-bit RGB888 on a 5-lane differential interface

Section 7.2.2.1.1 shows the required OpenLDI bus mapping for the supported data transfer formats.

#### 7.2.2.1.1 OpenLDI Interface Bit Mapping Modes



A. \* = Use is undefined/reserved

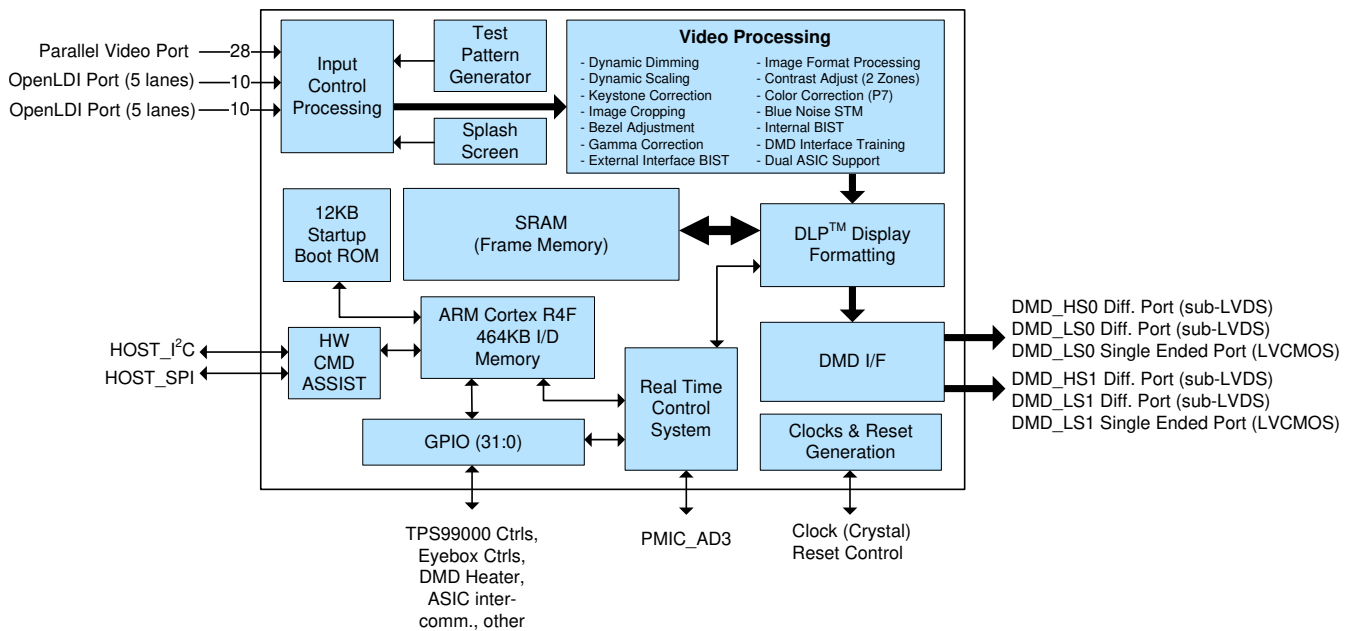
**Figure 7-2. OpenLDI 24-bit Single Port**

## 8 Detailed Description

### 8.1 Overview

The automotive DLP® Products chipset consists of three components – the DLP5530S-Q1, the DLPC230S-Q1, and the TPS9900S-Q1. The DLPC230S-Q1 is the display controller for the DMD - it formats incoming video and controls the timing of the DMD. It also controls TPS9900S-Q1 light source signal timing to coordinate with DMD timing in order to synchronize light output with DMD mirror movement. The DLPC230S-Q1 is designed for automotive applications with a wide operating temperature range and diagnostic features to identify and correct specific system-level failures. The DLPC230S-Q1 provides interfaces such as OpenLDI (video) and sub-LVDS (DMD interface) to minimize power consumption and EMI. Applications include head-up display (HUD) and adaptive high beam and smart headlight.

### 8.2 Functional Block Diagram



**Figure 8-1. Functional Block Diagram**



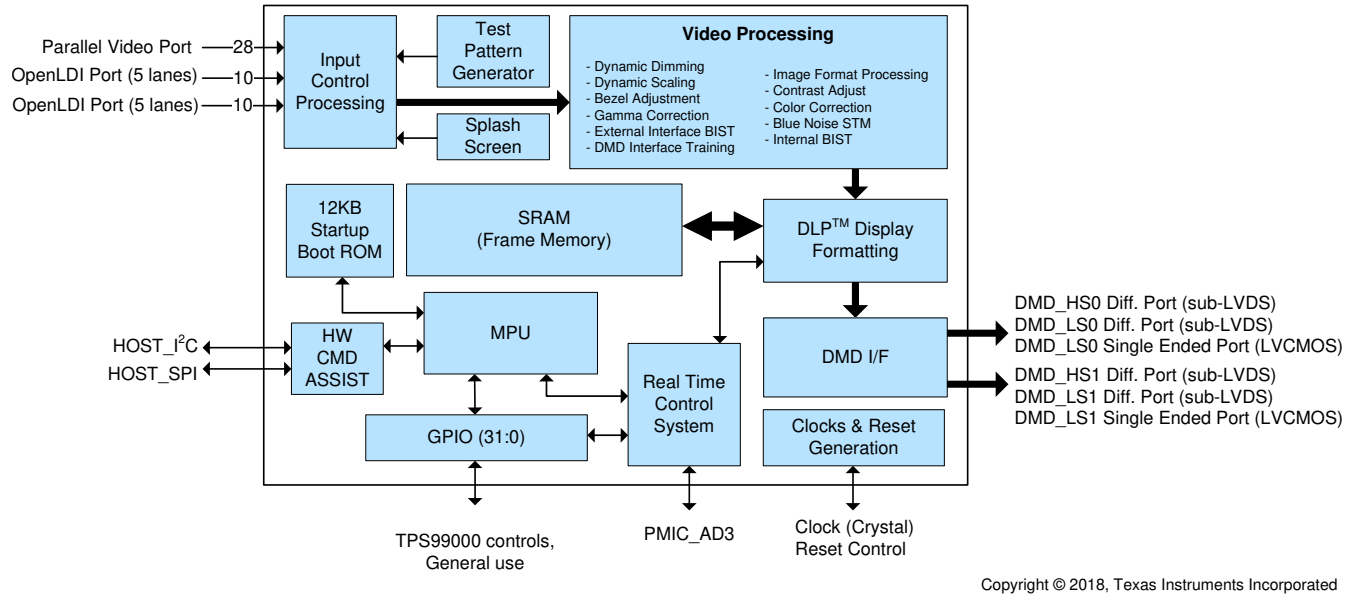


Figure 8-2. Alternate Functional Block Diagram

## 8.3 Feature Description

### 8.3.1 Parallel Interface

The parallel interface complies with standard graphics interface protocol, which includes a vertical sync signal (VSYNC), horizontal sync signal (HSYNC), data valid signal (DATEN), a 24-bit data bus (PDATA\_x), and a pixel clock (PCLK). Figure 6-5 shows the relationship of these signals.

#### Note

VSYNC must remain active at all times. If VSYNC is lost, the DMD must be transitioned to a safe state. When the system detects a VSYNC loss, it will switch to a test pattern or splash image as specified in flash by the Host.

The parallel interface supports intra-interface bit multiplexing (specified in flash) that can help with board layout as needed. The intra-interface bit multiplexing allows the mapping of any PDATA\_x input to any internal data bus bit. When utilizing this feature, each unique input pin can only be mapped to one unique destination bit. The typical mapping is shown in Figure 8-3. An example of an alternate mapping is shown in Figure 8-4.

