DSD1794A

24-BIT, 192-kHz SAMPLING, ADVANCED SEGMENT, AUDIO STEREO DIGITAL-TO-ANALOG CONVERTER

FEATURES

- **Supports Both DSD and PCM Formats**
- **24-Bit Resolution**
- **Analog Performance:**
  - **Dynamic Range:**
    - **132 dB (9 V RMS, Mono)**
    - **129 dB (4.5 V RMS, Stereo)**
    - **127 dB (2 V RMS, Stereo)**
  - **THD+N: 0.0004%**
- **Differential Current Output: 7.8 mA p-p**
- **8× Oversampling Digital Filter:**
  - **Stop-Band Attenuation: –130 dB**
  - **Pass-Band Ripple: ±0.00001 dB**
- **Sampling Frequency: 10 kHz to 200 kHz**
- **System Clock: 128, 192, 256, 384, 512, or 768 f_S With Autodetect**
- **Accepts 16-, 20-, and 24-Bit Audio Data**
- **PCM Data Formats: Standard, I^{2}S, and Left-Justified**
- **Optional Interface to External Digital Filter or DSP Available**
- **I^{2}C-Compatible Serial Port**
- **User-Programmable Mode Controls:**
  - **Digital Attenuation: 0 dB to –120 dB, 0.5 dB/Step**
  - **Digital De-Emphasis**
  - **Digital Filter Rolloff: Sharp or Slow**
  - **Soft Mute**
- **Dual-Supply Operation:**
  - **5-V Analog, 3.3-V Digital**
- **5-V Tolerant Digital Inputs**
- **Small 28-Lead SSOP Package**

APPLICATIONS

- **A/V Receivers**
- **SACD Player**
- **DVD Players**
- **HDTV Receivers**
- **Car Audio Systems**
- **Digital Multitrack Recorders**
- **Other Applications Requiring 24-Bit Audio**

DESCRIPTION

The DSD1794A is a monolithic CMOS integrated circuit that includes stereo digital-to-analog converters and support circuitry in a small 28-lead SSOP package. The data converters use TI's advanced-segment DAC architecture to achieve excellent dynamic performance and improved tolerance to clock jitter. The DSD1794A provides balanced current outputs, allowing the user to optimize analog performance externally. The DSD1794A accepts the PCM and DSD audio data formats, providing easy interfacing to audio DSP and decoder chips. The DSD1794A also interfaces with external digital filter devices (DF1704, DF1706, PMD200). Sampling rates up to 200 kHz are supported. A full set of user-programmable functions is accessible through an I^{2}C-compatible serial port.

This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.

DSD1794A

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ORDERING INFORMATION

PRODUCT	PACKAGE	PACKAGE CODE	OPERATION TEMPERATURE RANGE	PACKAGE MARKING	ORDERING NUMBER	TRANSPORT MEDIA
DSD1794ADB	28-lead SSOP	28DB	−25°C to 85°C	DSD1794A	DSD1794ADB	Tube
					DSD1794ADBR	Tape and reel

ABSOLUTE MAXIMUM RATINGS

over operating free-air temperature range unless otherwise noted[1]

		DSD1794A
Supply voltage	$V_{CC}1$, $V_{CC}2L$, $V_{CC}2R$	−0.3 V to 6.5 V
	V_{DD}	−0.3 V to 4 V
Supply voltage differences: $V_{CC}1$, $V_{CC}2L$, $V_{CC}2R$		±0.1 V
Ground voltage differences: AGND1, AGND2, AGND3L, AGND3R, DGND		±0.1 V
Digital input voltage	PLRCK, PDATA, PBCK, SCK, RST, SCL, SDA[2], ADR0, ADR1, DSDL[2], DSDR[2], DBCK	−0.3 V to 6.5 V
	DSDL[3], DSDR[3], SDA[3]	−0.3 V to (V_{DD} + 0.3 V) < 4 V
Analog input voltage		−0.3 V to (V_{CC} + 0.3 V) < 6.5 V
Input current (any pins except supplies)		±10 mA
Ambient temperature under bias		−40°C to 125°C
Storage temperature		−55°C to 150°C
Junction temperature		150°C
Lead temperature (soldering)		260°C, 5 s
Package temperature (IR reflow, peak)		250°C

[1] Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

[2] Input mode

[3] Output mode

ELECTRICAL CHARACTERISTICS

all specifications at T_A = 25°C, $V_{CC}1$ = $V_{CC}2L$ = $V_{CC}2R$ = 5 V, V_{DD} = 3.3. V, f_S = 44.1 kHz, system clock = 256 f_S, and 24-bit data, unless otherwise noted

PARAMETER		TEST CONDITIONS	DSD1794ADB			UNIT
			MIN	TYP	MAX	
RESOLUTION				24		Bits
DATA FORMAT (PCM Mode)						
	Audio data interface format			Standard, I^2S, left justified		
	Audio data bit length			16-, 20-, 24-bit selectable		
	Audio data format			MSB first, 2s complement		
f_S	Sampling frequency		10		200	kHz
	System clock frequency			128, 192, 256, 384, 512, 768 f_S		
DATA FORMAT (DSD Mode)						
	Audio data interface format			DSD (direct stream digital)		
	Audio data bit length			1 bit		
f_S	Sampling frequency			2.8224		MHz
	System clock frequency		2.8224		11.2896	MHz

ELECTRICAL CHARACTERISTICS (Continued)

all specifications at $T_A = 25°C$, $V_{CC}1 = V_{CC}2L = V_{CC}2R = 5$ V, $V_{DD} = 3.3$. V, $f_S = 44.1$ kHz, system clock = 256 f_S, and 24-bit data, unless otherwise noted

PARAMETER		TEST CONDITIONS	DSD1794ADB			UNIT
			MIN	TYP	MAX	
DIGITAL INPUT/OUTPUT						
	Logic family			TTL compatible		
V_{IH}	Input logic level		2			VDC
V_{IL}					0.8	
I_{IH}	Input logic current	$V_{IN} = V_{DD}$			10	μA
I_{IL}		$V_{IN} = 0$ V			−10	
V_{OH}	Output logic level	$I_{OH} = −2$ mA	2.4			VDC
V_{OL}		$I_{OL} = 2$ mA			0.4	
DYNAMIC PERFORMANCE (PCM MODE, 2-V RMS OUTPUT) [1][2]						
THD+N at $V_{OUT} = 0$ dB		$f_S = 44.1$ kHz		0.0004%	0.0008%	
		$f_S = 96$ kHz		0.0008%		
		$f_S = 192$ kHz		0.0015%		
Dynamic range		EIAJ, A-weighted, $f_S = 44.1$ kHz	123	127		dB
		EIAJ, A-weighted, $f_S = 96$ kHz		127		
		EIAJ, A-weighted, $f_S = 192$ kHz		127		
Signal-to-noise ratio		EIAJ, A-weighted, $f_S = 44.1$ kHz	123	127		dB
		EIAJ, A-weighted, $f_S = 96$ kHz		127		
		EIAJ, A-weighted, $f_S = 192$ kHz		127		
Channel separation		$f_S = 44.1$ kHz	120	123		dB
		$f_S = 96$ kHz		122		
		$f_S = 192$ kHz		120		
Level linearity error		$V_{OUT} = −120$ dB		±1		dB
DYNAMIC PERFORMANCE (PCM Mode, 4.5-V RMS Output) [1][3]						
THD+N at $V_{OUT} = 0$ dB		$f_S = 44.1$ kHz		0.0004%		
		$f_S = 96$ kHz		0.0008%		
		$f_S = 192$ kHz		0.0015%		
Dynamic range		EIAJ, A-weighted, $f_S = 44.1$ kHz		129		dB
		EIAJ, A-weighted, $f_S = 96$ kHz		129		
		EIAJ, A-weighted, $f_S = 192$ kHz		129		
Signal-to-noise ratio		EIAJ, A-weighted, $f_S = 44.1$ kHz		129		dB
		EIAJ, A-weighted, $f_S = 96$ kHz		129		
		EIAJ, A-weighted, $f_S = 192$ kHz		129		
Channel separation		$f_S = 44.1$ kHz		124		dB
		$f_S = 96$ kHz		123		
		$f_S = 192$ kHz		121		

[1] Filter condition:
THD+N: 20-Hz HPF, 20-kHz apogee LPF
Dynamic range: 20-Hz HPF, 20-kHz AES17 LPF, A-weighted
Signal-to-noise ratio: 20-Hz HPF, 20-kHz AES17 LPF, A-weighted
Channel separation: 20-Hz HPF, 20-kHz AES17 LPF
Analog performance specifications are measured using the System Two™ Cascade audio measurement system by Audio Precision™ in the averaging mode.

[2] Dynamic performance and dc accuracy are specified at the output of the postamplifier as shown in Figure 33.

[3] Dynamic performance and dc accuracy are specified at the output of the postamplifier as shown in Figure 34.

ELECTRICAL CHARACTERISTICS (Continued)

all specifications at T_A = 25°C, $V_{CC}1$ = $V_{CC}2L$ = $V_{CC}2R$ = 5 V, V_{DD} = 3.3. V, f_S = 44.1 kHz, system clock = 256 f_S, and 24-bit data, unless otherwise noted

PARAMETER	TEST CONDITIONS	DSD1794ADB			UNIT
		MIN	TYP	MAX	
DYNAMIC PERFORMANCE (MONO MODE) [1][2]					
THD+N at V_{OUT} = 0 dB	f_S = 44.1 kHz		0.0004%		
	f_S = 96 kHz		0.0008%		
	f_S = 192 kHz		0.0015%		
Dynamic range	EIAJ, A-weighted, f_S = 44.1 kHz		132		dB
	EIAJ, A-weighted, f_S = 96 kHz		132		
	EIAJ, A-weighted, f_S = 192 kHz		132		
Signal-to-noise ratio	EIAJ, A-weighted, f_S = 44.1 kHz		132		dB
	EIAJ, A-weighted, f_S = 96 kHz		132		
	EIAJ, A-weighted, f_S = 192 kHz		132		
DSD MODE DYNAMIC PERFORMANCE [1][3] **(44.1 kHz, 64 f_S)**					
THD+N at FS	4.5 V rms		0.0005%		
Dynamic range	–60 dB, EIAJ, A-weighted		128		dB
Signal-to-noise ratio	EIAJ, A-weighted		128		dB
ANALOG OUTPUT					
Gain error		–6	±2	6	% of FSR
Gain mismatch, channel-to-channel		–3	±0.5	3	% of FSR
Bipolar zero error	At BPZ	–2	±0.5	2	% of FSR
Output current	Full scale (0 dB)		7.8		mA p-p
Center current	At BPZ		–6.2		mA
DIGITAL FILTER PERFORMANCE					
De-emphasis error			±0.004		dB
FILTER CHARACTERISTICS-1: SHARP ROLLOFF					
Pass band	±0.00001 dB			0.454 f_S	
	–3 dB			0.49 f_S	
Stop band		0.546 f_S			
Pass-band ripple				±0.00001	dB
Stop-band attenuation	Stop band = 0.546 f_S	–130			dB
Delay time			55/f_S		s
FILTER CHARACTERISTICS-2: SLOW ROLLOFF					
Pass band	±0.04 dB			0.254 f_S	
	–3 dB			0.46 f_S	
Stop band		0.732 f_S			
Pass-band ripple				±0.001	dB
Stop-band attenuation	Stop band = 0.732 f_S	–100			dB
Delay time			18/f_S		s

[1] Filter condition:
THD+N: 20-Hz HPF, 20-kHz apogee LPF
Dynamic range: 20-Hz HPF, 20-kHz AES17 LPF, A-weighted
Signal-to-noise ratio: 20-Hz HPF, 20-kHz AES17 LPF, A-weighted
Channel separation: 20-Hz HPF, 20-kHz AES17 LPF
Analog performance specifications are measured using the System Two Cascade audio measurement system by Audio Precision in the averaging mode.

[2] Dynamic performance and dc accuracy are specified at the output of the postamplifier as shown in Figure 34.

[3] Dynamic performance and dc accuracy are specified at the output of the postamplifier as shown in Figure 35.
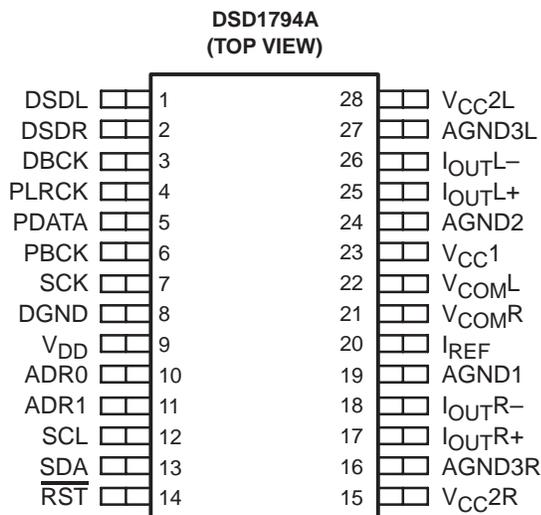
ELECTRICAL CHARACTERISTICS (Continued)

all specifications at T_A = 25°C, $V_{CC}1$ = $V_{CC}2L$ = $V_{CC}2R$ = 5 V, V_{DD} = 3.3. V f_S = 44.1 kHz, system clock = 256 f_S, and 24-bit data, unless otherwise noted

PARAMETER		TEST CONDITIONS	DSD1794ADB			UNIT
			MIN	TYP	MAX	
POWER SUPPLY REQUIREMENTS						
V_{DD}	Voltage range		3	3.3	3.6	VDC
$V_{CC}1$						
$V_{CC}2L$	Voltage range		4.75	5	5.25	VDC
$V_{CC}2R$						
I_{DD}	Supply current [1]	f_S = 44.1 kHz		12	15	mA
		f_S = 96 kHz		23		
		f_S = 192 kHz		45		
I_{CC}		f_S = 44.1 kHz		33	40	mA
		f_S = 96 kHz		35		
		f_S = 192 kHz		37		
	Power dissipation [1]	f_S = 44.1 kHz		205	250	mW
		f_S = 96 kHz		250		
		f_S = 192 kHz		335		
TEMPERATURE RANGE						
	Operation temperature		−25		85	°C
θ_{JA}	Thermal resistance	28-pin SSOP		100		°C/W

[1] Input is BPZ data.