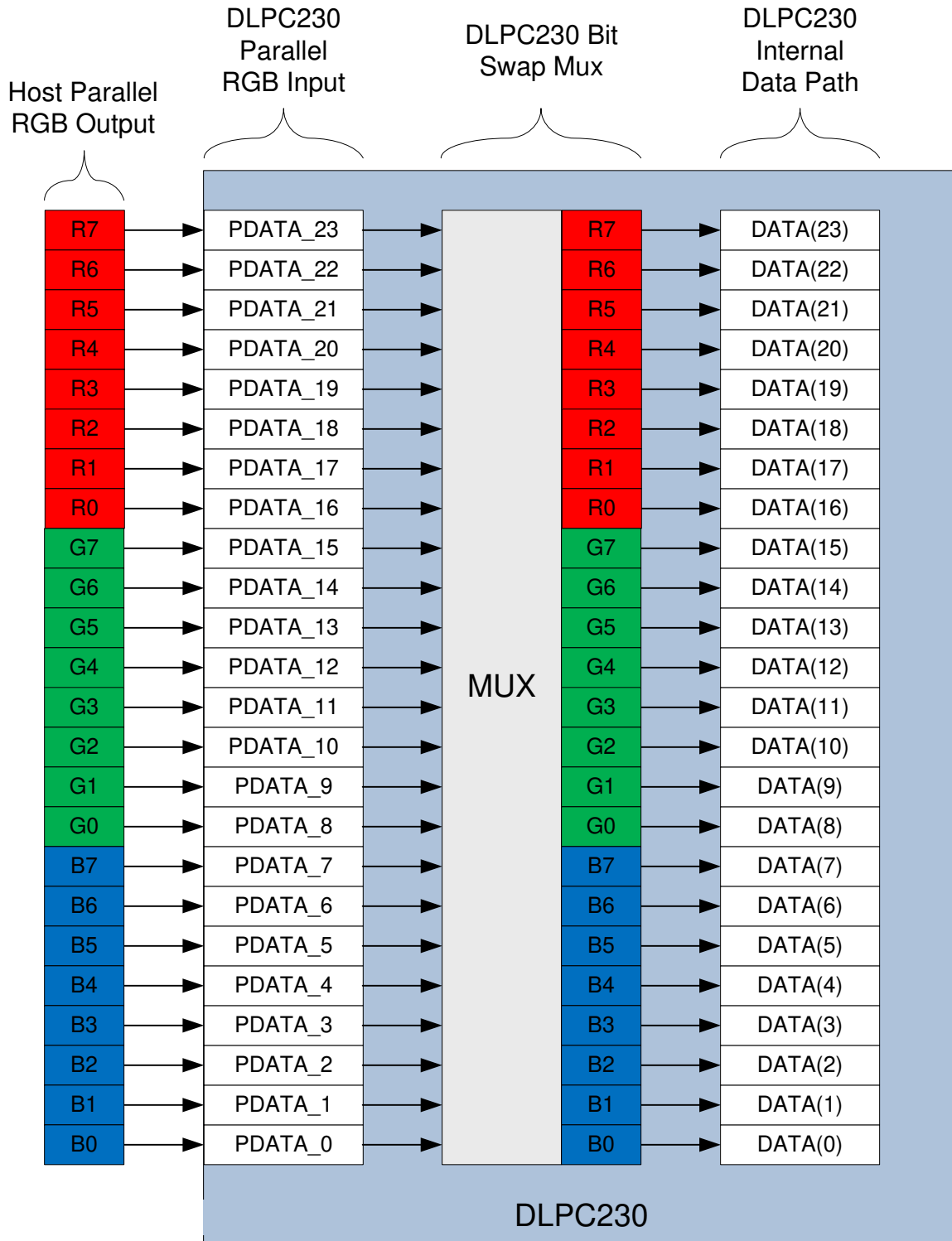


Figure 8-3. Example of Typical Parallel Port Bit Mapping

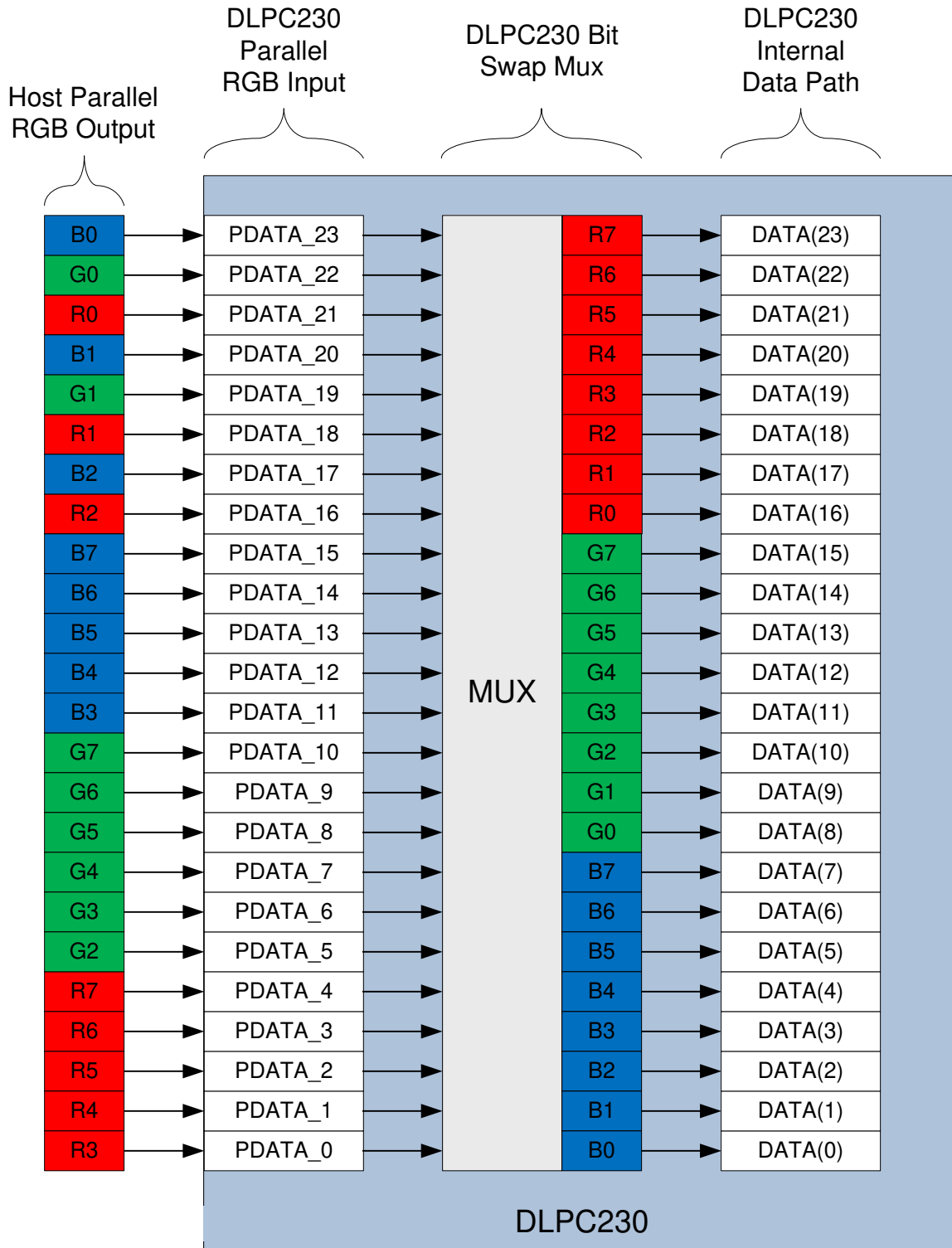


Figure 8-4. Example of Alternate Parallel Port Bit Mapping

### 8.3.2 OpenLDI Interface

Each DLPC230S-Q1 OpenLDI interface port supports intra-port lane multiplexing (specified in flash) that can help with board layout as needed. The intra-port multiplexing allows the mapping of any Lx\_DATA lane pair to any internal data lane pair. When utilizing this feature, each unique lane pair can only be mapped to one unique

destination lane pair. The typical lane mapping is shown in Figure 8-5. An example of an alternate lane mapping is shown in Figure 8-6.

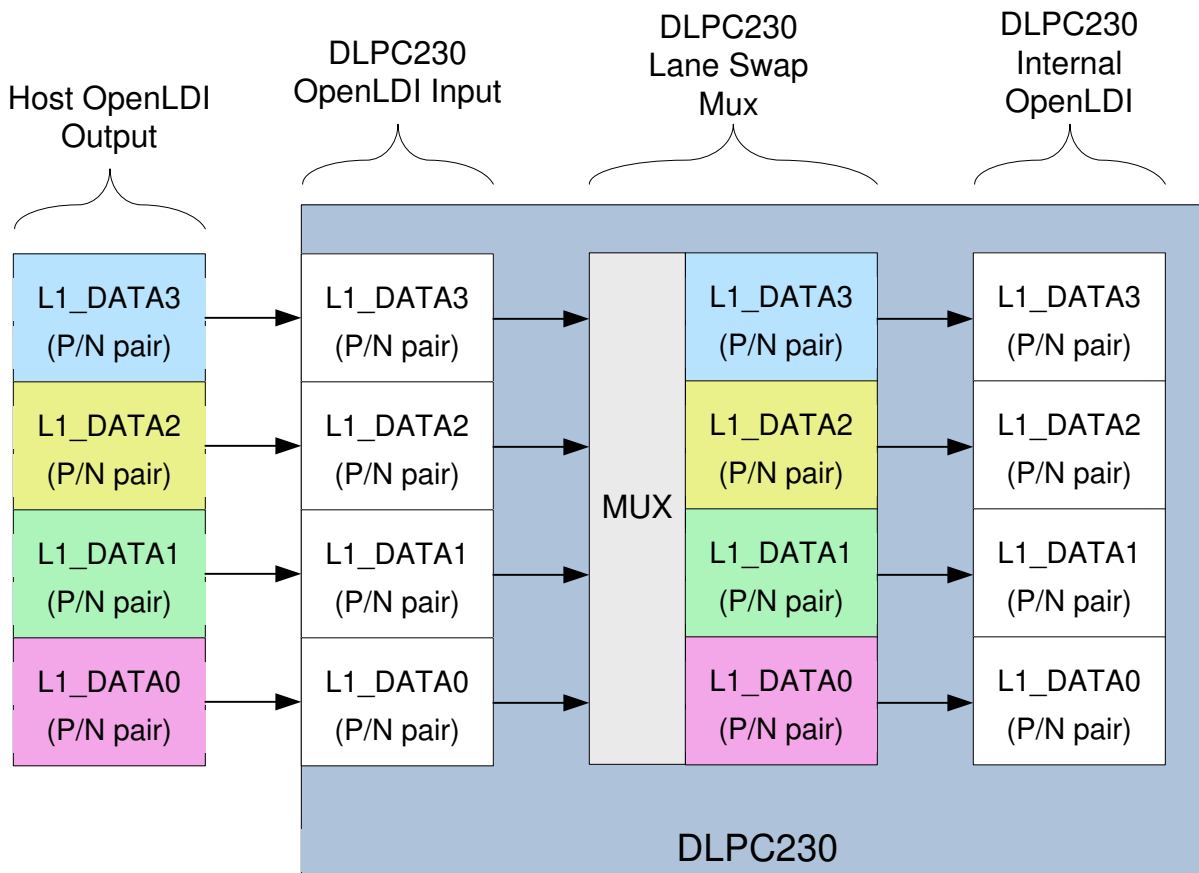
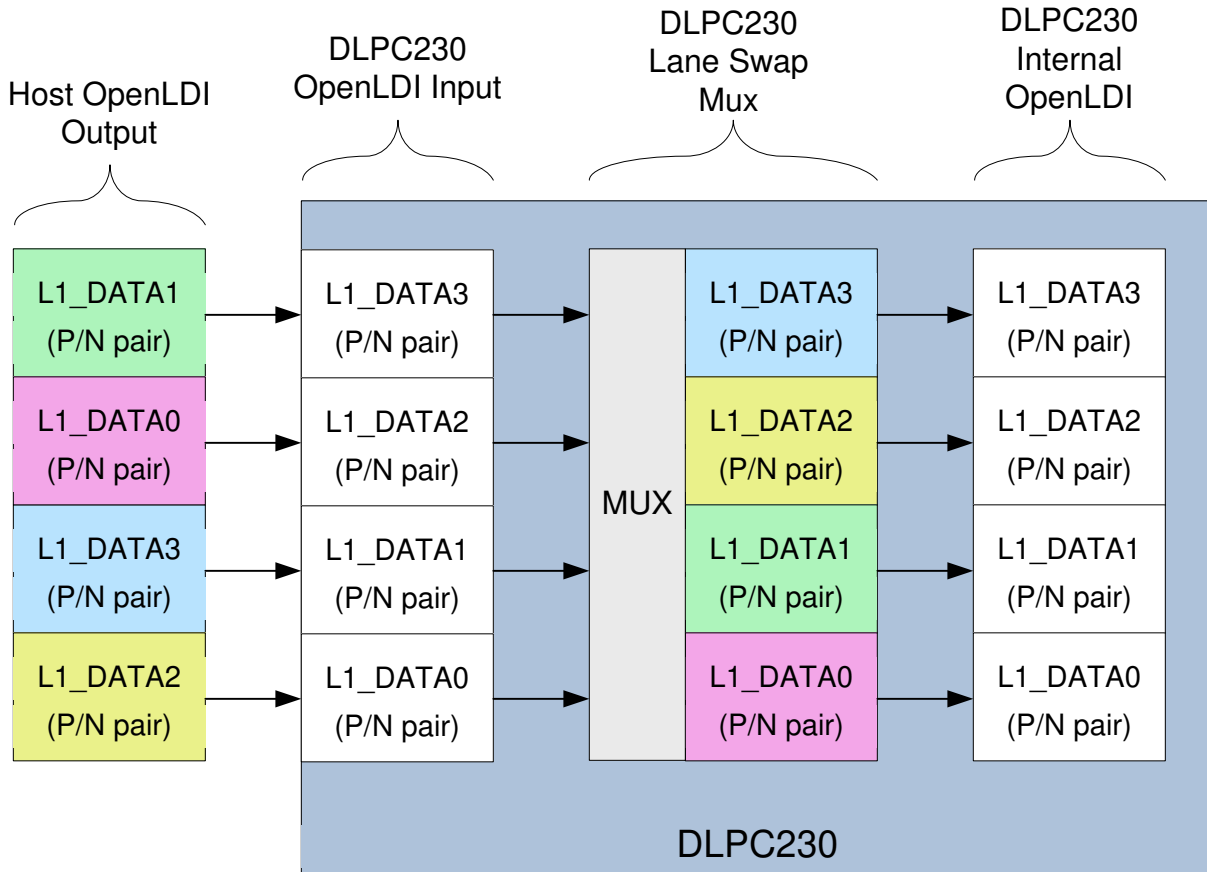


Figure 8-5. Example of Typical OpenLDI Port Lane Mapping



**Figure 8-6. Example of Alternate OpenLDI Port Lane Mapping**

### 8.3.3 DMD (Sub-LVDS) Interface

The DLPC230S-Q1 ASIC DMD interface supports two high-speed sub-LVDS output-only interfaces for data transmission, a single low-speed sub-LVDS output-only interface for command write transactions, as well as a low-speed single-ended input interface used for command read transactions. The DLPC230S-Q1 supports a limited number of DMD interface swap configurations (specified in Flash) that can help board layout by remapping specific combinations of DMD interface lines to other DMD interface lines as needed. [Table 8-1](#) shows some of the options available.

**Table 8-1. ASIC to 8-Lane DMD Pin Mapping Options**

DLPC230S-Q1 ASIC PIN ROUTING OPTIONS TO DMD PINS				
BASELINE	FULL FLIP HS0/HS1 180	SWAP HS0 PORT WITH HS1 PORT	SWAP HS0 PORT WITH HS1 PORT AND FULL FLIP 180	DMD PINS
HS0_WDATA0_P HS0_WDATA0_N	HS0_WDATA7_P HS0_WDATA7_N	HS1_WDATA0_P HS1_WDATA0_N	HS1_WDATA7_P HS1_WDATA7_N	D_AP(0) D_AN(0)
HS0_WDATA1_P HS0_WDATA1_N	HS0_WDATA6_P HS0_WDATA6_N	HS1_WDATA1_P HS1_WDATA1_N	HS1_WDATA6_P HS1_WDATA6_N	D_AP(1) D_AN(1)
HS0_WDATA2_P HS0_WDATA2_N	HS0_WDATA5_P HS0_WDATA5_N	HS1_WDATA2_P HS1_WDATA2_N	HS1_WDATA5_P HS1_WDATA5_N	D_AP(2) D_AN(2)
HS0_WDATA3_P HS0_WDATA3_N	HS0_WDATA4_P HS0_WDATA4_N	HS1_WDATA3_P HS1_WDATA3_N	HS1_WDATA4_P HS1_WDATA4_N	D_AP(3) D_AN(3)
HS0_WDATA4_P HS0_WDATA4_N	HS0_WDATA3_P HS0_WDATA3_N	HS1_WDATA4_P HS1_WDATA4_N	HS1_WDATA3_P HS1_WDATA3_N	D_AP(4) D_AN(4)
HS0_WDATA5_P HS0_WDATA5_N	HS0_WDATA2_P HS0_WDATA2_N	HS1_WDATA5_P HS1_WDATA5_N	HS1_WDATA2_P HS1_WDATA2_N	D_AP(5) D_AN(5)
HS0_WDATA6_P HS0_WDATA6_N	HS0_WDATA1_P HS0_WDATA1_N	HS1_WDATA6_P HS1_WDATA6_N	HS1_WDATA1_P HS1_WDATA1_N	D_AP(6) D_AN(6)
HS0_WDATA7_P HS0_WDATA7_N	HS0_WDATA0_P HS0_WDATA0_N	HS1_WDATA7_P HS1_WDATA7_N	HS1_WDATA0_P HS1_WDATA0_N	D_AP(7) D_AN(7)
HS1_WDATA0_P HS1_WDATA0_N	HS1_WDATA7_P HS1_WDATA7_N	HS0_WDATA0_P HS0_WDATA0_N	HS0_WDATA7_P HS0_WDATA7_N	D_BP(0) D_BN(0)
HS1_WDATA1_P HS1_WDATA1_N	HS1_WDATA6_P HS1_WDATA6_N	HS0_WDATA1_P HS0_WDATA1_N	HS0_WDATA6_P HS0_WDATA6_N	D_BP(1) D_BN(1)
HS1_WDATA2_P HS1_WDATA2_N	HS1_WDATA5_P HS1_WDATA5_N	HS0_WDATA2_P HS0_WDATA2_N	HS0_WDATA5_P HS0_WDATA5_N	D_BP(2) D_BN(2)
HS1_WDATA3_P HS1_WDATA3_N	HS1_WDATA4_P HS1_WDATA4_N	HS0_WDATA3_P HS0_WDATA3_N	HS0_WDATA4_P HS0_WDATA4_N	D_BP(3) D_BN(3)
HS1_WDATA4_P HS1_WDATA4_N	HS1_WDATA3_P HS1_WDATA3_N	HS0_WDATA4_P HS0_WDATA4_N	HS0_WDATA3_P HS0_WDATA3_N	D_BP(4) D_BN(4)
HS1_WDATA5_P HS1_WDATA5_N	HS1_WDATA2_P HS1_WDATA2_N	HS0_WDATA5_P HS0_WDATA5_N	HS0_WDATA2_P HS0_WDATA2_N	D_BP(5) D_BN(5)
HS1_WDATA6_P HS1_WDATA6_N	HS1_WDATA1_P HS1_WDATA1_N	HS0_WDATA6_P HS0_WDATA6_N	HS0_WDATA1_P HS0_WDATA1_N	D_BP(6) D_BN(6)
HS1_WDATA7_P HS1_WDATA7_N	HS1_WDATA0_P HS1_WDATA0_N	HS0_WDATA7_P HS0_WDATA7_N	HS0_WDATA0_P HS0_WDATA0_N	D_BP(7) D_BN(7)

### 8.3.4 Serial Flash Interface

The DLPC230S-Q1 uses an external SPI serial flash memory device for configuration and operational data. The minimum supported size is 16 Mb . Larger devices may be required based on operation data and splash image size. The maximum supported size is 128 Mb. It should be noted that the system will support 256 Mb and 512 Mb devices, however, only the first 128 Mb of space will be used.

The external serial flash device is supported on a single SPI interface and mostly complies with industry standard SPI flash protocol (See [Figure 6-8](#)). The Host will specify the maximum supported flash interface

frequency (which can be based on device limits, system limits, and/or other factors) and the system will program the closest obtainable value less than or equal to this specified maximum.

The DLPC230S-Q1 ASIC flash must be connected to the designated SPI flash interface (FLSH\_SPI\_XXX) to enable support for system initialization, configuration, and operation.

The DLPC230S-Q1 should support any flash device that is compatible with the modes of operation, features, and performance as defined in this section.

**Table 8-2. SPI Flash Required Features or Modes of Operation**

FEATURE	DLPC230S-Q1 REQUIREMENT	COMMENTS
SPI interface width	Single Wire, Two Wire, Four Wire	
SPI protocol	SPI mode 0	
Fast READ addressing	Auto-incrementing	
Programming mode	Page mode	
Page size	256 Bytes	
Sector (or sub-sector) size	4 KB	Required erase granularity
Block structure	Uniform sector / sub-sector	
Block protection bits	0 = Disabled (with Default = 0 = Disabled)	
Status register bit(0)	Write in progress (WIP) {also called flash busy}	
Status register bit(1)	Write enable latch (WEN)	
Status register bits(6:2)	A value of 0 disables programming protection	
Status register bit(7)	Status register write protect (SRWP)	
Status register bits(15:8) (expanded status register), or Secondary Status register	The DLPC230S-Q1 supports multi-byte status registers, as well as separate, additional status registers, but only for specific devices/register addresses. The supported registers and addresses are specified in <a href="#">Table 8-3</a> .	

**CAUTION**

The selected SPI flash device must block repeated status writes from being written to internal register. The boot application writes to the flash device status register once per 256 bytes during programming. Most flash devices discard status register writes when the status content does not change. Some flash parts, such as the Micron N25Q128A13ESFA0F, do not block status writes when the status data is repeated. This causes the status register to exceed its maximum write limit after several programming cycles, making them incompatible with the DLPC230S-Q1. Note that the main application does not write to the status register.

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For each write operation, the DLPC230S-Q1 boot application executes the following:

1. Write enable command
2. Write status command (to unprotect memory)
3. Read status command to poll the successful execution of the write status (repeated as needed)
4. Write enable command
5. Program or erase command
6. Read status command (repeated as needed) to poll the successful execution of the program or erase operation
7. Write disable command (during programming; this is not performed after erase command.)

For each write operation, the DLPC230S-Q1 main application executes the following:

1. Write enable command
2. Program or erase command
3. Read status command (repeated as needed) to poll the successful execution of the program or erase operation
4. Write disable command (during programming; this is not performed after erase command)

The specific instruction op-code and timing compatibility requirements are listed in [Table 8-3](#) and [Flash Interface Timing Requirements](#). Note that DLPC230S-Q1 does not read the flash's full electronic signature ID and thus cannot automatically adapt protocol and clock rates based on the ID.



**Table 8-3. SPI Flash Instruction Op-Code and Access Profile Compatibility Requirements**

SPI FLASH COMMAND	FIRST BYTE (OP-CODE)	SECOND BYTE	THIRD BYTE	FOURTH BYTE	FIFTH BYTE	SIXTH BYTE	NO. OF DUMMY CLOCKS	COMMENTS
Fast READ (1/1)	0x0B	ADDRS(0)	ADDRS(1)	ADDRS(2)	dummy	DATA(0) <sup>(1)</sup>	8	See <a href="#">Table 8-4</a>
Dual READ (1/2)	0x3B	ADDRS(0)	ADDRS(1)	ADDRS(2)	dummy	DATA(0) <sup>(1)</sup>	8	See <a href="#">Table 8-4</a>
2X READ (2/2)	0xBB	ADDRS(0)	ADDRS(1)	ADDRS(2)	dummy	DATA(0) <sup>(1)</sup>	4	See <a href="#">Table 8-4</a>
Quad READ (1/4)	0x6B	ADDRS(0)	ADDRS(1)	ADDRS(2)	dummy	DATA(0) <sup>(1)</sup>	8	See <a href="#">Table 8-4</a>
4X READ (4/4)	0xEB	ADDRS(0)	ADDRS(1)	ADDRS(2)	dummy	DATA(0) <sup>(1)</sup>	6	See <a href="#">Table 8-4</a>
Read status	0x05	n/a	n/a	STATUS(0)	STATUS(1)		0	Status(1) - Winbond only
Write status	0x01	STATUS(0)	STATUS(1)				0	Status(1) - Winbond only
Read Volatile Conf Reg	0x85	Data(0)					0	Micron Only
Write Volatile Conf Reg	0x81	Data(0)					0	Micron Only
Write Enable	0x06						0	
Write Disable	0x04						0	
Page program	0x02	ADDRS(0)	ADDRS(1)	ADDRS(2)	DATA(0) <sup>(1)</sup>		0	
Sector/Sub-sector Erase (4KB)	0x20	ADDRS(0)	ADDRS(1)	ADDRS(2)			0	
Full Chip Erase	0xC7						0	
Software Reset Enable	0x66							
Software Reset	0x99							
Read Id	0x9F	Data(0)	Data(1)	Data(2)				System will only read 1st three bytes

(1) Only the first data byte is shown, data continues.

More detailed information on the various read operations supported are shown in [Table 8-4](#).

**Table 8-4. SPI Flash Supported Read Operation Details**

READ TYPE <sup>(2)</sup>	NUMBER OF LINES FOR OP-CODE <sup>(1)</sup>	NUMBER OF LINES FOR ADDRESS	NUMBER OF LINES FOR DUMMY BYTES	NUMBER OF LINES FOR RETURN DATA
Fast Read (1/1)	1	1	1	1
Dual Read (1/2)	1	1	1	2
2X Read (2/2)	1	2	2	2
Quad Read (1/4)	1	1	1	4
4X Read (4/4)	1	4	4	4

(1) System does not support Read op-codes being spread across more than one data line.

(2) Flash vendors have diverged in naming and controlling their various read capabilities. As such, the Host needs to be very careful to fully understand what is and what is not supported by the DLPC230S-Q1. In general, for the supported devices, the DLPC230S-Q1 only supports "Extended SPI" or "SPI Mode" (as defined in the various Flash Data Sheets). It does not support "Dual SPI Mode", "Quad SPI Mode", "QPI", "QPI Mode", "Dual QPI", "Quad QPI", "DTR", or "DDR". If uncertain, most devices will support "Fast Reads" in a manner that is consistent with the DLPC230S-Q1.

**Table 8-5. DLPC230S-Q1 Compatible SPI Flash Device Options**

DENSITY (M-BITS) <sup>(2)</sup> <sup>(3)</sup>	VENDOR	PART NUMBER	PACKAGE SIZE
<b>3.3-V Compatible Devices</b>			
128	Micron <sup>(1)</sup>	MT25QL128ABA8ESF-OAAT	SO16
128	Macronix	MX25L12835FMR-10G	SO16
128	Macronix	MX25L12845GMR-10G	SO16
128	Macronix	MX25L12839FXDQ-10G	BGA25

- (1) Care should be used when considering Numonyx versions of Micron serial flash devices as they typically do not have the 4KB sector size needed to be DLPC230S-Q1 compatible.
- (2) For any devices not listed on this table, special care should be taken to insure that the requirements shown in [Table 8-2](#) and [Table 8-3](#) are met.
- (3) The boot application writes to the flash device status register once per 256 bytes during programming. Most flash devices discard status register writes when the status content does not change. Some flash parts, such as Micron N25Q128A13ESFA0F, do not block status writes when the status data is repeated. This causes the status register to exceed its maximum write limit after several programming cycles, making them incompatible with the DLPC230S-Q1. Note that the main application does not write to the status register.