PIN ASSIGNMENTS

DSD1794A
(TOP VIEW)

DSDL	1	28	$V_{CC}2L$
DSDR	2	27	AGND3L
DBCK	3	26	$I_{OUT}L-$
PLRCK	4	25	$I_{OUT}L+$
PDATA	5	24	AGND2
PBCK	6	23	$V_{CC}1$
SCK	7	22	$V_{COM}L$
DGND	8	21	$V_{COM}R$
V_{DD}	9	20	I_{REF}
ADR0	10	19	AGND1
ADR1	11	18	$I_{OUT}R-$
SCL	12	17	$I_{OUT}R+$
SDA	13	16	AGND3R
RST	14	15	$V_{CC}2R$

Terminal Functions

TERMINAL		I/O	DESCRIPTIONS
NAME	PIN		
ADR0	10	I	I^2C address 0 [1]
ADR1	11	I	I^2C address 1 [1]
AGND1	19	–	Analog ground (internal bias)
AGND2	24	–	Analog ground (internal bias)
AGND3L	27	–	Analog ground (L-channel DACFF)
AGND3R	16	–	Analog ground (R-channel DACFF)
DBCK	3	I	Bit clock input for DSD modes [1]
DGND	8	–	Digital ground
DSDL	1	I/O	L-channel audio data input when in DSD and external DF modes PCM-mode zero flag for L-channel when in zero-flag output mode [2]
DSDR	2	I/O	R-channel audio data input when in DSD and external DF modes [2] PCM-mode zero flag for R-channel when in zero-flag output mode
$I_{OUT}L+$	25	O	L-channel analog current output +
$I_{OUT}L-$	26	O	L-channel analog current output –
$I_{OUT}R+$	17	O	R-channel analog current output +
$I_{OUT}R-$	18	O	R-channel analog current output –
I_{REF}	20	–	Output current reference bias pin
PBCK	6	I	Bit clock input. Connected to GND in DSD mode [1]
PDATA	5	I	Serial audio data input for PCM-format operation [1]
PLRCK	4	I	Left and right clock (f_S) input for PCM-format operation. WDCK clock input for external DF mode. Connected to GND for DSD mode [1]
$\overline{RST}$	14	I	Reset [1]
SCL	12	I	I^2C clock [1]
SCK	7	I	System clock input [1]
SDA	13	I/O	I^2C data [3]
$V_{CC}1$	23	–	Analog power supply, 5 V
$V_{CC}2L$	28	–	Analog power supply (L-channel DACFF), 5 V
$V_{CC}2R$	15	–	Analog power supply (R-channel DACFF), 5 V
$V_{COM}L$	22	–	L-channel internal bias decoupling pin
$V_{COM}R$	21	–	R-channel internal bias decoupling pin
V_{DD}	9	–	Digital power supply, 3.3 V

[1] Schmitt-trigger input, 5-V tolerant
[2] Schmitt-trigger input and output. 5-V tolerant input, and CMOS output
[3] Schmitt-trigger 5-V tolerant input and open-drain/3-state output

FUNCTIONAL BLOCK DIAGRAM

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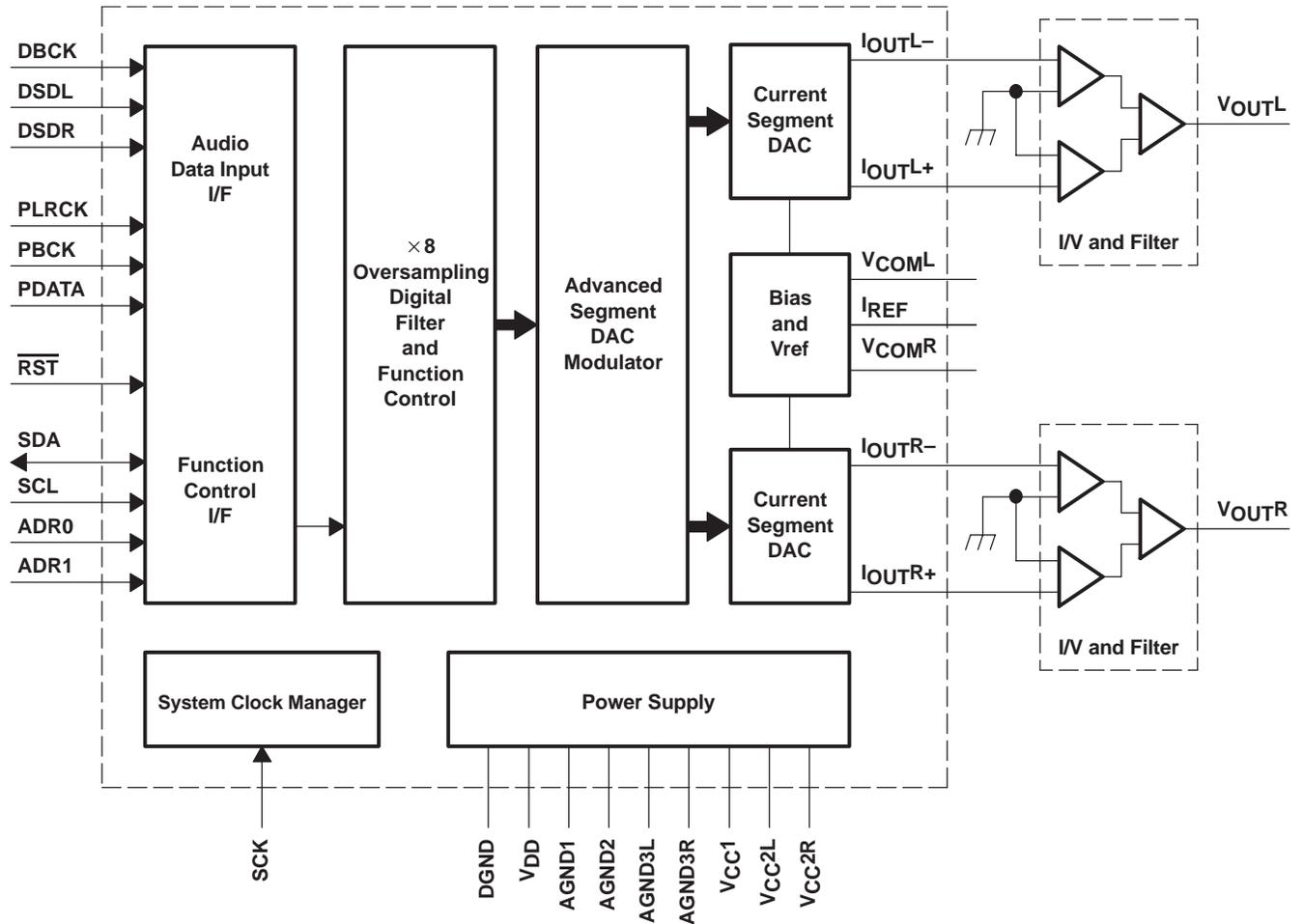
TYPICAL PERFORMANCE CURVES

DIGITAL FILTER

Digital Filter Response

AMPLITUDE vs FREQUENCY
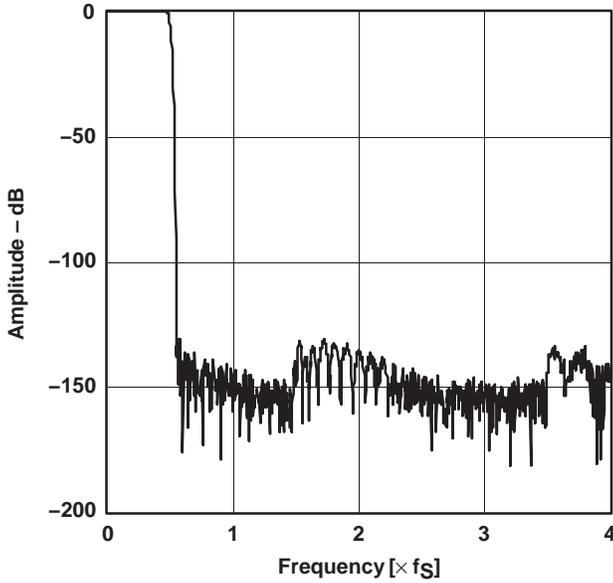
Figure 1. Frequency Response, Sharp Rolloff
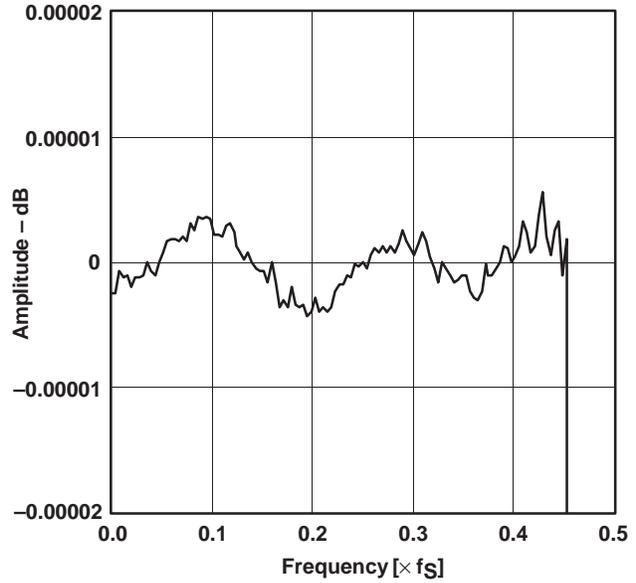
AMPLITUDE vs FREQUENCY

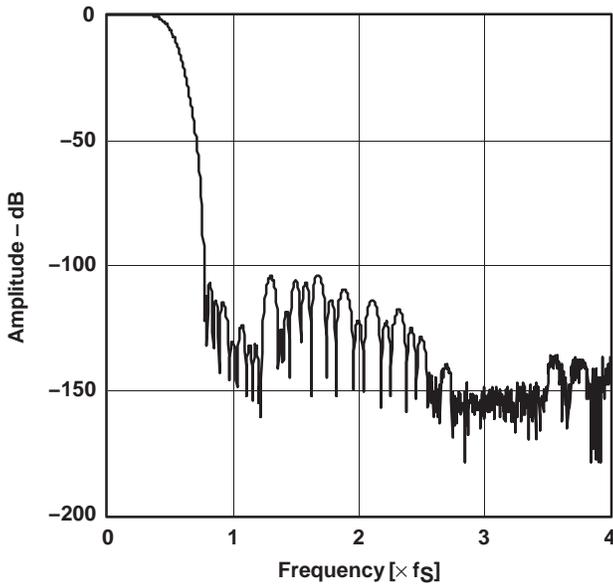Figure 2. Pass-Band Ripple, Sharp Rolloff

AMPLITUDE vs FREQUENCY

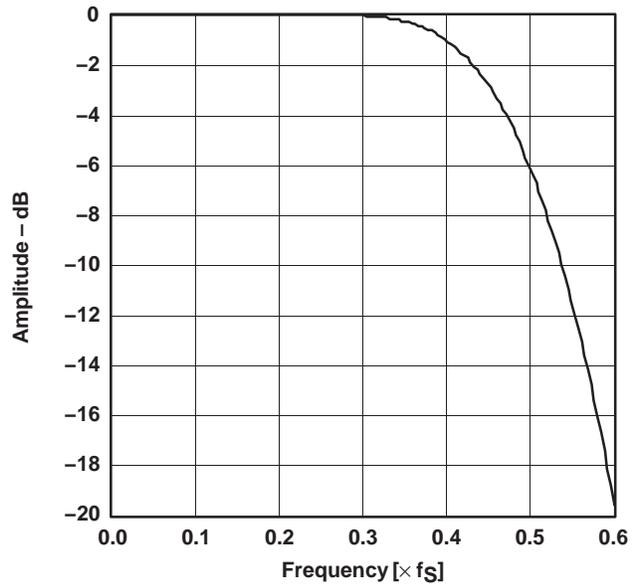Figure 3. Frequency Response, Slow Rolloff

AMPLITUDE vs FREQUENCY

Figure 4. Transition Characteristics, Slow Rolloff
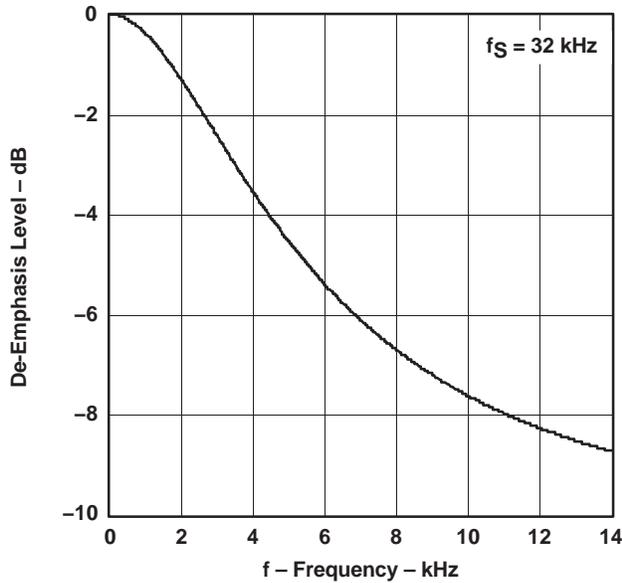
De-Emphasis Filter