While the DLPC230S-Q1 supports a variety of clock rates and read operation types, it does have a minimum flash read bandwidth requirement which is shown in [Table 8-6](#). This minimum read bandwidth can be met in any number of different ways, with the variables being clock rate and read type. The Host is required to select a flash device which can meet this minimum read bandwidth using the DLPC230S-Q1 supported interface capabilities. It should be noted that the Host will specify to the system (via flash parameter) the maximum supported clock rate as well as the supported read types for their selected flash device, with which the DLPC230S-Q1 SW will automatically select an appropriate combination to maximize this bandwidth (which should at least meet the minimum bandwidth requirement assuming a solution exists per the specified parameters).

**Table 8-6. SPI Flash Interface Bandwidth Requirements**

PARAMETER	MIN	MAX	UNIT
FLSH_RD <sub>BW</sub> Flash Read Interface Bandwidth	47.00		Mbps

### 8.3.5 Serial Flash Programming

The serial flash can be programmed through the DLPC230S-Q1 using Host commands through the SPI or I<sup>2</sup>C command and control interface.

### 8.3.6 Host Command and Diagnostic Processor Interfaces

The DLPC230S-Q1 provides an interface port for Host commands as well as an interface port for a *diagnostic processor*. There are two external communication ports dedicated for this use, one SPI interface and one I<sup>2</sup>C interface. The host is allowed to specify (via ASIC input pin) which port will be used for which purpose (for example, Host Command Interface → SPI, therefore "diagnostic processor" → I<sup>2</sup>C - or they can be reversed).

The timing requirements for the SPI interface are shown in [Section 6.16](#). The timing requirements for the I<sup>2</sup>C interface are shown in [Section 6.17](#). The I<sup>2</sup>C slave address pair is 36h/37h.

### 8.3.7 GPIO Supported Functionality

The DLPC230S-Q1 provides 32 general purpose I/O that are available to support a variety of functions for a number of different product configurations. In general, most of these I/O will only support one specific function based on a specific product configuration, although that function may be different for a different product configuration. There are also a few of these I/O that have been reserved for use by the Host for whatever function they might require. In addition, most of these I/O can also be made available for TI test and debug use. Definitions for the HUD and Headlight product configurations are shown in [Table 8-7](#) and [Table 8-8](#).

**Table 8-7. GPIO Supported Functionality - HUD Product Configuration**

GPIO	SIGNAL NAME	DESCRIPTION <sup>(1)</sup>
GPIO_00	PMIC_CNTRL_OUT (input)	LED control feedback from the TPS99000S-Q1. An external pull-down resistor should be used (connects to TPS99000S-Q1 Drive Enable).
GPIO_01	PMIC_SEQ_STRT (output)	Sequence start output from the DLPC230S-Q1. This should be connected to the TPS99000S-Q1 to time LED related actions and shadow TPS99000S-Q1 configuration registers. An external pull-down resistor should be used.
GPIO_02	PMIC_COMP_OUT (input)	LED optical comparison feedback. This is used to count light pulses during each frame. This signal is active-low. An external pull-down resistor should be used.
GPIO_03	PMIC_LED_SEN (output)	LED Shunt Enable - shunts current from LEDs to allow faster LED turn-off. An external pull-down resistor should be used.
GPIO_04	PMIC_LED_DEN (output)	LED FET Drive Enable - enables LED current switching and defines LED pulse length. An external pull-down resistor should be used.
GPIO_05	Reserved for Future Use	An external pull-down resistor should be used
GPIO_06	Host Available	Available for general host use via Host Commands
GPIO_07	Host Available	Available for general host use via Host Commands
GPIO_08	Host Available	Available for general host use via Host Commands
GPIO_09	Reserved for Future Use	An external pull-down resistor should be used
GPIO_10	Reserved for Future Use	An external pull-down resistor should be used
GPIO_11	Reserved for Future Use	An external pull-down resistor should be used
GPIO_12	Reserved for Future Use	An external pull-down resistor should be used
GPIO_13	Reserved for Future Use	An external pull-down resistor should be used
GPIO_14	Reserved for Future Use	An external pull-down resistor should be used
GPIO_15	PMIC_WD1 (output)	Periodic signal that the DLPC230S-Q1 processor generates during normal operation. TPS99000S-Q1 monitors this signal and reports if this signal stops pulsing. An external pull-down resistor should be used.
GPIO_16	Reserved for Future Use	An external pull-down resistor should be used
GPIO_17	Host Available	Available for general host use via Host Commands
GPIO_18	Reserved for Future Use	An external pull-down resistor should be used
GPIO_19	Reserved for Future Use	An external pull-down resistor should be used
GPIO_20	Reserved for Future Use	An external pull-down resistor should be used
GPIO_21	Reserved for Future Use	An external pull-down resistor should be used
GPIO_22	Reserved for Future Use	An external pull-down resistor should be used
GPIO_23	Reserved for Future Use	An external pull-down resistor should be used
GPIO_24	Reserved for Future Use	An external pull-down resistor should be used
GPIO_25	Reserved for Future Use	An external pull-down resistor should be used
GPIO_26	Host Available	Available for general host use via Host Commands
GPIO_27	Host Available	Available for general host use via Host Commands
GPIO_28	Host Available	Available for general host use via Host Commands
GPIO_29	Host Available	Available for general host use via Host Commands
GPIO_30	Host Available	Available for general host use via Host Commands
GPIO_31	Host Available	Available for general host use via Host Commands

(1) It is recommended that all unused Host Available GPIO be configured as a logic '0' output and be left unconnected in the system. If this is not done, an external pull-down resistor ( $\leq 10$  k $\Omega$ ) should be used to avoid floating inputs.

**Table 8-8. GPIO Supported Functionality - Headlight Product Configuration**

GPIO	SIGNAL NAME	DESCRIPTION <sup>(1)</sup>
GPIO_00	HL_PWM0 (output)	PWM 0 Output - This can be used for general purposes such as controlling the level of an external light source.
GPIO_01	PMIC_SEQ_STRT (output)	Sequence start output from the DLPC230S-Q1. This should be connected to the TPS99000S-Q1 to time LED related actions and shadow TPS99000S-Q1 configuration registers. An external pull-down resistor should be used.
GPIO_02	HL_PWM1(output)	PWM 1 Output - This can be used for general purposes such as controlling the level of an external light source.
GPIO_03	Reserved for Future Use	An external pull-down resistor should be used
GPIO_04	Reserved for Future Use	An external pull-down resistor should be used
GPIO_05	Reserved for Future Use	An external pull-down resistor should be used
GPIO_06	Host Available	Available for general host use via Host Commands
GPIO_07	Host Available	Available for general host use via Host Commands
GPIO_08	Host Available	Available for general host use via Host Commands
GPIO_09	Reserved for Future Use	An external pull-down resistor should be used
GPIO_10	Reserved for Future Use	An external pull-down resistor should be used
GPIO_11	Reserved for Future Use	An external pull-down resistor should be used
GPIO_12	Reserved for Future Use	An external pull-down resistor should be used
GPIO_13	Reserved for Future Use	An external pull-down resistor should be used
GPIO_14	Reserved for Future Use	An external pull-down resistor should be used
GPIO_15	PMIC_WD1 (output)	Periodic signal that the DLPC230S-Q1 processor generates during normal operation. TPS99000S-Q1 monitors this signal and reports if this signal stops pulsing. An external pull-down resistor should be used.
GPIO_16	Reserved for Future Use	An external pull-down resistor should be used
GPIO_17	HL_PWM2 (output)	PWM 2 Output - This can be used for general purposes such as controlling the level of an external light source.
GPIO_18	EXT_SMPL	Connects to TPS99000S-Q1 EXT_SMPL input. This sequence-aligned signal can be configured to trigger TPS99000S-Q1 ADC sampling.
GPIO_19	Reserved for Future Use	An external pull-down resistor should be used
GPIO_20	Reserved for Future Use	An external pull-down resistor should be used
GPIO_21	Reserved for Future Use	An external pull-down resistor should be used
GPIO_22	Reserved for Future Use	An external pull-down resistor should be used
GPIO_23	Reserved for Future Use	An external pull-down resistor should be used
GPIO_24	Reserved for Future Use	An external pull-down resistor should be used
GPIO_25	Reserved for Future Use	An external pull-down resistor should be used
GPIO_26	Host Available	Available for general host use via Host Commands
GPIO_27	Host Available	Available for general host use via Host Commands
GPIO_28	Host Available	Available for general host use via Host Commands
GPIO_29	Host Available	Available for general host use via Host Commands
GPIO_30	Host Available	Available for general host use via Host Commands
GPIO_31	Host Available	Available for general host use via Host Commands

(1) It is recommended that all unused Host Available GPIO be configured as a logic '0' output and be left unconnected in the system. If this is not done, an external pull-down resistor ( $\leq 10$  k $\Omega$ ) should be used to avoid floating inputs.

### 8.3.8 Built-In Self Test (BIST)

The DLPC230S-Q1 provides a significant amount of BIST support to help ensure the operational integrity of the system. This BIST support is divided into two general BIST types, which are Non-Periodic and Periodic.

Non-Periodic BISTs are tests that are typically run one time, and are run outside of normal operation since their activity will disturb the operation of the system. These tests are specified to be run either by a Flash parameter or by a Host command. The Flash parameter specifies which tests are to be run during system power-up and initialization. The Host command is used to select and specify the running of these tests when the system is in Standby Mode (often just before the system is powered down). Some examples of non-periodic tests are: tests for all of the ASIC memories, tests for the main data processing path, and testing of the DMD memory.

Periodic BISTs are tests that are run on an almost continual basis during normal ASIC operation. These tests are managed (set up, enabled, results gathered and evaluated) automatically by the ASIC embedded software. Some examples of periodic tests are: tuning and verification of the DMD High-Speed Interface, input source monitoring (clock, active pixels, active lines), and external video checksum monitoring.

For more information on BISTs, refer to *DLPC230S-Q1 Programmer's Guide (DLPU041 for HUD and DLPU048 for Headlight)*.

### 8.3.9 EEPROMs

The DLPC230S-Q1 may optionally use an external I<sup>2</sup>C EEPROM memory device for storage of calibration data as an alternative to storing calibration data in the SPI flash memory. The EEPROM must be connected to the designated DLPC230S-Q1 master I<sup>2</sup>C interface (MSTR\_XXX).

The DLPC230S-Q1 supports the EEPROM devices listed in [Table 8-9](#).

**Table 8-9. DLPC230S-Q1 Supported EEPROMs**

MANUFACTURER	PART NUMBER	DENSITY (Kb)	PACKAGE SIZE
STMicro	M24C64A125	64	S08
STMicro	M24C128A125	128	S08
Atmel	A24C64D	64	S08
Atmel	A24C128C	128	S08

### 8.3.10 Temperature Sensor

The DLPC230S-Q1 requires an external temperature sensor (TMP411) to measure the DMD temperature through a remote temperature sense diode residing within the DMD. The DLPC230S-Q1 will also read the local temperature reported by the TMP411 device. The TMP411 must be connected to the designated DLPC230S-Q1 master I<sup>2</sup>C interface (MSTR\_XXX).

The DLPC230S-Q1 uses an averaged DMD temperature reading to manage the thermal environment and/or operation of the DMD. This management occurs over the full range of temperatures supported by the DMD. This temperature reading is used change sequence operation across the temperature range, and park the DMD when it is operated outside of its allowable temperature specification.

### 8.3.11 Debug Support

The DLPC230S-Q1 contains a test point output port, TSTPT\_(7:0), which provides the Host with the ability to specify a number of initial system configurations, as well as to provide for ASIC debug support. These test points are tri-stated while reset is applied, are sampled as inputs approximately 1.5 μs after reset is released, and then switch to outputs once the input values have been sampled. The sampled and captured input state for each of these signals is used to configure initial system configurations as specified in the table Pin Functions - Parallel Port Input Data and Control in [Section 5](#).

There are three other signals (JTAGTDO(3:1)) that are sampled as inputs approximately 1.5 μs after reset is released, and then switched to outputs. The sampled and captured state for each of these JTAGTDO signals is used to configure the initial test mode output state of the TSTPT\_(7:0) signals. [Table 8-10](#) defines the test mode selection for a few programmable output states for TSTPT\_(7:0) as defined by JTAGTDO(3:1). For normal use (that is, no debug required), the default state of x111 (using weak internal pull-ups) should be used to allow for the normal use of these JTAG TDO signals.

To allow TI to make use of this debug capability, a jumper to an external pull-down is recommended for JTAGTDO(3:1).

**Table 8-10. Test Mode Selection Scenario Defined by JTAGTDO(3:1)**

TSTPT_(7:0) OUTPUT <sup>(1)</sup>	JTAGTDO(3:1) CAPTURED VALUE	
	x111 (DEFAULT) (NO SWITCHING ACTIVITY)	x010 CLOCK DEBUG OUTPUT
TSTPT(0)	HI-Z	60 MHz
TSTPT(1)	HI-Z	30 MHz
TSTPT(2)	HI-Z	7.5 MHz
TSTPT(3)	HI-Z	LOW
TSTPT(4)	HI-Z	15 MHz
TSTPT(5)	HI-Z	60 MHz
TSTPT(6)	HI-Z	LOW
TSTPT(7)	HI-Z	LOW

(1) These are only the default output selections. Software can reprogram the selection at any time.

## 8.4 Device Functional Modes

The DLPC230S-Q1 has three operational modes which are enabled via software command via the Host control interface. These modes are Standby, Display, and Calibration.

### 8.4.1 Standby Mode

The system will automatically enter Standby mode after power is applied. This is a reduced functional mode that allows Flash update operations and Non-Periodic test operations. The DMD will be parked while the system is operating in this mode and no source may be displayed.

### 8.4.2 Display Mode

This is the main operational mode of the system. In this mode, normal display activities occur. In this mode the system may display video data and execute periodic BISTs. After system initialization, a host command can be used to transition to this mode from Standby mode. Alternatively, a flash configuration setting can be set to allow the system to automatically transition from standby to display mode after system initialization.

### 8.4.3 Calibration Mode

This mode is used to calibrate the system's light sources for the desired display properties. For head-up display applications, this includes the ability to adjust individual color light sources to achieve the desired brightness and color point.

## 9 Application and Implementation

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### Note

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

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### 9.1 Application Information

The DLPC230S-Q1 is designed to support projection-based automotive applications such as head-up display (HUD) and high resolution headlight.

This DLP<sup>®</sup> Products chipset consists of three components—the DLP5530S-Q1 Digital Micromirror Device (DMD), the DLPC230S-Q1, and the TPS99000S-Q1. The DMD is a light modulator consisting of tiny mirrors that are used to form and project images. The DLPC230S-Q1 is a controller for the DMD; it formats incoming video sources and controls the timing of the DMD illumination sources and the DMD in order to display the incoming video source. The TPS99000S-Q1 is a controller for the illumination sources (LEDs or lasers) and a management IC for the entire chipset. In conjunction, the DLPC230S-Q1 and the TPS99000S-Q1 can also be used for system-level monitoring, diagnostics, and failure detection features.

## 9.2 Typical Application

### 9.2.1 Head-Up Display

Figure 9-1 shows the system block diagram for a DLP® technology HUD.

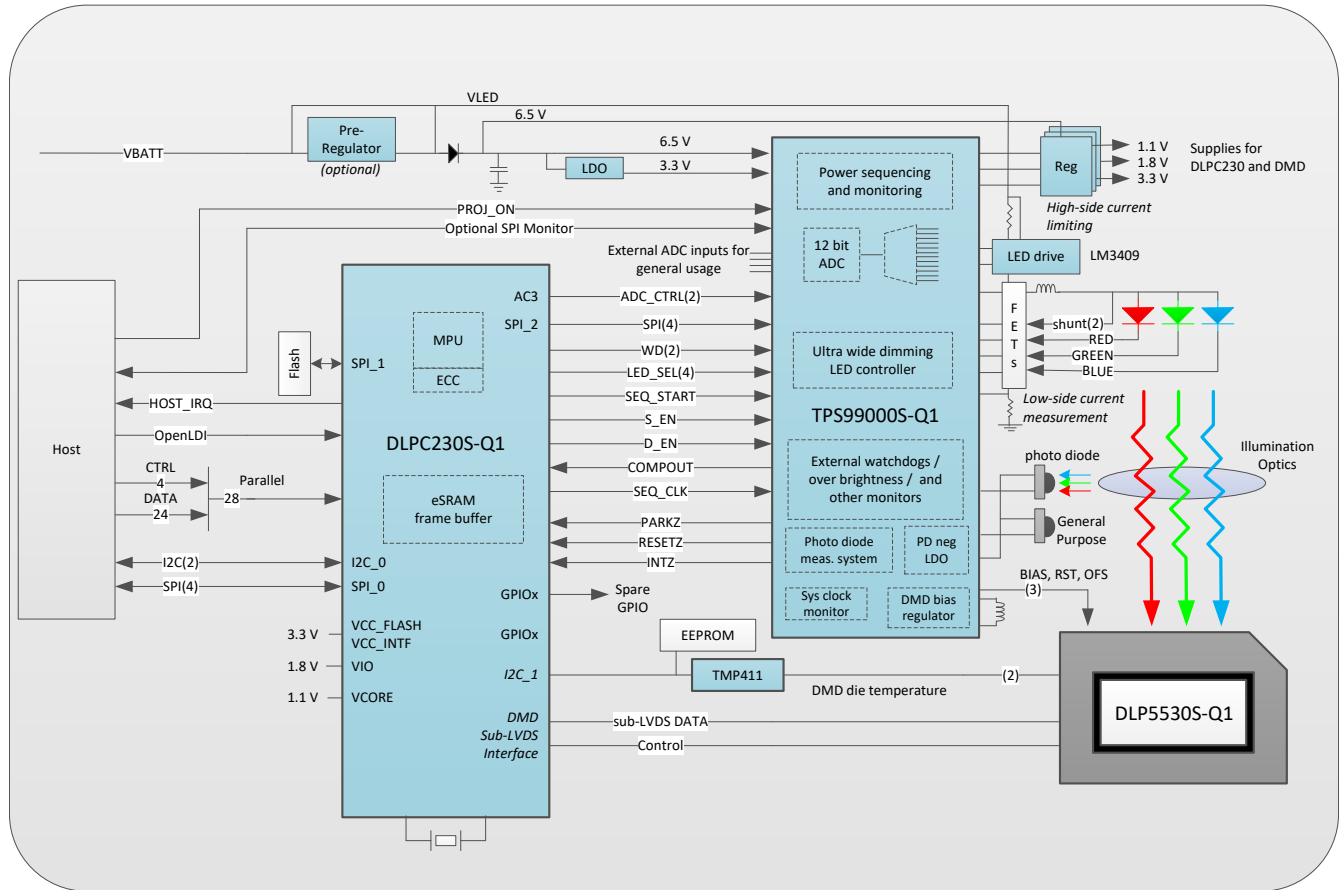


Figure 9-1. HUD System Block Diagram

#### 9.2.1.1 Design Requirements

The DLPC230S-Q1 is a controller for the DMD and the timing of the RGB LEDs in the HUD. It requests the proper timing and amplitude from the LEDs to achieve the requested color and brightness from the HUD across the entire operating range. It synchronizes the DMD with these LEDs in order to display full-color video content sent by the host.

The DLPC230S-Q1 receives command and input video data from a host processor in the vehicle. Read and write (R/W) commands can be sent using either the I<sup>2</sup>C bus or SPI bus. The bus that is not being used for R/W commands can be used as a read-only bus for diagnostic purposes. Input video can be sent over an OpenLDI bus or a parallel 24-bit bus. The SPI flash memory provides the embedded software for the DLPC230S-Q1's embedded processor, color calibration data, and default settings. The TPS99000S-Q1 provides diagnostic and monitoring information to the DLPC230S-Q1 using a SPI bus and several other control signals such as PARKZ, INTZ, and RESETZ to manage power-up and power-down sequencing. The DLPC230S-Q1 interfaces to a TMP411 via I<sup>2</sup>C for temperature information.

The outputs of the DLPC230S-Q1 are LED drive information to the TPS99000S-Q1, control signals to the DMD, and monitoring and diagnostics information to the host processor. Based on a host requested brightness and the operating temperature, the DLPC230S-Q1 determines the proper timing and amplitudes for the LEDs. It passes this information to the TPS99000S-Q1 using a SPI bus and several additional control signals such as D\_EN,



S\_EN, and SEQ\_STRT. It controls the DMD mirrors by sending data over a sub-LVDS bus. It can alert the host about any critical errors using a HOST\_IRQ signal.

The TPS99000S-Q1 is a highly-integrated mixed-signal IC that controls DMD power, the analog response of the LEDs, and provides monitoring and diagnostics information for the HUD system. The power sequencing and monitoring blocks of the TPS99000S-Q1 properly power up the DMD, provide accurate DMD voltage rails, as well as monitor the system's power rails during operation. The integration of these functions into one IC significantly reduces design time and complexity. The highly accurate photodiode (PD) measurement system and the dimming controller block precisely control the LED response. This enables a DLP technology HUD to achieve a very high dimming range (> 5000:1) with accurate brightness and color across the temperature range of the system. Finally, the TPS99000S-Q1 has several general-purpose ADCs that developers can use for system-level monitoring, such as over-brightness detection.

The TPS99000S-Q1 receives inputs from the DLPC230S-Q1, power rail voltages for monitoring, a photodiode that is used to measure LED response, the host processor, and potentially several other ADC ports. The DLPC230S-Q1 sends commands to the TPS99000S-Q1 over a SPI port and several other control signals. The TPS99000S-Q1 includes watchdogs to monitor the DLPC230S-Q1 and ensure that it is operating as expected. The power rails are monitored by the TPS99000S-Q1 to detect power failures or glitches and request a proper power down of the DMD in case of an error. The photodiode's current is measured and amplified using a transimpedance amplifier (TIA) within the TPS99000S-Q1. The host processor can read diagnostics information from the TPS99000S-Q1 using a dedicated SPI bus, adding an independent monitoring path from the host processor. Additionally the host can request the system to be turned on or off using a PROJ\_ON signal. The TPS99000S-Q1 has several general-purpose ADCs that can be used to implement other system features such as over-brightness and over-temperature detection.

The outputs of the TPS99000S-Q1 are LED drive signals, diagnostic information, and error alerts to the DLPC230S-Q1. The TPS99000S-Q1 has signals connected to the LM3409 buck controller for high power LEDs and to discrete hardware that control the LEDs. The TPS99000S-Q1 can output diagnostic information to the host and the DLPC230S-Q1 over two SPI busses. It also has signals such as RESETZ, PARKZ, and INTZ that can be used to trigger power down or reset sequences.

The DMD is a micro-electro-mechanical system (MEMS) device that receives electrical signals as an input (video data) and produces a mechanical output (mirror position). The electrical interface to the DMD is a sub-LVDS interface driven with the DLPC230S-Q1. The mechanical output is the state of more than 1.3 million mirrors in the DMD array that can be tilted  $\pm 12^\circ$ . In a projection system, the mirrors are used as pixels in order to display an image.