DE-EMPHASIS LEVEL
vs
FREQUENCY

Figure 5

DE-EMPHASIS ERROR
vs
FREQUENCY
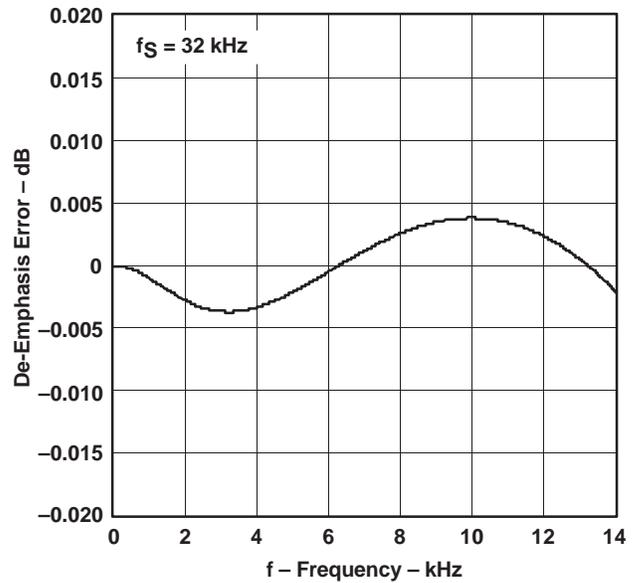
Figure 6

DE-EMPHASIS LEVEL
vs
FREQUENCY
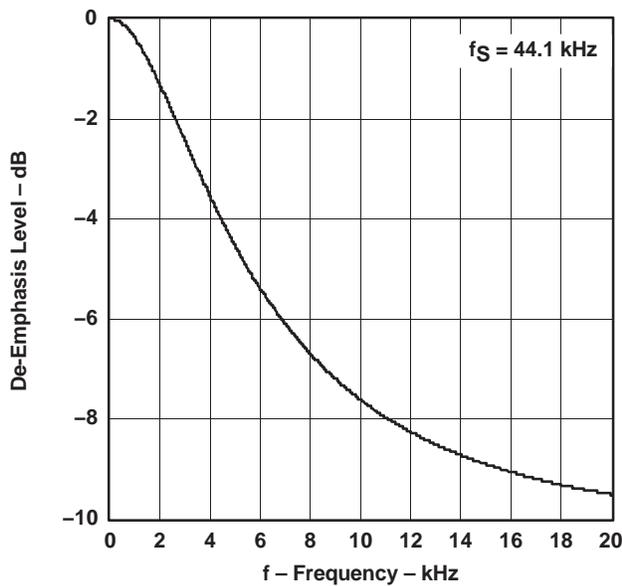
Figure 7

DE-EMPHASIS ERROR
vs
FREQUENCY
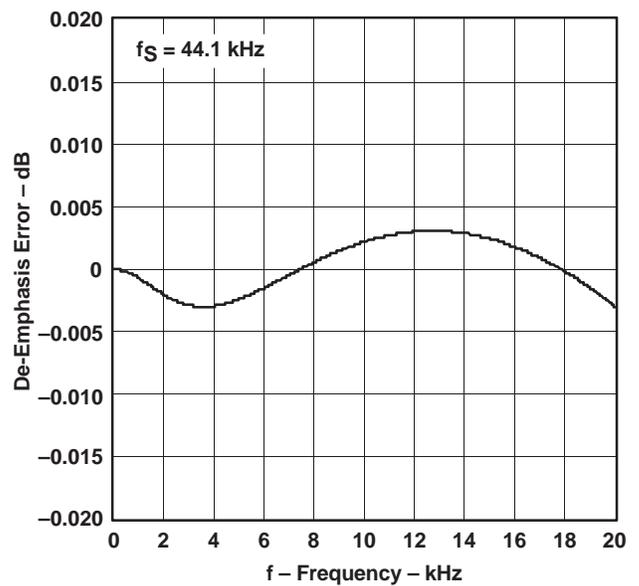
Figure 8

De-Emphasis Filter (Continued)

**DE-EMPHASIS LEVEL
vs
FREQUENCY**

f_S = 48 kHz

Figure 9

**DE-EMPHASIS ERROR
vs
FREQUENCY**

f_S = 48 kHz
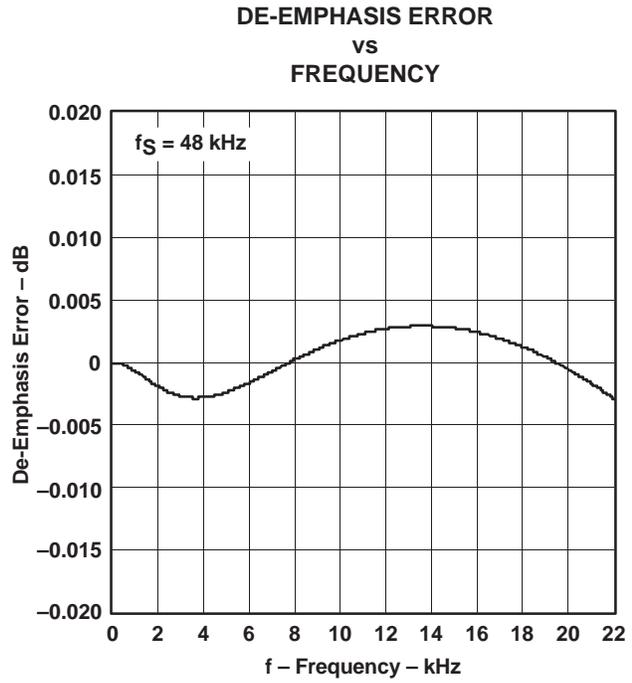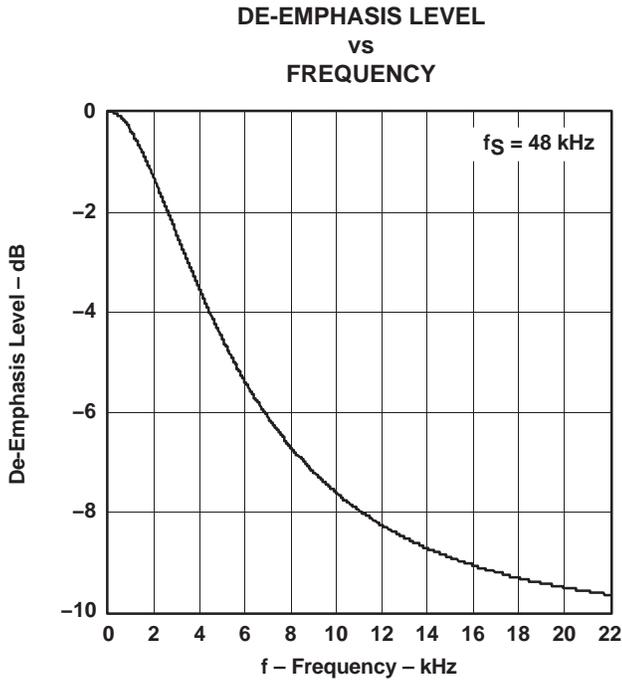
Figure 10

ANALOG DYNAMIC PERFORMANCE

Supply Voltage Characteristics

TOTAL HARMONIC DISTORTION + NOISE
vs
SUPPLY VOLTAGE

Figure 11

DYNAMIC RANGE
vs
SUPPLY VOLTAGE

Figure 12

SIGNAL-to-NOISE RATIO
vs
SUPPLY VOLTAGE

Figure 13

CHANNEL SEPARATION
vs
SUPPLY VOLTAGE
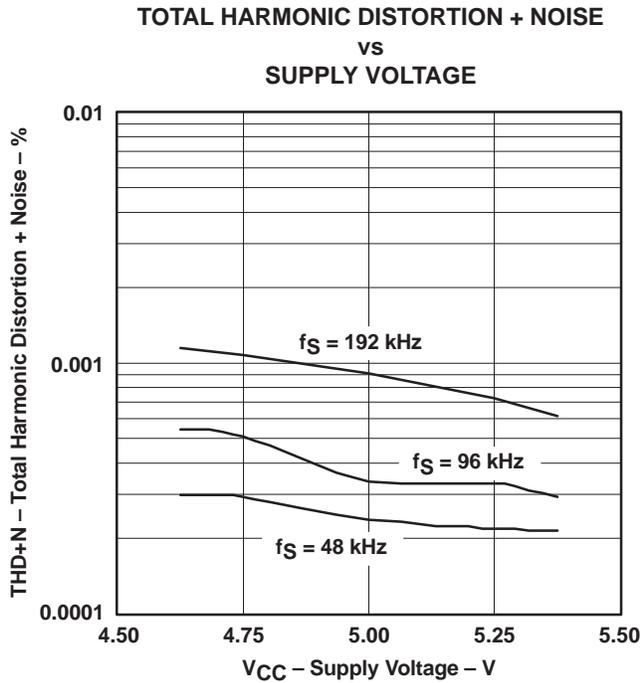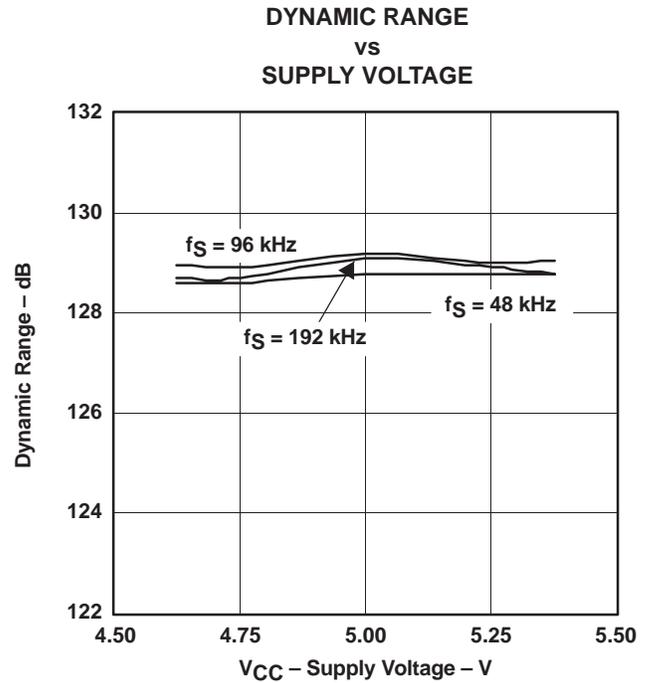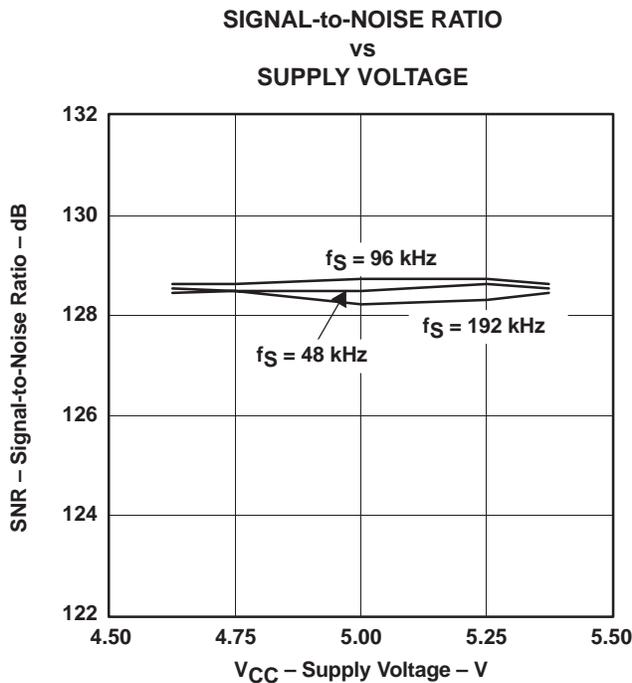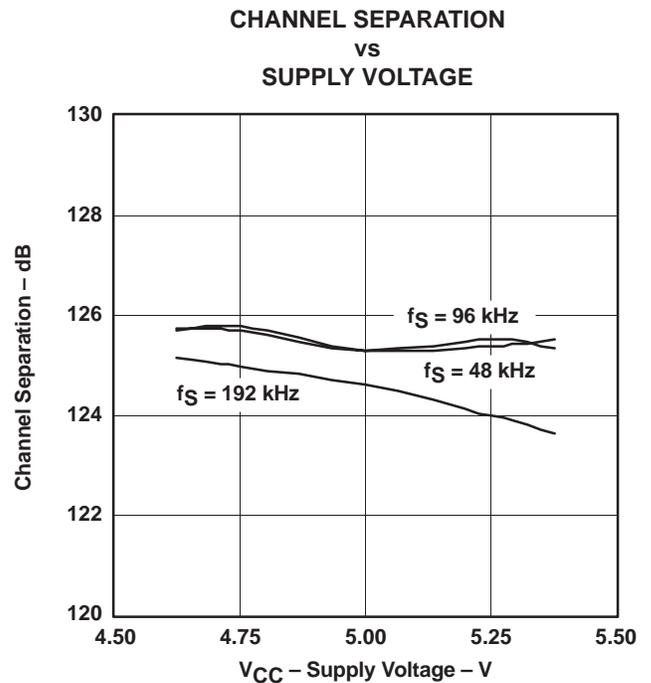
Figure 14

NOTE: PCM mode, T_A = 25°C, V_{DD} = 3.3 V, measurement circuit is Figure 34 (V_{OUT} = 4.5 V rms).

Temperature Characteristics

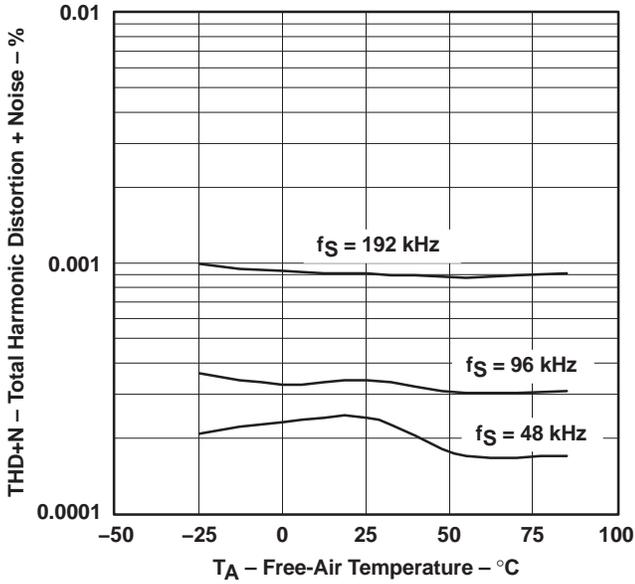TOTAL HARMONIC DISTORTION + NOISE
vs
FREE-AIR TEMPERATURE

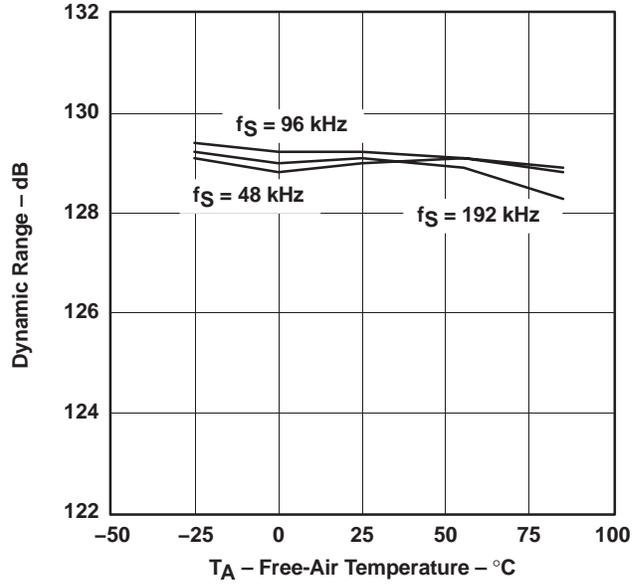**Figure 15**

DYNAMIC RANGE
vs
FREE-AIR TEMPERATURE

Figure 16

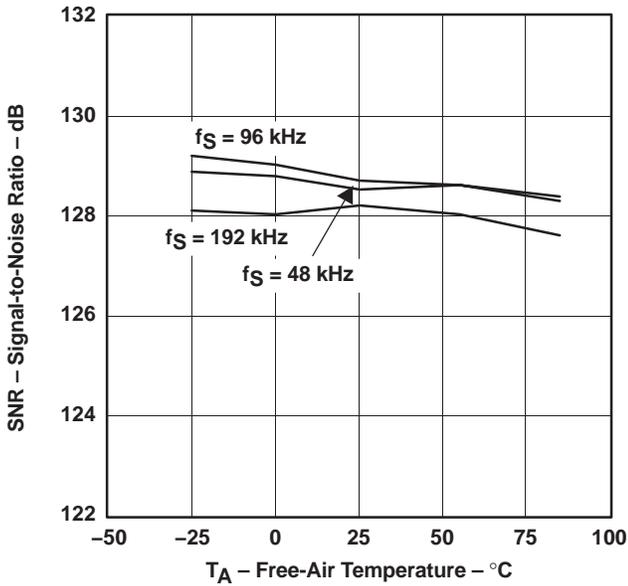SIGNAL-to-NOISE RATIO
vs
FREE-AIR TEMPERATURE

Figure 17

CHANNEL SEPARATION
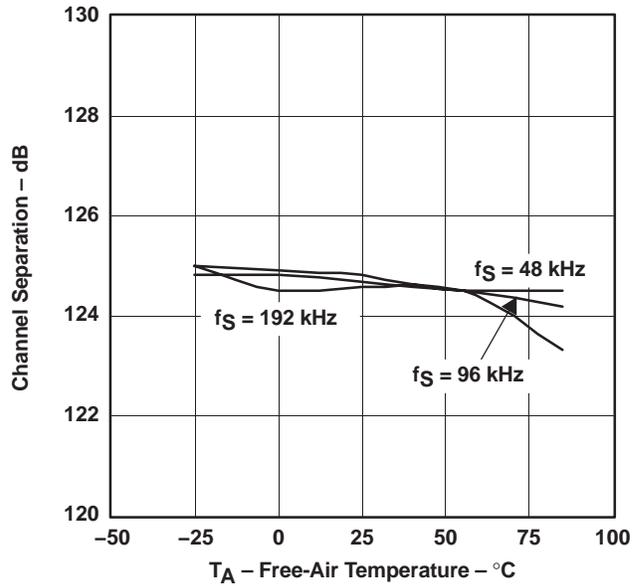vs
FREE-AIR TEMPERATURE

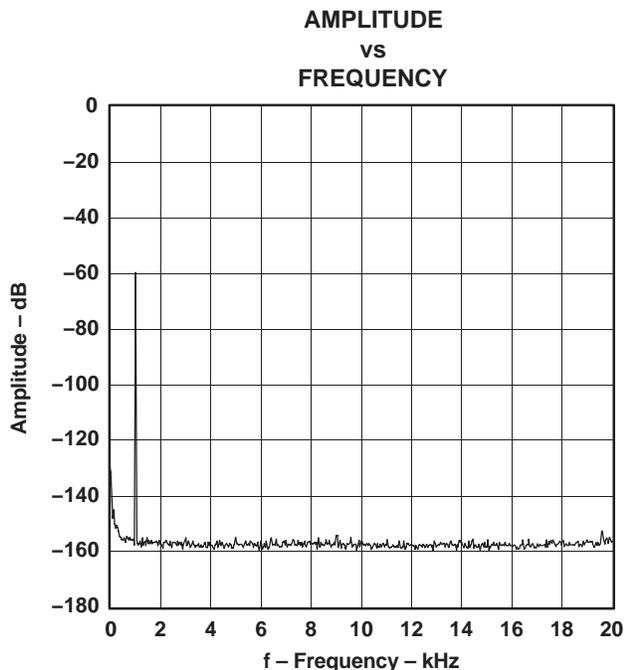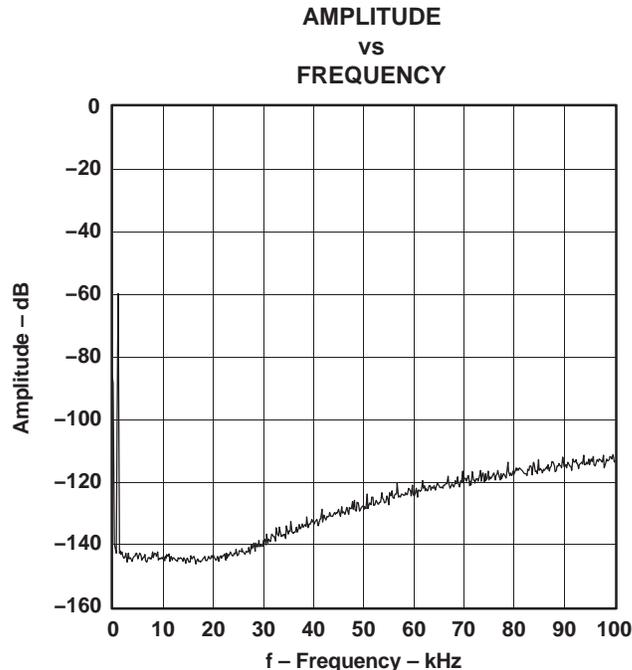**Figure 18**

NOTE: PCM mode, V_{DD} = 3.3 V, V_{CC} = 5 V, measurement circuit is Figure 34 (V_{OUT} = 4.5 V rms).

**AMPLITUDE
vs
FREQUECY**

NOTE: PCM mode, f_S = 48 kHz, 32768 point 8 average, T_A = 25°C, V_{DD} = 3.3 V, V_{CC} = 5 V, measurement circuit is Figure 34.

Figure 19. –60-db Output Spectrum, BW = 20 kHz

**AMPLITUDE
vs
FREQUENCY**
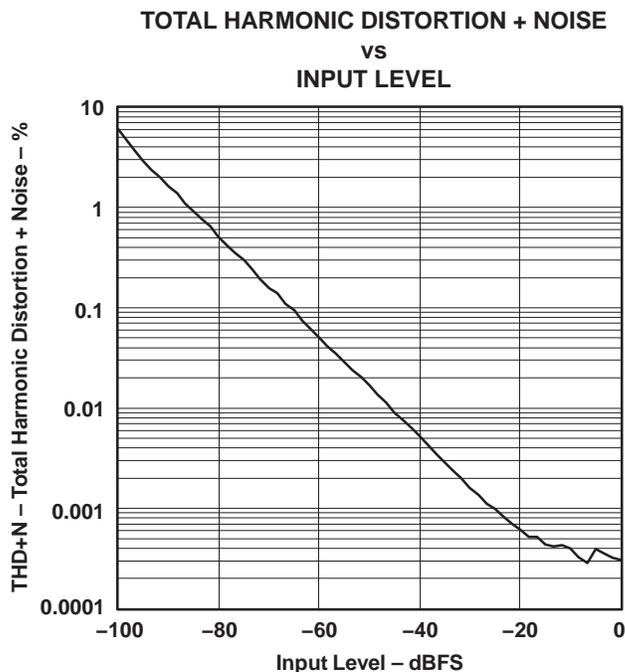

NOTE: PCM mode, f_S = 48 kHz, 32768 point 8 average, T_A = 25°C, V_{DD} = 3.3 V, V_{CC} = 5 V, measurement circuit is Figure 34.

Figure 20. –60-db Output Spectrum, BW = 100 kHz

**TOTAL HARMONIC DISTORTION + NOISE
vs
INPUT LEVEL**

NOTE: PCM mode, f_S = 48 kHz, T_A = 25°C, V_{DD} = 3.3 V, V_{CC} = 5 V, measurement circuit is Figure 34.

Figure 21. THD+N vs Input Level, PCM Mode

**AMPLITUDE
vs
FREQUENCY**
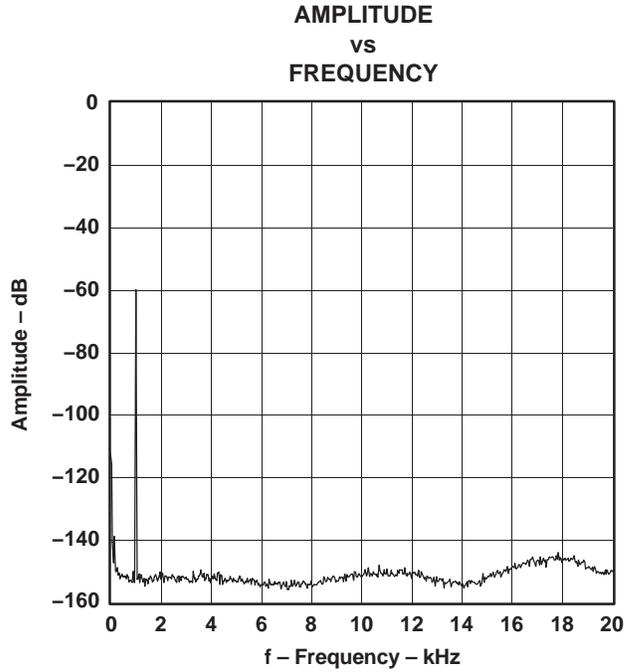
Figure 22. –60-dB Output Spectrum, DSD Mode

**AMPLITUDE
vs
FREQUENCY**

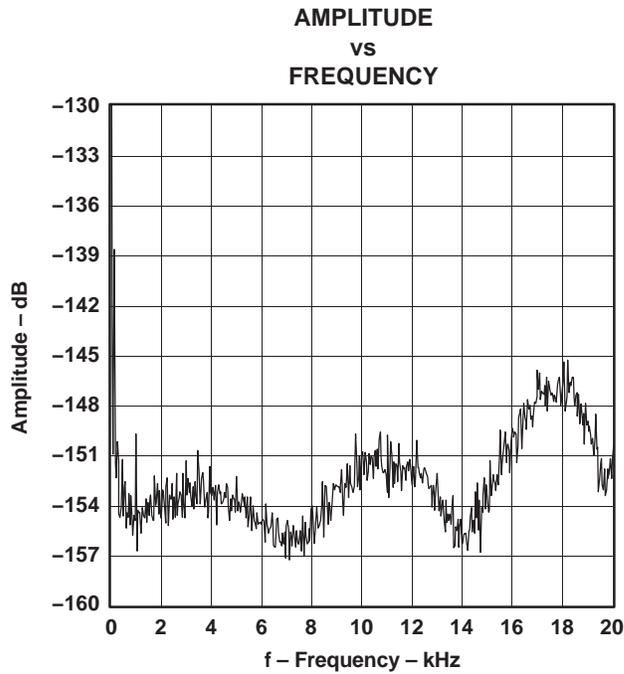**Figure 23. –150-dB Output Spectrum, DSD Mono Mode**

NOTE: DSD mode (FIR-4), 32768 point 8 average, T_A = 25°C, V_{DD} = 3.3 V, V_{CC} = 5 V, measurement circuit is Figure 36.

SYSTEM CLOCK AND RESET FUNCTIONS

System Clock Input

The DSD1794A requires a system clock for operating the digital interpolation filters and advanced segment DAC modulators. The system clock is applied at the SCK input (pin 7). The DSD1794A has a system clock detection circuit that automatically senses the frequency at which the system clock is operating. Table 1 shows examples of system clock frequencies for common audio sampling rates. If the oversampling rate of the delta-sigma modulator is selected as 128 f_S, the system clock frequency is required to be over 256 f_S.

Figure 24 shows the timing requirements for the system clock input. For optimal performance, it is important to use a clock source with low phase jitter and noise. One of the Texas Instruments PLL1700 family of multiclock generators is an excellent choice for providing the DSD1794A system clock.