## 10 Power Supply Recommendations

### 10.1 Power Supply Management

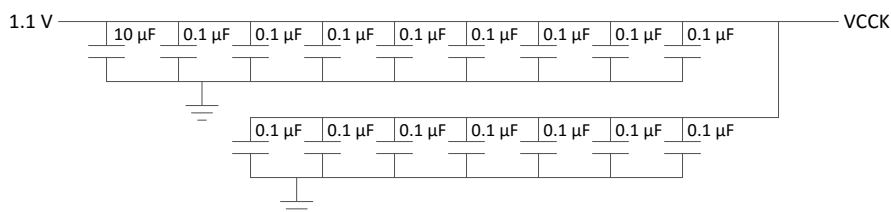
The TPS99000S-Q1 manages power for the DLPC230S-Q1 and DMD. See [Section 6.12](#) for all power sequencing and timing requirements.

### 10.2 Hot Plug Usage

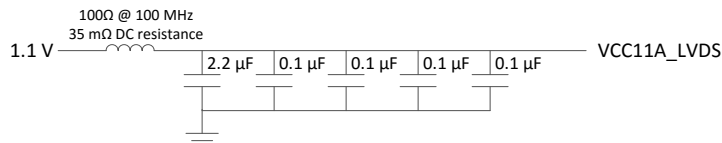
The DLPC230S-Q1 does not support Hot Plug use (for itself or for any DMD connected to the system). As such, the system should always be powered down prior to removal of the ASIC or DMD from any system.

### 10.3 Power Supply Filtering

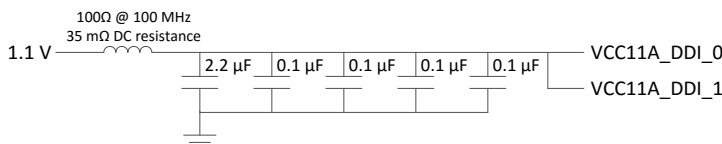
The following filtering circuits are recommended for the various supply inputs. High frequency 0.1- $\mu\text{F}$  capacitors should be evenly distributed amongst the power balls and placed as close to the power balls as possible.



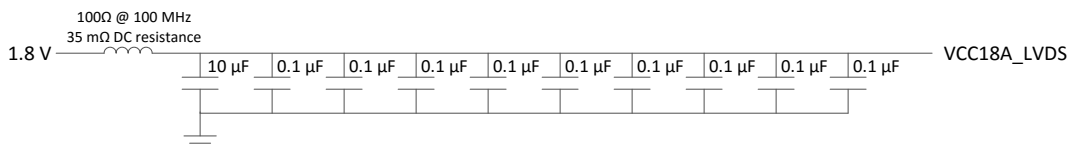
**Figure 10-1. VCCK Recommended Filter**



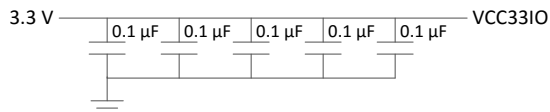
**Figure 10-2. VCC11A\_LVDS Recommended Filter**



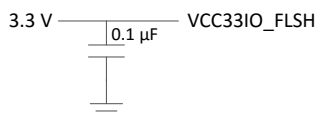
**Figure 10-3. VCC11A\_DDI Recommended Filter**



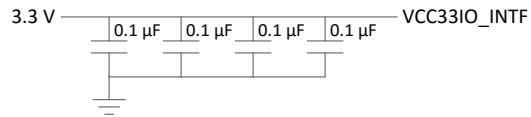
**Figure 10-4. VCC18A\_LVDS Recommended Filter**



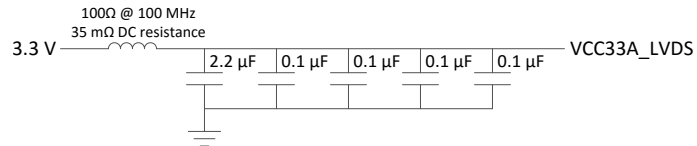
**Figure 10-5. VCC33IO Recommended Filter**



**Figure 10-6. VCC33IO\_FLSH Recommended Filter**



**Figure 10-7. VCC33IO\_INTF Recommended Filter**



**Figure 10-8. VCC33A\_LVDS Recommended Filter**

## 11 Layout

### 11.1 Layout Guidelines

#### 11.1.1 PCB Layout Guidelines for Internal ASIC PLL Power

The following guidelines are recommended to achieve desired ASIC performance relative to the internal PLL. The DLPC230S-Q1 contains two internal PLLs which have dedicated analog supplies (VCC11AD\_PLLM, GND11AD\_PLLM, VCC11AD\_PLLD, GND11AD\_PLLD). At a minimum, VCC11AD\_PLLx power and GND11AD\_PLLx ground pins should be isolated using a simple passive filter consisting of two series Ferrites and two shunt capacitors (to widen the spectrum of noise absorption). Recommended values and layout are shown in [Table 11-1](#) and [Figure 11-1](#) respectively.

**Table 11-1. Recommended PLL Filter Components**

COMPONENT	PARAMETER	RECOMMENDED VALUE	UNIT
Shunt Capacitor	Capacitance	0.1	$\mu\text{F}$
Shunt Capacitor	Capacitance	1.0	$\mu\text{F}$
Series Ferrite	Impedance at 100 MHz	> 100	$\Omega$
	DC Resistance	< 0.40	

Since the PCB layout is critical to PLL performance, it is vital that the quiet ground and power are treated like analog signals. Additional design guidelines are as follows:

- All four components should be placed as close to the ASIC as possible
- It's especially important to keep the leads of the high frequency capacitors as short as possible
- A capacitor of each value should be connected across VCC11AD\_PLLM / GND11AD\_PLLM and VCC11AD\_PLLD / GND11AD\_PLLD respectively on the ASIC side of the Ferrites
- VCC11AD\_PLLM and VCC11AD\_PLLD must be a single trace from the DLPC230S-Q1 to both capacitors and then through the series ferrites to the power source
- The power and ground traces should be as short as possible, parallel to each other, and as close as possible to each other

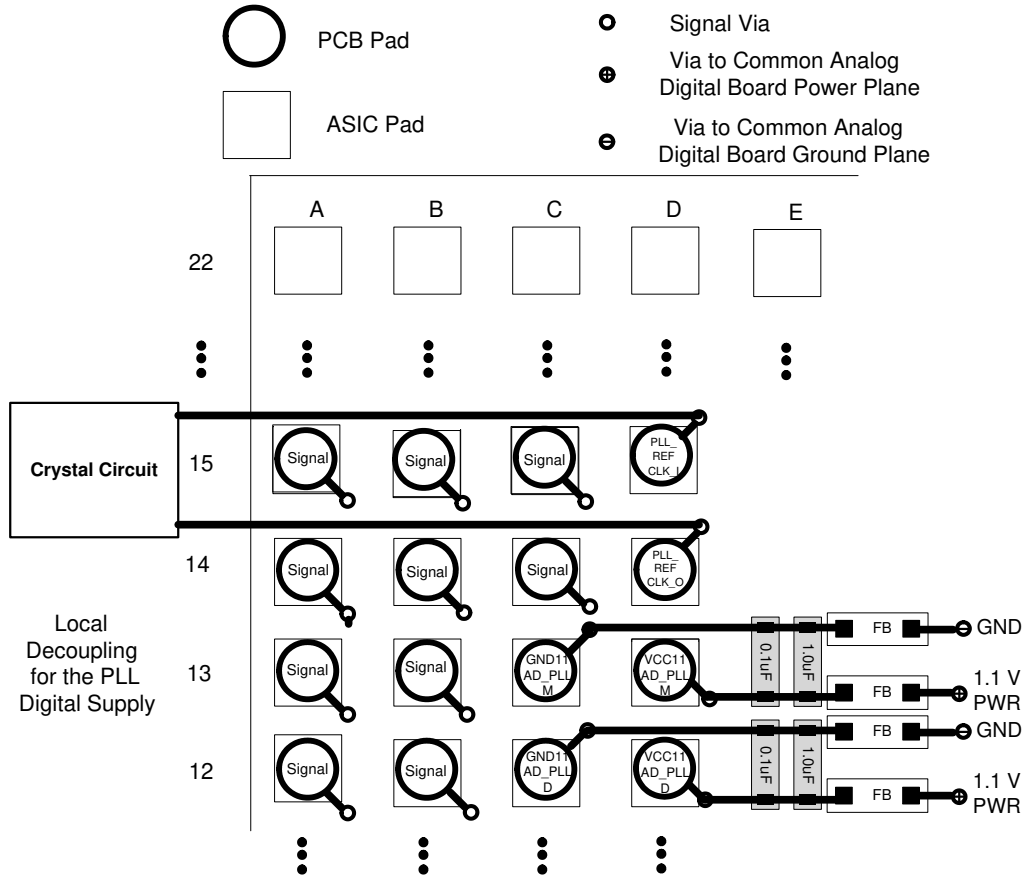
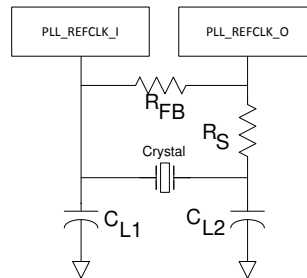


Figure 11-1. PLL Filter Layout

### 11.1.2 DLPC230S-Q1 Reference Clock

The DLPC230S-Q1 requires an external reference clock to feed its internal PLL. A crystal or oscillator can supply this reference. The recommended crystal configurations and reference clock frequencies are listed in [Table 11-2](#), with additional required discrete components shown in [Figure 11-2](#) and defined in [Table 11-2](#).



- A.  $C_L$  = Crystal load capacitance  
 B.  $R_{FB}$  = Feedback Resistor

**Figure 11-2. Discrete Components Required When Using Crystal**

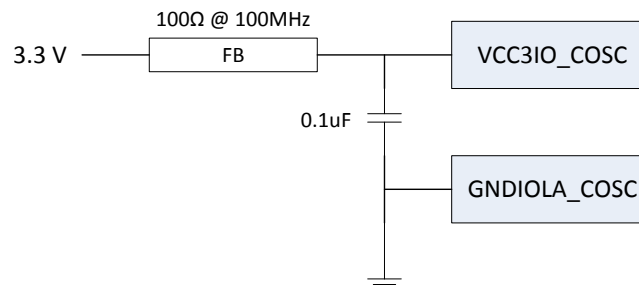
#### 11.1.2.1 Recommended Crystal Oscillator Configuration

**Table 11-2. Recommended Crystal Configuration**

PARAMETER	RECOMMENDED	UNIT
Crystal circuit configuration	Parallel resonant	
Crystal type	Fundamental (first harmonic)	
Crystal nominal frequency	16	MHz
Crystal frequency tolerance (including accuracy, temperature, aging and trim sensitivity)	±200	PPM
Maximum crystal equivalent series resistance (ESR)	50	Ω
Crystal load capacitance	10	pF
Temperature range	–40°C to +105°C	°C
Drive level (nominal)	100	μW
$R_{FB}$ feedback resistor (nominal)	1	MΩ
$C_{L1}$ external crystal load capacitor	See equation in <a href="#">(1)</a>	pF
$C_{L2}$ external crystal load capacitor	See equation in <a href="#">(2)</a>	pF
PCB layout	A ground isolation ring around the crystal is recommended	

- (1)  $CL1 = 2 \times (CL - C_{stray\_pll\_refclk\_i})$ , where:  $C_{stray\_pll\_refclk\_i}$  = Sum of package and PCB stray capacitance at the crystal pin associated with the ASIC pin  $pll\_refclk\_i$ .  
 (2)  $CL2 = 2 \times (CL - C_{stray\_pll\_refclk\_o})$ , where:  $C_{stray\_pll\_refclk\_o}$  = Sum of package and PCB stray capacitance at the crystal pin associated with the ASIC pin  $pll\_refclk\_o$ .

The crystal circuit in the DLPC230S-Q1 ASIC has dedicated power ( $VCC3IO\_COSC$ ) and ground ( $GNDIOLA\_COSC$ ) pins, with the recommended filtering shown in [Figure 11-3](#).