Table 1. System Clock Rates for Common Audio Sampling Frequencies

SAMPLING FREQUENCY	SYSTEM CLOCK FREQUENCY (f_{SCK}) (MHz)					
	128 f_S	192 f_S	256 f_S	384 f_S	512 f_S	768 f_S
32 kHz	4.096[1]	6.144[1]	8.192	12.288	16.384	24.576
44.1 kHz	5.6488[1]	8.4672	11.2896	16.9344	22.5792	33.8688
48 kHz	6.144[1]	9.216	12.288	18.432	24.576	36.864
96 kHz	12.288	18.432	24.576	36.864	49.152[1]	73.728[1]
192 kHz	24.576	36.864	49.152[1]	73.728[1]	[2]	[2]

[1] This system clock rate is not supported in I2C fast mode.
[2] This system clock rate is not supported for the given sampling frequency.
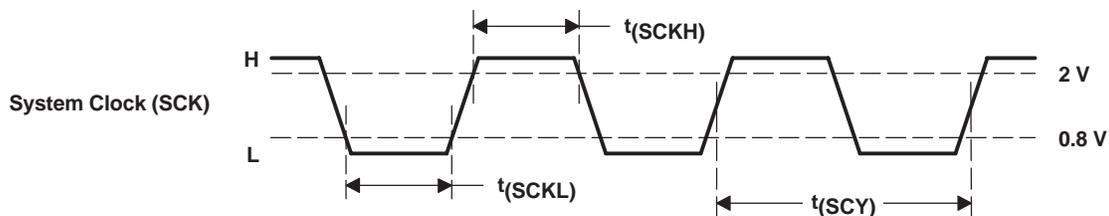
PARAMETERS		MIN	MAX	UNITS
$t_{(SCY)}$	System clock pulse cycle time	13		ns
$t_{(SCKH)}$	System clock pulse duration, HIGH	$0.4t_{(SCY)}$		ns
$t_{(SCKL)}$	System clock pulse duration, LOW	$0.4t_{(SCY)}$		ns

Figure 24. System Clock Input Timing

Power-On and External Reset Functions

The DSD1794A includes a power-on reset function. Figure 25 shows the operation of this function. With $V_{DD} > 2$ V, the power-on reset function is enabled. The initialization sequence requires 1024 system clocks from the time $V_{DD} > 2$ V. After the initialization period, the DSD1794A is set to its default reset state, as described in the *MODE CONTROL REGISTERS* section of this data sheet.

The DSD1794A also includes an external reset capability using the $\overline{RST}$ input (pin 14). This allows an external controller or master reset circuit to force the DSD1794A to initialize to its default reset state.

Figure 26 shows the external reset operation and timing. The $\overline{RST}$ pin is set to logic 0 for a minimum of 20 ns. The $\overline{RST}$ pin is then set to a logic 1 state, thus starting the initialization sequence, which requires 1024 system clock periods. The external reset is especially useful in applications where there is a delay between the DSD1794A power up and system clock activation.
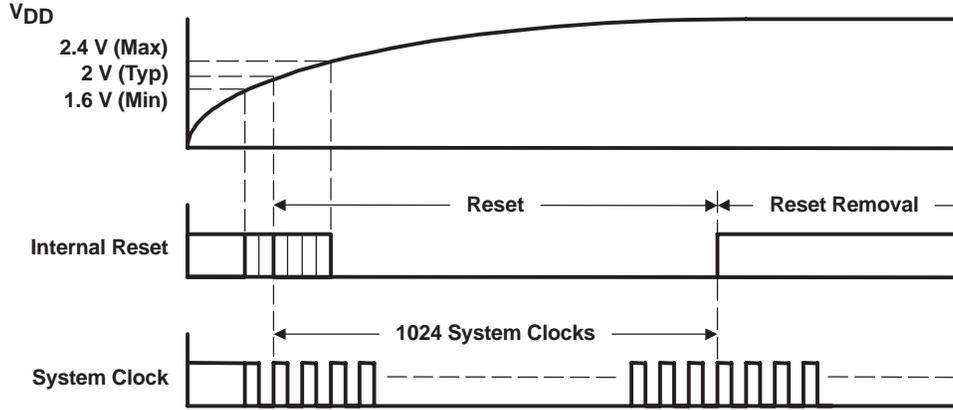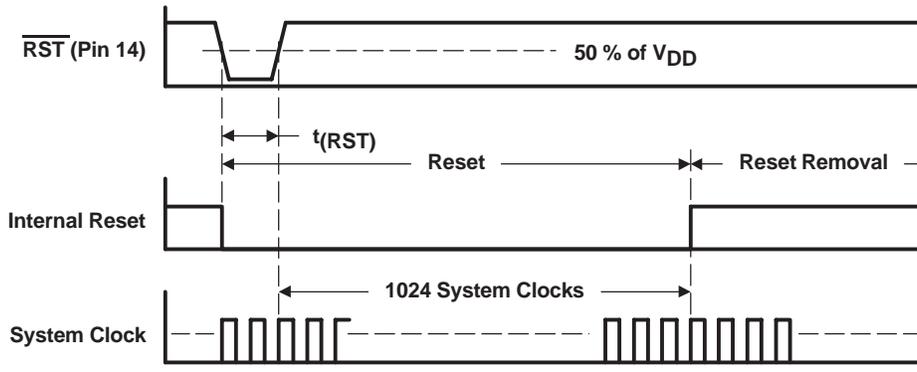
Figure 25. Power-On Reset Timing

PARAMETERS		MIN	MAX	UNITS
t(RST)	Reset pulse duration, LOW	20		ns

Figure 26. External Reset Timing

AUDIO DATA INTERFACE

Audio Serial Interface

The audio interface port is a 3-wire serial port. It includes PLRCK (pin 4), PBCK (pin 6), and PDATA (pin 5). PBCK is the serial audio bit clock, and it is used to clock the serial data present on PDATA into the serial shift register of the audio interface. Serial data is clocked into the DSD1794A on the rising edge of PBCK. PLRCK is the serial audio left/right word clock.

The DSD1794A requires the synchronization of PLRCK and the system clock, but does not need a specific phase relation between PLRCK and the system clock.

If the relationship between PLRCK and the system clock changes more than ±6 PBCK, internal operation is initialized within $1/f_S$ and analog outputs are forced to the bipolar zero level until resynchronization between PLRCK and the system clock is completed.

PCM Audio Data Formats and Timing

The DSD1794A supports industry-standard audio data formats, including standard right-justified, I^{2}S, and left-justified. The data formats are shown in Figure 28. Data formats are selected using the format bits, FMT[2:0], in control register 18. The default data format is 24-bit I^{2}S. All formats require binary twos-complement, MSB-first audio data. Figure 27 shows a detailed timing diagram for the serial audio interface.
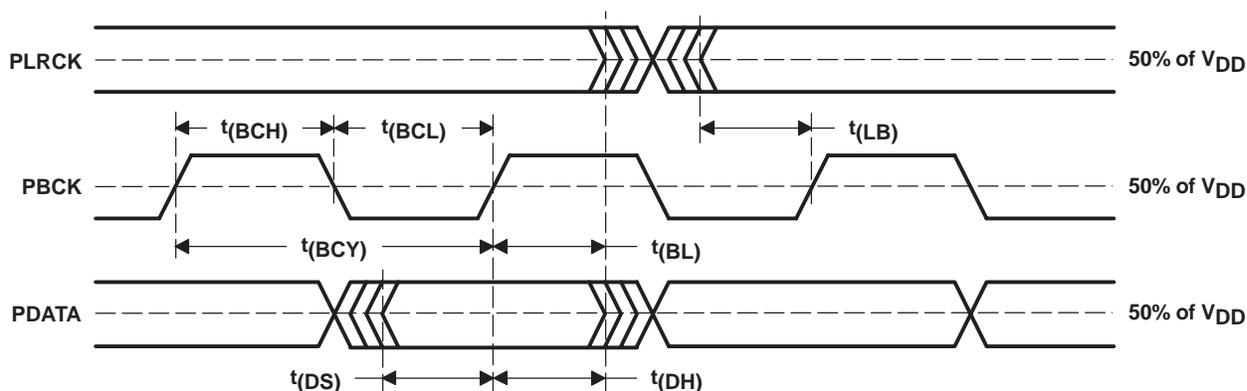
	PARAMETERS	MIN	MAX	UNITS
t(BCY)	PBCK pulse cycle time	70		ns
t(BCL)	PBCK pulse duration, LOW	30		ns
t(BCH)	PBCK pulse duration, HIGH	30		ns
t(BL)	PBCK rising edge to PLRCK edge	10		ns
t(LB)	PLRCK edge to PBCK rising edge	10		ns
t(DS)	PDATA Setup time	10		ns
t(DH)	PDATA hold time	10		ns
—	PLRCK clock data	50% ± 2 bit clocks		

Figure 27. Timing of Audio Interface

(1) Standard Data Format (Right Justified); L-Channel = HIGH, R-Channel = LOW
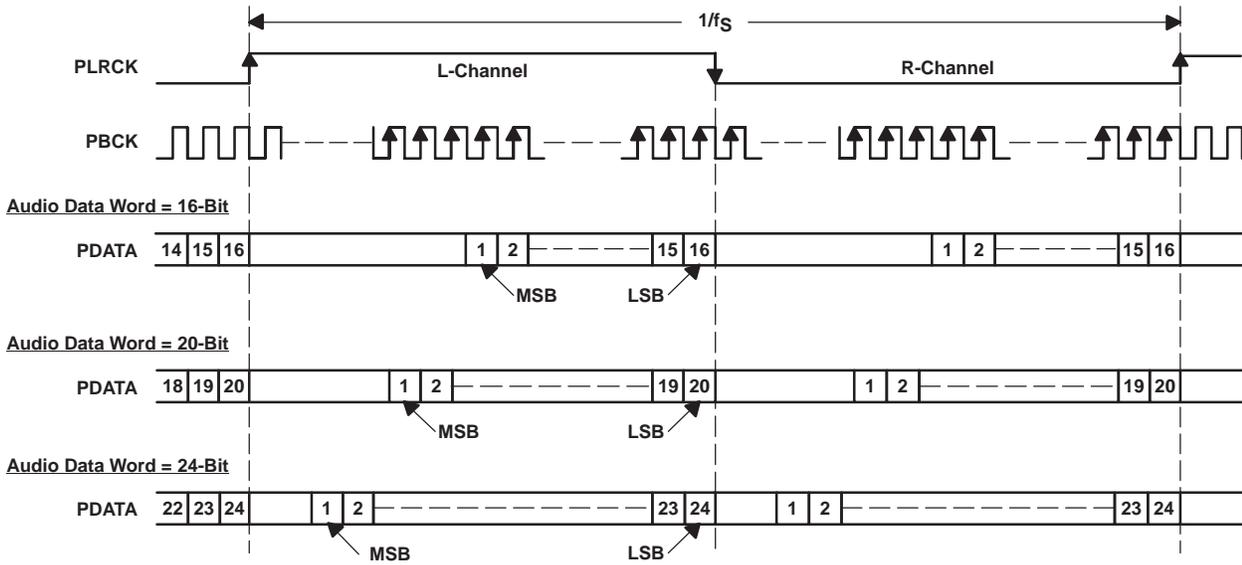
(2) Left Justified Data Format; L-Channel = HIGH, R-Channel = LOW
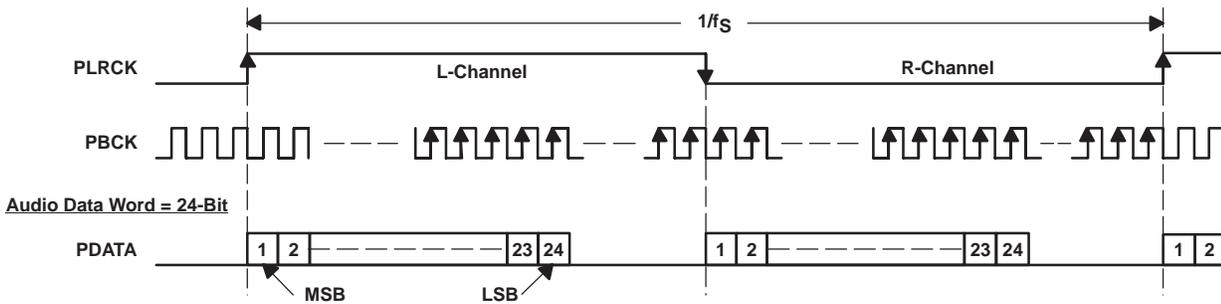
(3) I^{2}S Data Format; L-Channel = LOW, R-Channel = HIGH
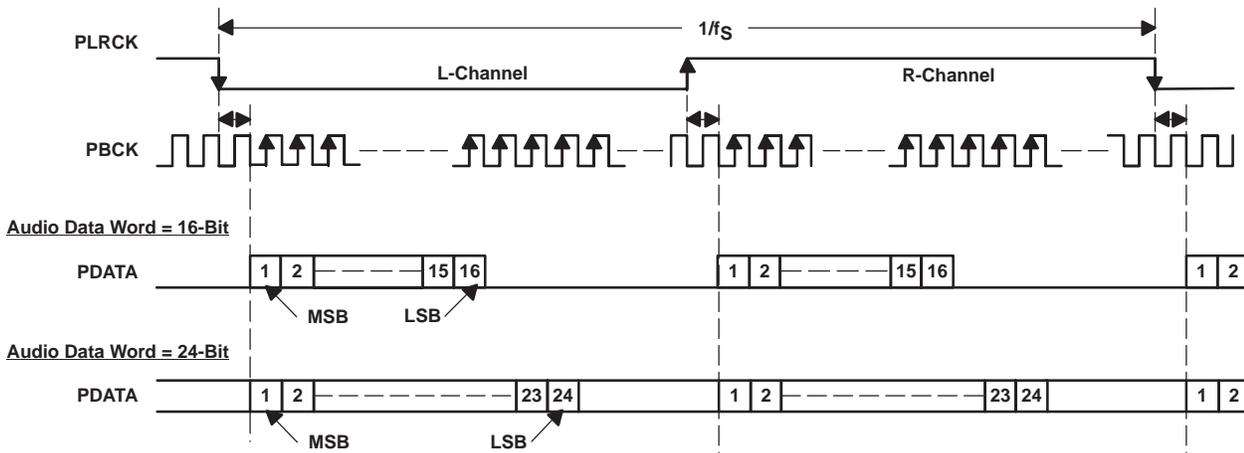
Figure 28. Audio Data Input Formats

External Digital Filter Interface and Timing

The DSD1794A supports an external digital filter interface with a 3- or 4-wire synchronous serial port, which allows the use of an external digital filter. External filters include the Texas Instruments DF1704 and DF1706, the Pacific Microsonics PMD200, or a programmable digital signal processor.

In the external DF mode, PLRCK (pin 4), PBCK (pin 6) and PDATA (pin 5) are defined as WDCK, the word clock; BCK, the bit clock; and DATA, the monaural data, respectively. The external digital filter interface is selected by using the DFTH bit of control register 20, which functions to bypass the internal digital filter of the DSD1794A.

When the DFMS bit of control register 19 is set, the DSD1794A can process stereo data. In this case, DSDL (pin 1) and DSDR (pin 2) are defined as L-channel data and R-channel data input, respectively.

Detailed information for the external digital filter interface mode is provided in the *APPLICATION FOR EXTERNAL DIGITAL FILTER INTERFACE* section of this data sheet.

Direct Stream Digital (DSD) Format Interface and Timing

The DSD1794A supports the DSD-format interface operation, which includes out-of-band noise filtering using an internal analog FIR filter. The DSD-format interface consists of a 3-wire synchronous serial port, which includes DBCK (pin 3), DSDL (pin 1), and DSDR (pin 2). DBCK is the serial bit clock. DSDL and DSDR are the L-channel and R-channel DSD data inputs, respectively. They are clocked into the DSD1794A on the rising edge of DBCK. PLRCK (pin 4) and PBCK (pin 6) are connected to GND in the DSD mode. The DSD-format interface is activated by setting the DSD bit of control register 20.

Detailed information for the DSD mode is provided in the *APPLICATION FOR DSD FORMAT (DSD MODE) INTERFACE* section of this data sheet.

SERIAL CONTROL INTERFACE (I^{2}C)

The DSD1794A supports the I^{2}C serial bus and the data transmission protocol for standard and fast mode as a slave device. This protocol is explained in I^{2}C specification 2.0.

Slave Address

MSB							LSB
1	0	0	1	1	ADR1	ADR0	R/$\overline{\text{W}}$

The DSD1794A has 7 bits for its own slave address. The first five bits (MSBs) of the slave address are factory preset to 10011. The next two bits of the address byte are the device select bits which can be user-defined by the ADR1 and ADR0 terminals. A maximum of four DSD1794As can be connected on the same bus at one time. Each DSD1794A responds when it receives its own slave address.

Packet Protocol

A master device must control packet protocol, which consists of start condition, slave address, read/write bit, data if write or acknowledge if read, and stop condition. The DSD1794A supports only slave receivers and slave transmitters.

R/$\overline{W}$: Read Operation if 1; Otherwise, Write Operation
ACK: Acknowledgement of a Byte if 0
DATA: 8 Bits (Byte)
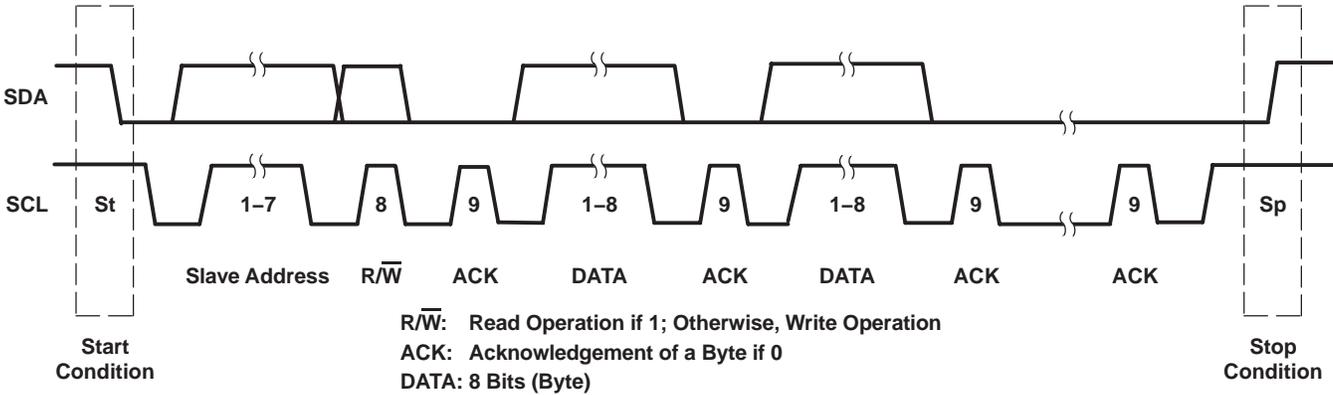
Write operation

Transmitter	M	M	M	S	M	S	M	S	…	S	M
Data Type	St	Slave Address	R/$\overline{W}$	ACK	DATA	ACK	DATA	ACK	…	ACK	Sp

Read operation

Transmitter	M	M	M	S	S	M	S	M	…	M	M
Data Type	St	Slave Address	R/$\overline{W}$	ACK	DATA	ACK	DATA	ACK	…	NACK	Sp

M: Master Device   *S*: Slave Device

St: Start Condition   Sp: Stop Condition

Figure 29. Basic I^{2}C Framework

Write Register

A master can write to any DSD1794A registers using single or multiple accesses. The master sends a DSD1794A slave address with a write bit, a register address, and the data. If multiple access is required, the address is that of the starting register, followed by the data to be transferred. When the data are received properly, the index register is incremented by 1 automatically. When the index register reaches 0x7F, the next value is 0x0. When undefined registers are accessed, the DSD1794A does not send an acknowledgement. Figure 30 is a diagram of the write operation.

Transmitter	M	M	M	S	M	S	M	S	M	S	…	S	M
Data Type	St	Slave Address	W	ACK	Reg Address	ACK	Write Data 1	ACK	Write Data 2	ACK	…	ACK	Sp

M: Master Device   *S*: Slave Device

St: Start Condition   W: Write   ACK: Acknowledge   Sp: Stop Condition

Figure 30. Write Operation

Read Register

A master can read the DSD1794A register. The value of the register address is stored in an indirect index register in advance. The master sends a DSD1794A slave address with a read bit after storing the register address. Then the DSD1794A transfers the data which the index register points to. When the data are transferred during a multiple access, the index register is incremented by 1 automatically. (When first going into read mode immediately following a write, the index register is not incremented. The master can read the register that was previously written.) When the index register reaches 0x7F, the next value is 0x0. The DSD1794A outputs some data when the index register is 0x10 to 0x1F, even if it is not defined in Table 3. Figure 31 is a diagram of the read operation.

Transmitter	M	M	M	S	M	S	M	M	M	S	S	M	…	M	M
Data Type	St	Slave Address	W	ACK	Reg Address	ACK	Sr	Slave Address	R	ACK	Data	ACK	…	NACK	Sp

M: Master Device   *S*: Slave Device

St: Start Condition   Sr: Repeated Start Condition   ACK: Acknowledge

Sp: Stop Condition   NACK: Not Acknowledge   W: Write   R: Read

NOTE: The slave address after the repeat start condition must be the same as the previous slave address.

Figure 31. Read Operation

Noise Suppression

The DSD1794A incorporates noise suppression using the system clock (SCK). However, there must be no more than two noise spikes in 600 ns. The noise suppression works for SCK frequencies between 8 MHz and 40 MHz in fast mode. However, it works incorrectly in the following conditions.

Case 1:

1. $t_{(SCK)}$ > 120 ns ($t_{(SCK)}$: period of SCK)
2. $t_{(HI)} + t_{(D-HD)} < t_{(SCK)} \times 5$
3. Spike noise exists on the first half of the SCL HIGH pulse.
4. Spike noise exists on the SDA HIGH pulse just before SDA goes LOW.
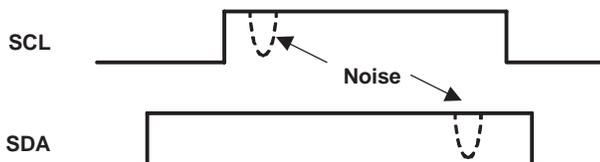
When these conditions occur at the same time, the data is recognized as LOW.

Case 2:

1. $t_{(SCK)}$ > 120 ns
2. $t_{(S-HD)}$ or $t_{(RS-HD)} < t_{(SCK)} \times 5$
3. Spike noise exists on both SCL and SDA during the hold time.
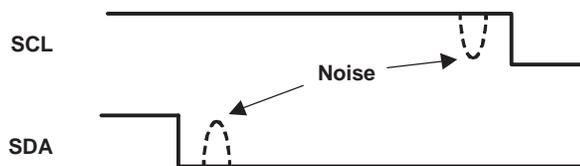
When these conditions occur at the same time, the DSD1794A fails to detect a start condition.

Case 3:

1. $t_{(SCK)}$ < 50 ns
2. $t_{(SP)} > t_{(SCK)}$
3. Spike noise exists on SCL just after SCL goes LOW.
4. Spike noise exists on SDA just before SCL goes LOW.
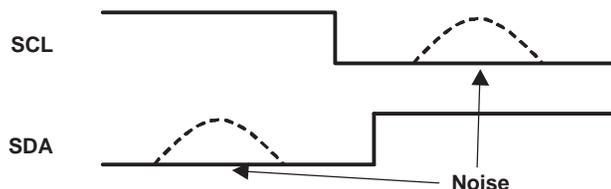
When these conditions occur at the same time, the DSD1794A erroneously detects a start or stop condition.

TIMING DIAGRAM

TIMING CHARACTERISTICS

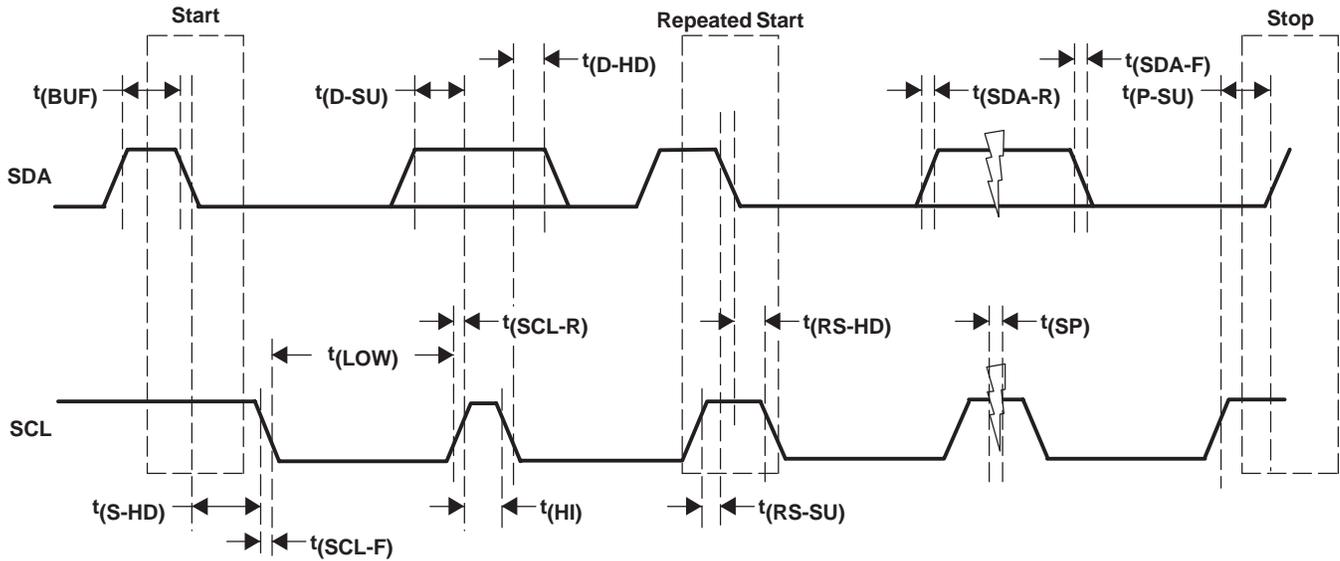| PARAMETER		CONDITIONS	MIN	MAX	UNIT
$t_{(SCL)}$	SCL clock frequency	Standard		100	kHz
		Fast		400	
$t_{(BUF)}$	Bus free time between stop and start conditions	Standard	4.7		µs
		Fast	1.3		
$t_{(LOW)}$	Low period of the SCL clock	Standard	4.7		µs
		Fast	1.3		
$t_{(HI)}$	High period of the SCL clock	Standard	4		µs
		Fast	600		ns
$t_{(RS-SU)}$	Setup time for (repeated) start condition	Standard	4.7		µs
		Fast	600		ns
$t_{(S-HD)}$ $t_{(RS-HD)}$	Hold time for (repeated) start condition	Standard	4		µs
		Fast	600		ns
$t_{(D-SU)}$	Data setup time	Standard	250		ns
		Fast	100		
$t_{(D-HD)}$	Data hold time	Standard	0	900	ns
		Fast	0	900	
$t_{(SCL-R)}$	Rise time of SCL signal	Standard	$20 + 0.1\,C_B$	1000	ns
		Fast	$20 + 0.1\,C_B$	300	
$t_{(SCL-R1)}$	Rise time of SCL signal after a repeated start condition and after an acknowledge bit	Standard	$20 + 0.1\,C_B$	1000	ns
		Fast	$20 + 0.1\,C_B$	300	
$t_{(SCL-F)}$	Fall time of SCL signal	Standard	$20 + 0.1\,C_B$	1000	ns
		Fast	$20 + 0.1\,C_B$	300	
$t_{(SDA-R)}$	Rise time of SDA signal	Standard	$20 + 0.1\,C_B$	1000	ns
		Fast	$20 + 0.1\,C_B$	300	
$t_{(SDA-F)}$	Fall time of SDA signal	Standard	$20 + 0.1\,C_B$	1000	ns
		Fast	$20 + 0.1\,C_B$	300	
$t_{(P-SU)}$	Setup time for stop condition	Standard	4		µs
		Fast	600		ns
$C_{(B)}$	Capacitive load for SDA and SCL line			400	pF
$t_{(SP)}$	Pulse duration of suppressed spike	Fast		50	ns
V_{NH}	Noise margin at high level for each connected device (including hysteresis)	Standard			
		Fast	$0.2\,V_{DD}$		V

MODE CONTROL REGISTERS

User-Programmable Mode Controls

The DSD1794A includes a number of user-programmable functions which are accessed via mode control registers. The registers are programmed using the serial control interface, discussed in the *SERIAL CONTROL INTERFACE (I2C)* section of this data sheet. Table 2 lists the available mode-control functions, along with their default reset conditions and associated register index.