**Figure 11-3. Crystal Power Supply Filtering**

**Table 11-3. DLPC230S-Q1 Recommended Crystal Parts**

MANUFACTURER	PART NUMBER	SPEED	FREQUENCY TOLERANCE, FREQUENCY STABILITY, AGING/YEAR	ESR	LOAD CAPACITANCE	OPERATING TEMPERATURE
TXC	AM16070006 <sup>(1)</sup>	16 MHz	Freq Tolerance: ±10 ppm	50-Ω max	10 pF	-40°C to +125°C
			Freq Stability: ±50 ppm			
			Aging/Year: ±3 ppm			

(1) This device requires a 3-kΩ series resistor to limit power.

If an external oscillator is used, the oscillator output must drive the PLL\_REFCLK\_O pin on the DLPC230S-Q1 ASIC, the PLL\_REFCLK\_I pin should be left unconnected, and the OSC\_BYPASS pin must = logic HIGH.

### 11.1.3 DMD Interface Layout Considerations

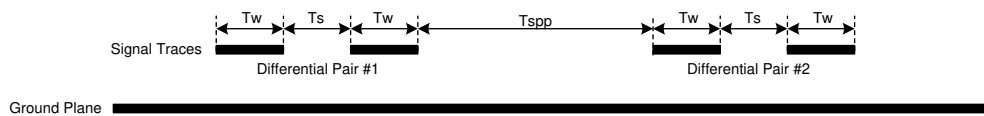
The DLPC230S-Q1 ASIC sub-LVDS HS/LS differential interface waveform quality and timing is dependent on the total length of the interconnect system, the spacing between traces, the characteristic impedance, etch losses, and how well matched the lengths are across the interface. Thus, ensuring positive timing margin requires attention to many factors.

DLPC230S-Q1 I/O timing parameters as well as DMD I/O timing parameters can be found in their corresponding data sheets. Similarly, PCB routing mismatch can be budgeted and met through controlled PCB routing. PCB design recommendations are provided in [Table 11-4](#) and [Figure 11-4](#) as a starting point for the customer.

**Table 11-4. PCB Recommendations for DMD Interface**

PARAMETER <sup>(1) (2)</sup>		MIN	MAX	UNIT
T <sub>W</sub>	Trace Width	4		mils
T <sub>S</sub>	Intra-lane Trace Spacing	4		mils
T <sub>SPP</sub>	Inter-lane Trace Spacing	2 * (T <sub>S</sub> + T <sub>W</sub> )		mils
R <sub>BGR</sub>	Resistor - Bandgap Reference	42.2 (1%)		kΩ

- (1) Recommendations to achieve the desired nominal differential impedance as specified by Tx<sub>load</sub> in [Section 6.7](#) and [Section 6.8](#).
- (2) If using the minimum trace width and spacing to escape the ASIC ball field, widening these out after escape would be desirable if practical to achieve the target 100-Ω impedance (e.g. to reduce transmission line losses).



**Figure 11-4. DMD Differential Layout Recommendations**

#### 11.1.4 General PCB Recommendations

TI recommends 1-oz copper power planes and 2-oz copper ground planes in the PCB design to achieve the required thermal connectivity, with:

- A minimum of 4 power and ground planes
- A copper plane beneath the thermal ball array containing a via farm with the following attributes
  - Copper plane area (top side of PCB, under package) = 8.0 mm × 8.0 mm
  - Copper plane area (bottom side of PCB, opposite of package) = 6.0 mm × 6.0 mm
  - Thermal via quantity = 7 × 7 array of vias
  - Thermal via size = 0.25 mm (10 mils)
  - Thermal via plating thickness = 0.05-mm (2-mils) wall thickness
- PCB copper coverage per layer
  - Power and Ground layers: 90% minimum coverage
  - Top/Bottom signal layers (ground fill to achieve coverage): 70% minimum coverage with 1.5-oz copper.

#### 11.1.5 General Handling Guidelines for Unused CMOS-Type Pins

To avoid potentially damaging current caused by floating CMOS input-only pins, TI recommends that unused ASIC input pins be tied through a pull-up resistor to its associated power supply or a pull-down to ground unless specifically noted otherwise in [Section 5](#). For ASIC inputs with an internal pull-up or pull-down resistors, it is unnecessary to add an external pull-up or pull-down unless specifically recommended. Note that internal pull-up and pull-down resistors are weak and should not be expected to drive the external line. When external pull-up or pull-down resistors are needed for pins that have built-in weak pull-ups or pull-downs, use the value specified in [Table 5-2](#).

Unused output-only pins should never be tied directly to power or ground, but can be left open.

When possible, TI recommends that unused bidirectional I/O pins be configured to their output state such that the pin can be left open. If this control is not available and the pins may become an input, then they should be pulled-up (or pulled-down) using an appropriate, dedicated resistor.



### 11.1.6 Maximum Pin-to-Pin, PCB Interconnects Etch Lengths

**Table 11-5. Max Pin-to-Pin PCB Interconnect Recommendations - DMD**

ASIC INTERFACE	SIGNAL INTERCONNECT TOPOLOGY <sup>(1)</sup> <sup>(2)</sup>		UNIT
	SINGLE BOARD SIGNAL ROUTING LENGTH	MULTI-BOARD SIGNAL ROUTING LENGTH	
DMD_HS0_CLK_P DMD_HS0_CLK_N	6.0 (152.4)	See <sup>(3)</sup>	in (mm)
DMD_HS0_WDATA0_P DMD_HS0_WDATA0_N	6.0 (152.4)	See <sup>(3)</sup>	in (mm)
DMD_HS0_WDATA1_P DMD_HS0_WDATA1_N			
DMD_HS0_WDATA2_P DMD_HS0_WDATA2_N			
DMD_HS0_WDATA3_P DMD_HS0_WDATA3_N			
DMD_HS0_WDATA4_P DMD_HS0_WDATA4_N			
DMD_HS0_WDATA5_P DMD_HS0_WDATA5_N			
DMD_HS0_WDATA6_P DMD_HS0_WDATA6_N			
DMD_HS0_WDATA7_P DMD_HS0_WDATA7_N			
DMD_HS1_CLK_P DMD_HS1_CLK_N			
DMD_HS1_WDATA0_P DMD_HS1_WDATA0_N	6.0 (152.4)	See <sup>(3)</sup>	in (mm)
DMD_HS1_WDATA1_P DMD_HS1_WDATA1_N			
DMD_HS1_WDATA2_P DMD_HS1_WDATA2_N			
DMD_HS1_WDATA3_P DMD_HS1_WDATA3_N			
DMD_HS1_WDATA4_P DMD_HS1_WDATA4_N			
DMD_HS1_WDATA5_P DMD_HS1_WDATA5_N			
DMD_HS1_WDATA6_P DMD_HS1_WDATA6_N			
DMD_HS1_WDATA7_P DMD_HS1_WDATA7_N			
DMD_LS0_CLK_P DMD_LS0_CLK_N	6.5 (165.1)	See <sup>(3)</sup>	in (mm)
DMD_LS0_WDATA_P DMD_LS0_WDATA_N	6.5 (165.1)	See <sup>(3)</sup>	in (mm)
DMD_LS0_RDATA	6.5 (165.1)	See <sup>(3)</sup>	in (mm)
DMD_LS1_RDATA	6.5 (165.1)	See <sup>(3)</sup>	in (mm)
DMD_DEN_ARSTZ	N/A	N/A	in (mm)

(1) Max signal routing length includes escape routing.

(2) Multi-board DMD routing length is more restricted due to the impact of the connector.

(3) Due to board variations, these are impossible to define. Any board designs should SPICE simulate with the ASIC IBIS models to ensure signal routing lengths do not exceed requirements.

**Table 11-6. Max Pin-to-Pin PCB Interconnect Recommendations - TPS9900S-Q1**

ASIC INTERFACE	SIGNAL INTERCONNECT TOPOLOGY <sup>(1) (2)</sup>		UNIT
	SINGLE BOARD SIGNAL ROUTING LENGTH	MULTI-BOARD SIGNAL ROUTING LENGTH	
TPS9900S-Q1			
PMIC_LEDSEL(3)	6.0 (152.4)	See <sup>(3)</sup>	in (mm)
PMIC_LEDSEL(2)			
PMIC_LEDSEL(1)			
PMIC_LEDSEL(0)			
PMIC_ADC3_CLK			
PMIC_ADC3_MOSI			
PMIC_ADC3_MISO			
PMIC_SEQ_STRT			

(1) Max signal routing length includes escape routing.

(2) Multi-board DMD routing length is more restricted due to the impact of the connector.

(3) Due to board variations, these are impossible to define. Any board designs should SPICE simulate with the ASIC IBIS models to ensure signal routing lengths do not exceed requirements.

**Table 11-7. High-Speed PCB Signal Routing Matching Requirements**

SIGNAL GROUP LENGTH MATCHING <sup>(1) (2)</sup>				
INTERFACE	SIGNAL GROUP	REFERENCE SIGNAL	MAX MISMATCH	UNIT
DMD <sup>(3)</sup>	DMD_HS0_WDATA0_P DMD_HS0_WDATA0_N	DMD_HS0_CLK_P DMD_HS0_CLK_N	±1.0 (±25.4)	in (mm)
	DMD_HS0_WDATA1_P DMD_HS0_WDATA1_N			
	DMD_HS0_WDATA2_P DMD_HS0_WDATA2_N			
	DMD_HS0_WDATA3_P DMD_HS0_WDATA3_N			
	DMD_HS0_WDATA4_P DMD_HS0_WDATA4_N			
	DMD_HS0_WDATA5_P DMD_HS0_WDATA5_N			
	DMD_HS0_WDATA6_P DMD_HS0_WDATA6_N			
	DMD_HS0_WDATA7_P DMD_HS0_WDATA7_N			
DMD <sup>(4)</sup>	DMD_HS0_x_P	DMD_HS0_x_N	±0.025 (±0.635)	in (mm)
DMD <sup>(3)</sup>	DMD_HS1_WDATA0_P DMD_HS1_WDATA0_N	DMD_HS1_CLK_P DMD_HS1_CLK_N	±1.0 (±25.4)	in (mm)
	DMD_HS1_WDATA1_P DMD_HS1_WDATA1_N			
	DMD_HS1_WDATA2_P DMD_HS1_WDATA2_N			
	DMD_HS1_WDATA3_P DMD_HS1_WDATA3_N			
	DMD_HS1_WDATA4_P DMD_HS1_WDATA4_N			
	DMD_HS1_WDATA5_P DMD_HS1_WDATA5_N			
	DMD_HS1_WDATA6_P DMD_HS1_WDATA6_N			
	DMD_HS1_WDATA7_P DMD_HS1_WDATA7_N			

**Table 11-7. High-Speed PCB Signal Routing Matching Requirements (continued)**

SIGNAL GROUP LENGTH MATCHING <sup>(1) (2)</sup>				
INTERFACE	SIGNAL GROUP	REFERENCE SIGNAL	MAX MISMATCH	UNIT
DMD <sup>(4)</sup>	DMD_HS1_x_P	DMD_HS1_x_N	±0.025 (±0.635)	in (mm)
DMD <sup>(3)</sup>	DMD_LS0_WDATA_P DMD_LS0_WDATA_N	DMD_LS0_CLK_P DMD_LS0_CLK_N	±1.0 (±25.4)	in (mm)
DMD <sup>(4)</sup>	DMD_LS0_x_P	DMD_LS0_x_N	±0.025 (±0.635)	in (mm)
DMD	DMD_LS0_RDATA DMD_LS1_RDATA	N/A	N/A <sup>(5)</sup>	in (mm)
DMD	DMD_DEN_ARSTZ	N/A	N/A	in (mm)
TPS99000S-Q1	PMIC_LEDSEL(3)	PMIC_ADC3_CLK	±1.0 (±25.4)	in (mm)
	PMIC_LEDSEL(2)			
	PMIC_LEDSEL(1)			
	PMIC_LEDSEL(0)			
	PMIC_SEQ_STRT			
	PMIC_ADC3_MOSI			

- (1) These routing requirements are specific to the PCB routing. Internal package routing mismatches in the DLPC230S-Q1 and DMD have already been accounted for in these requirements.
- (2) Training is applied to DMD HS data lines, so defined matching requirements are slightly relaxed.
- (3) This is an inter-pair specification (that is, differential pair to differential pair within the group).
- (4) This is an intra-pair specification (that is, length mismatch between P and N for the same pair).
- (5) For legacy DMD support, the ASIC provides a single-ended low-speed write interface. The primary low-speed write control interface to the DMD is differential. The low-speed read control interface from the DMD is single-ended, and makes use of the differential write clock. As such, a routing mismatch between these is not applicable.