Table 2. User-Programmable Function Controls

FUNCTION	DEFAULT	REGISTER	BIT	PCM	DSD	DF BYPASS
Digital attenuation control 0 dB to –120 dB and mute, 0.5 dB step	0 dB	Register 16 Register 17	ATL[7:0] (for L-ch) ATR[7:0] (for R-ch)	yes		
Attenuation load control—Disabled, enabled	Attenuation disabled	Register 18	ATLD	yes		
Input audio data format selection 16-, 20-, 24-bit standard (right-justified) format 24-bit MSB-first left-justified format 16-/24-bit I2S format	24-bit I2S format	Register 18	FMT[2:0]	yes		yes
Sampling rate selection for de-emphasis Disabled, 44.1 kHz, 48 kHz, 32 kHz	De-emphasis disabled	Register 18	DMF[1:0]	yes	yes[1]	
De-emphasis control—Disabled, enabled	De-emphasis disabled	Register 18	DME	yes		
Soft mute control—Mute disabled, enabled	Mute disabled	Register 18	MUTE	yes		
Output phase reversal—Normal, reverse	Normal	Register 19	REV	yes	yes	yes
Attenuation speed selection ×1 f_S, ×(1/2)f_S, ×(1/4)f_S, ×(1/8)f_S	×1 f_S	Register 19	ATS[1:0]	yes		
DAC operation control—Enabled, disabled	DAC operation enabled	Register 19	OPE	yes	yes	yes
Zero flag pin operation control DSD data input, zero flag output	DSD data input	Register 19	ZOE	yes		yes
Stereo DF bypass mode select Monaural, stereo	Monaural	Register 19	DFMS			yes
Digital filter rolloff selection Sharp rolloff, slow rolloff	Sharp rolloff	Register 19	FLT	yes		
Infinite zero mute control Disabled, enabled	Disabled	Register 19	INZD	yes		yes
System reset control Reset operation , normal operation	Normal operation	Register 20	SRST	yes	yes	yes
DSD interface mode control DSD enabled, disabled	Disabled	Register 20	DSD		yes	
Digital-filter bypass control DF enabled, DF bypass	DF enabled	Register 20	DFTH			yes
Monaural mode selection Stereo, monaural	Stereo	Register 20	MONO	yes	yes	yes
Channel selection for monaural mode data L-channel, R-channel	L-channel	Register 20	CHSL	yes	yes	yes
Delta-sigma oversampling rate selection ×64 f_S, ×128 f_S, ×32 f_S	×64 f_S	Register 20	OS[1:0]	yes	yes[2]	yes
PCM zero output enable	Enabled	Register 21	PCMZ	yes		yes
DSD zero output enable	Disabled	Register 21	DZ[1:0]		yes	
Function available only for read						
Zero detection flag Not zero, zero detected	Not zero = 0 Zero detected = 1	Register 22	ZFGL (for L-ch) ZFGR (for R-ch)	yes	yes	yes

[1] When in DSD mode, DMF[1:0] is defined as DSD filter (analog FIR) performance selection.
[2] When in DSD mode, OS[1:0] is defined as DSD filter (analog FIR) operation rate selection.

Register Map

The mode control register map is shown in Table 3. Registers 16–21 include an R/$\overline{W}$ bit, which determines whether a register read (R/$\overline{W}$ = 1) or write (R/$\overline{W}$ = 0) operation is performed. Register 22 is read-only.

Table 3. Mode Control Register Map

	B15	B14	B13	B12	B11	B10	B9	B8	B7	B6	B5	B4	B3	B2	B1	B0
Register 16	R/$\overline{W}$	0	0	1	0	0	0	0	ATL7	ATL6	ATL5	ATL4	ATL3	ATL2	ATL1	ATL0
Register 17	R/$\overline{W}$	0	0	1	0	0	0	1	ATR7	ATR6	ATR5	ATR4	ATR3	ATR2	ATR1	ATR0
Register 18	R/$\overline{W}$	0	0	1	0	0	1	0	ATLD	FMT2	FMT1	FMT0	DMF1	DMF0	DME	MUTE
Register 19	R/$\overline{W}$	0	0	1	0	0	1	1	REV	ATS1	ATS0	OPE	ZOE	DFMS	FLT	INZD
Register 20	R/$\overline{W}$	0	0	1	0	1	0	0	RSV	SRST	DSD	DFTH	MONO	CHSL	OS1	OS0
Register 21	R/$\overline{W}$	0	0	1	0	1	0	1	RSV	RSV	RSV	RSV	RSV	DZ1	DZ0	PCMZ
Register 22	R	0	0	1	0	1	1	0	RSV	RSV	RSV	RSV	RSV	RSV	ZFGR	ZFGL

Register Definitions

	B15	B14	B13	B12	B11	B10	B9	B8	B7	B6	B5	B4	B3	B2	B1	B0
Register 16	R/$\overline{W}$	0	0	1	0	0	0	0	ATL7	ATL6	ATL5	ATL4	ATL3	ATL2	ATL1	ATL0
Register 17	R/$\overline{W}$	0	0	1	0	0	0	1	ATR7	ATR6	ATR5	ATR4	ATR3	ATR2	ATR1	ATR0

R/$\overline{W}$: Read/Write Mode Select

When R/$\overline{W}$ = 0, a write operation is performed.

When R/$\overline{W}$ = 1, a read operation is performed.

Default value: 0

ATx[7:0]: Digital Attenuation Level Setting

These bits are available for read and write.

Default value: 1111 1111b

Each DAC output has a digital attenuator associated with it. The attenuator can be set from 0 dB to –120 dB, in 0.5-dB steps. Alternatively, the attenuator can be set to infinite attenuation (or mute).

The attenuation data for each channel can be set individually. However, the data load control (the ATLD bit of control register 18) is common to both attenuators. ATLD must be set to 1 in order to change an attenuator setting. The attenuation level can be set using the following formula:

Attenuation level (dB) = 0.5 dB • (ATx[7:0]$_{DEC}$ − 255)

where ATx[7:0]$_{DEC}$ = 0 through 255

For ATx[7:0]$_{DEC}$ = 0 through 14, the attenuator is set to infinite attenuation. Table 4 shows attenuation levels for various settings.

Table 4. Digital Attenuation Levels

ATx[7:0]	Decimal Value	Attenuation Level Setting
1111 1111b	255	0 dB, no attenuation (default)
1111 1110b	254	−0.5 dB
1111 1101b	253	−1.0 dB
⋮	⋮	⋮
0001 0000b	16	−119.5 dB
0000 1111b	15	−120.0 dB
0000 1110b	14	Mute
⋮	⋮	⋮
0000 0000b	0	Mute

	B15	B14	B13	B12	B11	B10	B9	B8	B7	B6	B5	B4	B3	B2	B1	B0
Register 18	R/$\overline{\text{W}}$	0	0	1	0	0	1	0	ATLD	FMT2	FMT1	FMT0	DMF1	DMF0	DME	MUTE

R/$\overline{\text{W}}$: Read/Write Mode Select

When R/$\overline{\text{W}}$ = 0, a write operation is performed.

When R/$\overline{\text{W}}$ = 1, a read operation is performed.

Default value: 0

ATLD: Attenuation Load Control

This bit is available for read and write.

Default value: 0

ATLD = 0	Attenuation control disabled (default)
ATLD = 1	Attenuation control enabled

The ATLD bit is used to enable loading of the attenuation data contained in registers 16 and 17. When ATLD = 0, the attenuation settings remain at the previously programmed levels, ignoring new data loaded from registers 16 and 17. When ATLD = 1, attenuation data written to registers 16 and 17 is loaded normally.

FMT[2:0]: Audio Interface Data Format

These bits are available for read and write.

Default value: 101

For the external digital filter interface mode (DFTH mode), this register is operated as shown in the *Application for Interfacing With an External Digital Filter* section of this data sheet.

FMT[2:0]	Audio Data Format Selection
000	16-bit standard format, right-justified data
001	20-bit standard format, right-justified data
010	24-bit standard format, right-justified data
011	24-bit MSB-first, left-justified format data
100	16-bit I^{2}S-format data
101	24-bit I^{2}S-format data (default)
110	Reserved
111	Reserved

The FMT[2:0] bits are used to select the data format for the serial audio interface.

DMF[1:0]: Sampling Frequency Selection for the De-Emphasis Function

These bits are available for read and write.

Default value: 00

DMF[1:0]	De-Emphasis Sampling Frequency Selection
00	Disabled (default)
01	48 kHz
10	44.1 kHz
11	32 kHz

The DMF[1:0] bits are used to select the sampling frequency used by the digital de-emphasis function when it is enabled by setting the DME bit. The de-emphasis curves are shown in the *TYPICAL PERFORMANCE CURVES* section of this data sheet.

For the DSD mode, analog FIR filter performance can be selected using this register. A register map and filter response plots are shown in the *APPLICATION FOR DSD FORMAT (DSD MODE) INTERFACE* section of this data sheet.

DME: Digital De-Emphasis Control

This bit is available for read and write.

Default value: 0

DME = 0	De-emphasis disabled (default)
DME = 1	De-emphasis enabled

The DME bit is used to enable or disable the de-emphasis function for both channels.

MUTE: Soft Mute Control

This bit is available for read and write.

Default value: 0

MUTE = 0	MUTE disabled (default)
MUTE = 1	MUTE enabled

The MUTE bit is used to enable or disable the soft mute function for both channels.

Soft mute is operated as a 256-step attenuator. The speed for each step to –∞ dB (mute) is determined by the attenuation rate selected in the ATS register.

	B15	B14	B13	B12	B11	B10	B9	B8	B7	B6	B5	B4	B3	B2	B1	B0
Register 19	R/$\overline{\text{W}}$	0	0	1	0	0	1	1	REV	ATS1	ATS0	OPE	ZOE	DFMS	FLT	INZD

R/$\overline{\text{W}}$: Read/Write Mode Select

When R/$\overline{\text{W}}$ = 0, a write operation is performed.

When R/$\overline{\text{W}}$ = 1, a read operation is performed.

Default value: 0

REV: Output Phase Reversal

This bit is available for read and write.

Default value: 0

REV = 0	Normal output (default)
REV = 1	Inverted output

The REV bit is used to invert the output phase for both channels.

ATS[1:0]: Attenuation Rate Select

These bits are available for read and write.

Default value: 00

ATS[1:0]	Attenuation Rate Selection
00	Every PLRCK (default)
01	PLRCK/2
10	PLRCK/4
11	PLRCK/8

The ATS[1:0] bits are used to select the rate at which the attenuator is decremented/incremented during level transitions.

OPE: DAC Operation Control

This bit is available for read and write.

Default value: 0

OPE = 0	DAC operation enabled (default)
OPE = 1	DAC operation disabled

The OPE bit is used to enable or disable the analog output for both channels. Disabling the analog outputs forces them to the bipolar zero level (BPZ) even if digital audio data is present on the input.

ZOE: Zero Flag Pin Operation Control

This bit is available for read and write.

Default value: 0

ZOE = 0	DSD data input (default)
ZOE = 1	Zero flag output

The ZOE bit is used to change the DSDL (pin 1) and DSDR (pin 2) pin assignments. When the ZOE bit is set to 0, DSDL and DSDR are inputs for L-channel and R-channel data. When the ZOE bit is set to 1, DSDL and DSDR become outputs for the L-channel and R-channel zero flags, respectively. See the PCMZ and DZ[1:0] bit descriptions of register 21.

DFMS: Stereo DF Bypass Mode Select

This bit is available for read and write.

Default value: 0

DFMS = 0	Monaural (default)
DFMS = 1	Stereo input enabled

The DFMS bit is used to enable stereo operation in DF bypass mode. In the DF bypass mode, when DFMS is set to 0, the pin for the input data is PDATA (pin 5) only, therefore the DSD1794A operates as a monaural DAC. When DFMS is set to 1, the DSD1794A can operate as a stereo DAC with inputs of L-channel and R-channel data on DSDL (pin 1) and DSDR (pin 2), respectively.

FLT: Digital Filter Rolloff Control

This bit is available for read and write.

Default value: 0

FLT = 0	Sharp rolloff (default)
FLT = 1	Slow rolloff

The FLT bit is used to select the digital filter rolloff characteristic. The filter responses for these selections are shown in the *TYPICAL PERFORMANCE CURVES* section of this data sheet.

INZD: Infinite Zero Detect Mute Control

This bit is available for read and write.

Default value: 0

INZD = 0	Infinite zero detect mute disabled (default)
INZD = 1	Infinite zero detect mute enabled

The INZD bit is used to enable or disable the zero detect mute function. Setting INZD to 1 forces muted analog outputs to hold a bipolar zero level when the DSD1794A detects zero data in both channels continuously for 1024 sampling periods ($1/f_S$). The infinite zero detect mute function does not work in the DSD mode.

	B15	B14	B13	B12	B11	B10	B9	B8	B7	B6	B5	B4	B3	B2	B1	B0
Register 20	R/$\overline{W}$	0	0	1	0	1	0	0	RSV	SRST	DSD	DFTH	MONO	CHSL	OS1	OS0

R/$\overline{W}$: Read/Write Mode Select

When R/$\overline{W}$ = 0, a write operation is performed.