### 11.1.7 Number of Layer Changes

- Single-ended signals: Minimize the number of layer changes.
- Differential signals: Individual differential pairs can be routed on different layers, but the signals of a given pair should not change layers.

### 11.1.8 Stubs

- Stubs should be avoided.

### 11.1.9 Terminations

- No external termination resistors are required on the DMD\_HS or DMD\_LS differential signals.
- The DMD\_LS0\_RDATA and DMD\_LS1\_RDATA single-ended signal paths should include a 10-Ω series termination resistor located as close as possible to the corresponding DMD pin.
- DMD\_DEN\_ARSTZ does not typically require a series resistor, however, for a long trace, one might be needed to reduce undershoot/overshoot.

### 11.1.10 Routing Vias

- The number of vias on each DMD\_HS and DMD\_LS signal should be minimized and should not exceed two. If two are required, one should be placed at each end of the line (one at the ASIC and one at the DMD).

### 11.2 Thermal Considerations

The underlying thermal limitation for the DLPC230S-Q1 is that the maximum operating junction temperature ( $T_J$ ) not be exceeded (this is defined in the [Section 6.3](#)). This temperature is dependent on operating ambient temperature, airflow, PCB design (including the component layout density and the amount of copper used), power dissipation of the DLPC230S-Q1, and power dissipation of surrounding components. The DLPC230S-Q1's package is designed primarily to extract heat through the power and ground planes of the PCB. Thus, copper content and airflow over the PCB are important factors.

TI highly recommends that once the host PCB is designed and built that the thermal performance be measured and validated.

To do this, measure the top center case temperature under the worse case product scenario (max power dissipation, max voltage, max ambient temperature) and validate that the maximum recommended case temperature ( $T_C$ ) is not exceeded. This specification is based on the measured  $\phi_{JT}$  for the DLPC230S-Q1 package and provides a relatively accurate correlation to junction temperature. Take care when measuring this case temperature to prevent accidental cooling of the package surface. TI recommends a small (approximately 40 gauge) thermocouple. The bead and thermocouple wire should contact the top of the package and be covered with a minimal amount of thermally conductive epoxy. The wires should be routed closely along the package and the board surface to avoid cooling the bead through the wires.

## 12 Device and Documentation Support

### 12.1 Device Support

#### 12.1.1 Third-Party Products Disclaimer

TI'S PUBLICATION OF INFORMATION REGARDING THIRD-PARTY PRODUCTS OR SERVICES DOES NOT CONSTITUTE AN ENDORSEMENT REGARDING THE SUITABILITY OF SUCH PRODUCTS OR SERVICES OR A WARRANTY, REPRESENTATION OR ENDORSEMENT OF SUCH PRODUCTS OR SERVICES, EITHER ALONE OR IN COMBINATION WITH ANY TI PRODUCT OR SERVICE.

#### 12.1.2 Device Nomenclature

##### 12.1.2.1 Device Markings



← Line 1  
← Line 2  
← Line 3

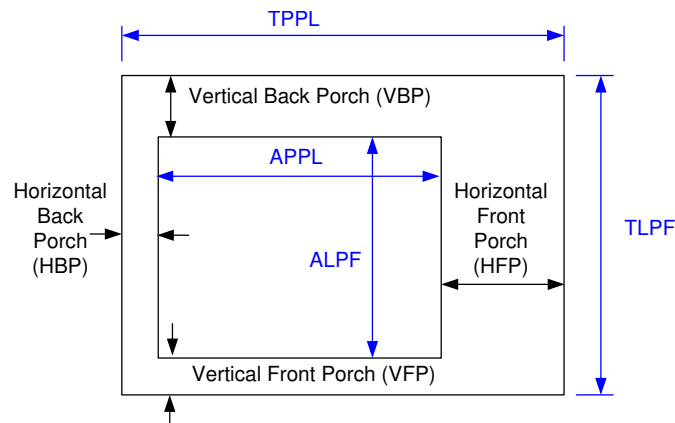
#### Marking Definitions:

Line 1:	<p>TI Part Number: Engineering Samples</p> <p>TI Part Number: Production</p>	<p>X = Engineering Samples DLPC230 = Device ID blank or A, B, C ... = Part Revision T = Temperature designator ZDQ = Package designator Q1 = Automotive qualified</p> <p>DLPC230 = Device ID blank or A, B, C ... = Part Revision S = Functional Safety T = Temperature designator ZDQ = Package designator Q1 = Automotive qualified</p>
Line 2:	Vendor Lot and Fab Information	<p>XXXXX = Fab lot number -XX = Fab sub-lot X (last X) = Assembly sub-lot The Fab is UMC12A. As such, the first character of the lot number is K</p>
Line 3:	Vendor Year and Week code	<p>YY = Year WW = Week Example, 1614 - parts built the 14<sup>th</sup> week of 2016</p>

##### 12.1.2.2 Video Timing Parameter Definitions

<b>Active Lines Per Frame (ALPF)</b>	Defines the number of lines in a frame containing displayable data: ALPF is a subset of the TLPF.
<b>Active Pixels Per Line (APPL)</b>	Defines the number of pixel clocks in a line containing displayable data: APPL is a subset of the TPPL.

<b>Horizontal Back Porch (HBP) Blanking</b>	Number of blank pixel clocks after horizontal sync but before the first active pixel. Note: HBP times are reference to the leading (active) edge of the respective sync signal.
<b>Horizontal Front Porch (HFP) Blanking</b>	Number of blank pixel clocks after the last active pixel but before Horizontal Sync.
<b>Horizontal Sync (HS)</b>	Timing reference point that defines the start of each horizontal interval (line). The absolute reference point is defined by the active edge of the HS signal. The active edge (either rising or falling edge as defined by the source) is the reference from which all horizontal blanking parameters are measured.
<b>Total Lines Per Frame (TLPF)</b>	Defines the vertical period (or frame time) in lines: TLPF = Total number of lines per frame (active and inactive).
<b>Total Pixel Per Line (TPPL)</b>	Defines the horizontal line period in pixel clocks: TPPL = Total number of pixel clocks per line (active and inactive).
<b>Vertical Sync (VS)</b>	Timing reference point that defines the start of the vertical interval (frame). The absolute reference point is defined by the active edge of the VS signal. The active edge (either rising or falling edge as defined by the source) is the reference from which all vertical blanking parameters are measured.
<b>Vertical Back Porch (VBP) Blanking</b>	Number of blank lines after vertical sync but before the first active line.
<b>Vertical Front Porch (VFP) Blanking</b>	Number of blank lines after the last active line but before vertical sync.



## 12.2 Trademarks

DLP® is a registered trademark of Texas Instruments.

is a registered trademark of Texas Instruments.

All trademarks are the property of their respective owners.

## 12.3 Electrostatic Discharge Caution



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

## 12.4 Glossary

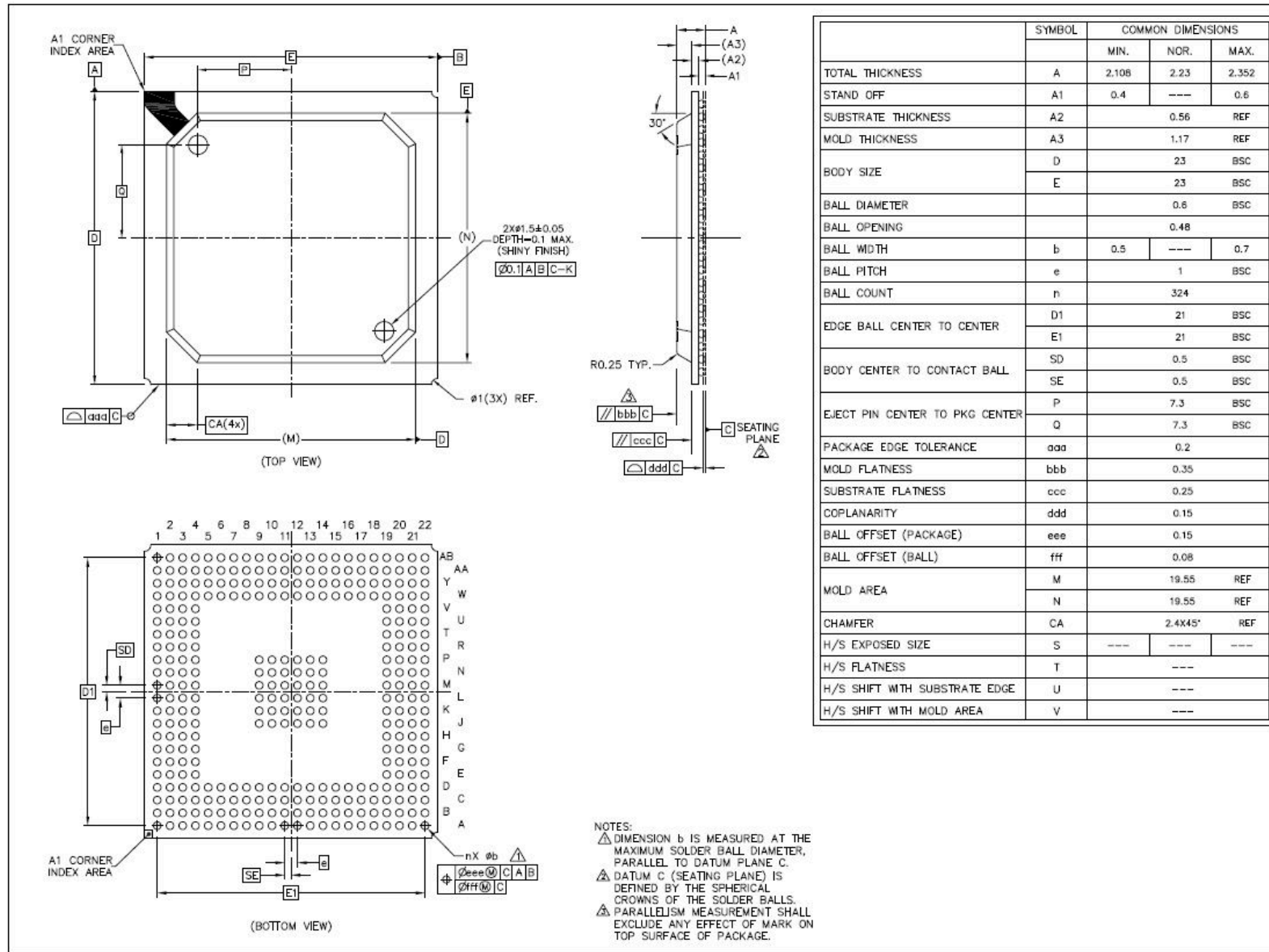
### TI Glossary

This glossary lists and explains terms, acronyms, and definitions.

## 13 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

### 13.1 DLPC230S-Q1 Mechanical Data



23-mm x 23-mm Package – Plastic Ball Grid Array



**PACKAGING INFORMATION**

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
DLPC230STZDQQ1	ACTIVE	BGA	ZDQ	324	1	TBD	Call TI	Call TI	-40 to 105		Samples
DLPC23STZDQRQ1	ACTIVE	BGA	ZDQ	324	250	TBD	Call TI	Call TI	-40 to 105		Samples

(1) The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

**LIFEBUY:** TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

**NRND:** Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSELETE:** TI has discontinued the production of the device.

(2) **RoHS:** TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

**RoHS Exempt:** TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

**Green:** TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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