When R/$\overline{W}$ = 1, a read operation is performed.

Default value: 0

SRST: System Reset Control

This bit is available for write only.

Default value: 0

SRST = 0	Normal operation (default)
SRST = 1	System reset operation (generate one reset pulse)

The SRST bit is used to reset the DSD1794A to the initial system condition.

DSD: DSD Interface Mode Control

This bit is available for read and write.

Default value: 0

DSD = 0	DSD interface mode disabled (default)
DSD = 1	DSD interface mode enabled

The DSD bit is used to enable or disable the DSD interface mode.

DFTH: Digital Filter Bypass (or Through Mode) Control

This bit is available for read and write.

Default value: 0

DFTH = 0	Digital filter enabled (default)
DFTH = 1	Digital filter bypassed for external digital filter

The DFTH bit is used to enable or disable the external digital filter interface mode.

MONO: Monaural Mode Selection

This bit is available for read and write.

Default value: 0

MONO = 0	Stereo mode (default)
MONO = 1	Monaural mode

The MONO function is used to change the operation mode from the normal stereo mode to the monaural mode. When the monaural mode is selected, both DACs operate in a balanced mode for one channel of audio input data. Channel selection is available for L-channel or R-channel data, determined by the CHSL bit as described immediately following.

CHSL: Channel Selection for Monaural Mode

This bit is available for read and write.

Default value: 0

This bit is available when MONO = 1.

CHSL = 0	L-channel selected (default)
CHSL = 1	R-channel selected

The CHSL bit selects L-channel or R-channel data to be used in monaural mode.

OS[1:0]: Delta-Sigma Oversampling Rate Selection

These bits are available for read and write.

Default value: 00

OS[1:0]	Operation Speed Select
00	64 times f_S (default)
01	32 times f_S
10	128 times f_S
11	Reserved

The OS bits are used to change the oversampling rate of delta-sigma modulation. Use of this function enables the designer to stabilize the conditions at the post low-pass filter for different sampling rates. As an application example, programming to set 128 times in 44.1-kHz operation, 64 times in 96-kHz operation, and 32 times in 192-kHz operation allows the use of only a single type (cutoff frequency) of post low-pass filter. The 128-f_S oversampling rate is not available at sampling rates above 100 kHz. If the 128-f_S oversampling rate is selected, a system clock of more than 256 f_S is required.

In DSD mode, these bits are used to select the speed of the bit clock for DSD data coming into the analog FIR filter.

	B15	B14	B13	B12	B11	B10	B9	B8	B7	B6	B5	B4	B3	B2	B1	B0
Register 21	R/$\overline{W}$	0	0	1	0	1	0	1	RSV	RSV	RSV	RSV	RSV	DZ1	DZ0	PCMZ

R/$\overline{W}$: Read/Write Mode Select

When R/$\overline{W}$ = 0, a write operation is performed.

When R/$\overline{W}$ = 1, a read operation is performed.

Default value: 0

DZ[1:0]: DSD Zero Output Enable

These bits are available for read and write.

Default value: 00

DZ[1:0]	Zero Output Enable
00	Disabled (default)
01	Even pattern detect
1x	96_H pattern detect

The DZ bits are used to enable or disable the output zero flags, and to select the zero pattern in the DSD mode. The DSD1794A sets zero flags when the number of 1s and 0s are equal in every 8 bits of DSD input data, or the DSD input data is 1001 0110 continuously for 200 ms.

PCMZ: PCM Zero Output Enable

These bits are available for read and write.

Default value: 1

PCMZ = 0	PCM zero output disabled
PCMZ = 1	PCM zero output enabled (default)

The PCMZ bit is used to enable or disable the output zero flags in the PCM mode and the external DF mode. The DSD1794A sets the zero flags when the input data is continuously zero for 1024 LRCKs in the PCM mode or 1024 WDCKs in the external filter mode.

	B15	B14	B13	B12	B11	B10	B9	B8	B7	B6	B5	B4	B3	B2	B1	B0
Register 22	R	0	0	1	0	1	1	0	RSV	RSV	RSV	RSV	RSV	RSV	ZFGR	ZFGL

R: Read Mode Select

Value is always 1, specifying the readback mode.

ZFGx: Zero-Detection Flag

Where x = L or R, corresponding to the DAC output channel. These bits are available only for readback.

Default value: 00

ZFGx = 0	Not zero
ZFGx = 1	Zero detected

When the DSD1794A detects that audio input data is continuously zero, the ZFGx bit is set to 1 for the corresponding channel(s).
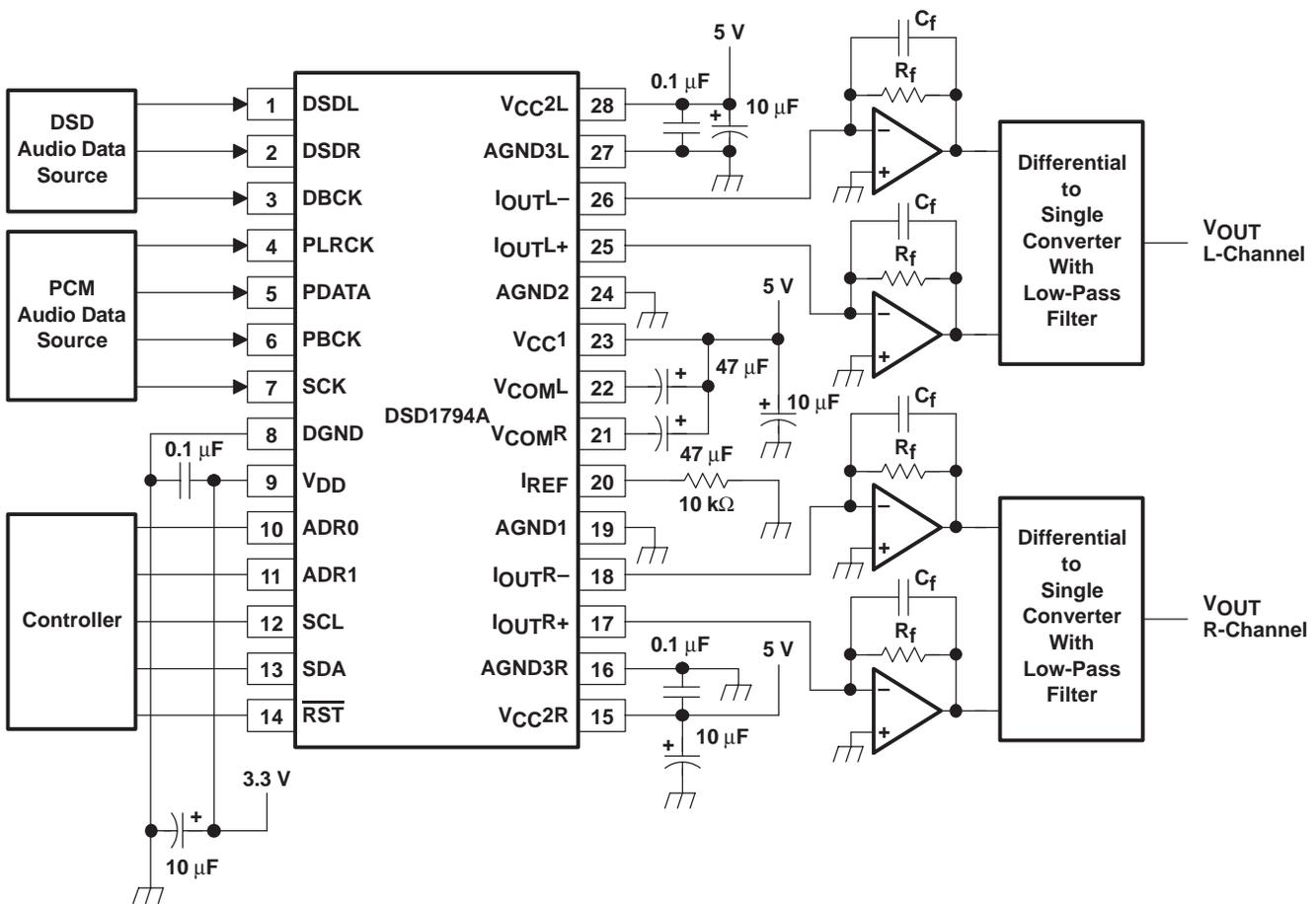
TYPICAL CONNECTION DIAGRAM

Figure 32. Typical Application Circuit

APPLICATION INFORMATION

APPLICATION CIRCUIT

The design of the application circuit is very important in order to actually realize the high S/N ratio of which the DSD1794A is capable. This is because noise and distortion that are generated in an application circuit are not negligible.

In the circuit of Figure 33, the output level is 2 V RMS and 127 dB S/N is achieved. The circuit of Figure 34 can realize the highest performance. In this case the output level is set to 4.5 V rms and 129 dB S/N is achieved (stereo mode). In monaural mode, if the output of the L-channel and R-channel is used as a balanced output, 132 dB S/N is achieved (see Figure 36).

Figure 35 shows a circuit for the DSD mode, which is a 4th-order LPF in order to reduce the out-of-band noise.

I/V Section

The current of the DSD1794A on each of the output pins ($I_{OUT}L+$, $I_{OUT}L-$, $I_{OUT}R+$, $I_{OUT}R-$) is 7.8 mA p-p at 0 dB (full scale). The voltage output level of the I/V converter (Vi) is given by following equation:

$$Vi = 7.8 \text{ mA p-p} \times R_f \quad (R_f: \text{feedback resistance of I/V converter})$$

An NE5534 operational amplifier is recommended for the I/V circuit to obtain the specified performance. Dynamic performance such as the gain bandwidth, settling time, and slew rate of the operational amplifier affects the audio dynamic performance of the I/V section.

Differential Section

The DSD1794A voltage outputs are followed by differential amplifier stages, which sum the differential signals for each channel, creating a single-ended I/V op-amp output. In addition, the differential amplifiers provide a low-pass filter function.

The operational amplifier recommended for the differential circuit is the Linear Technology LT1028, because its input noise is low.
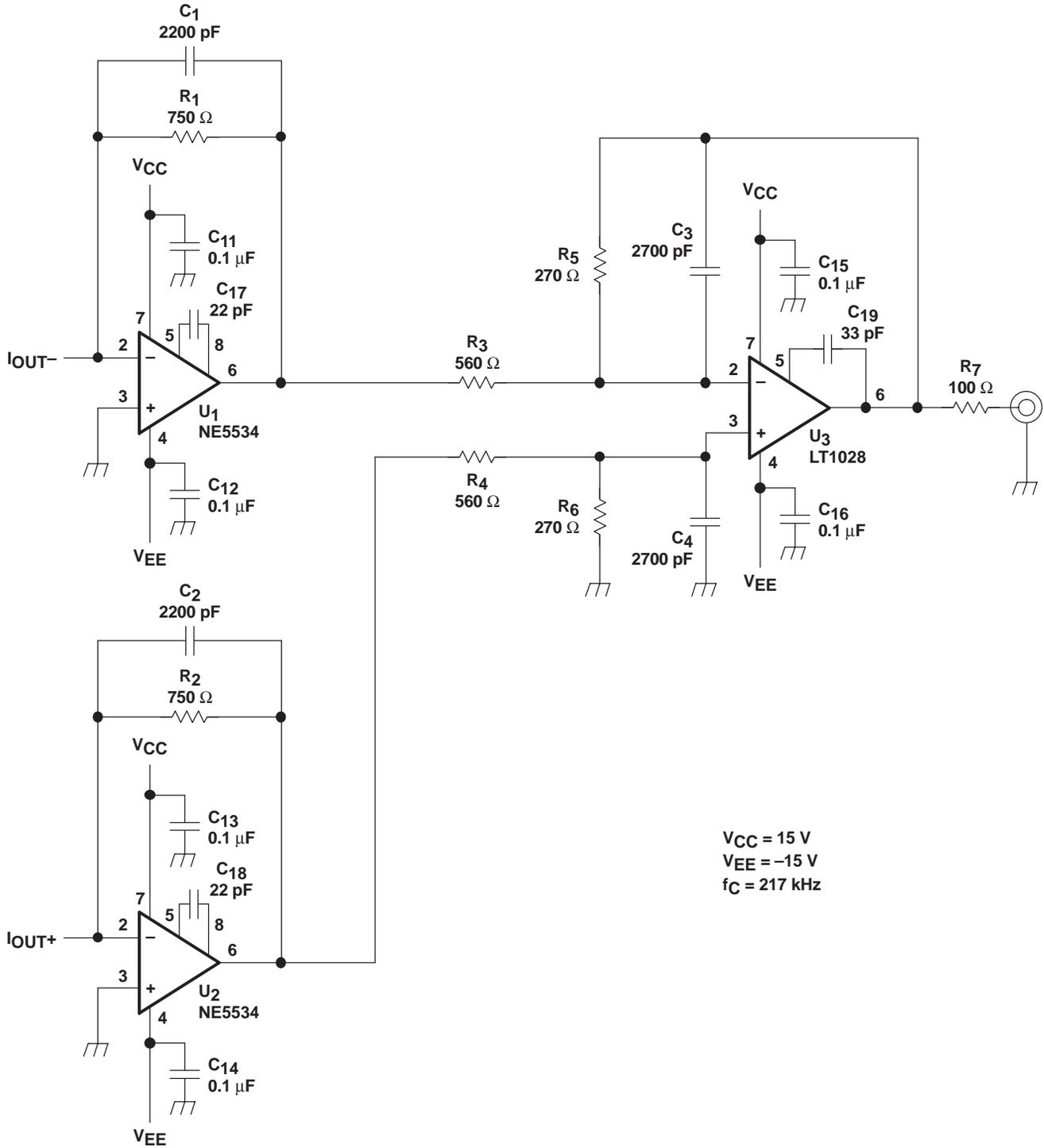
Figure 33. Measurement Circuit for PCM, V_{OUT} = 2 V RMS
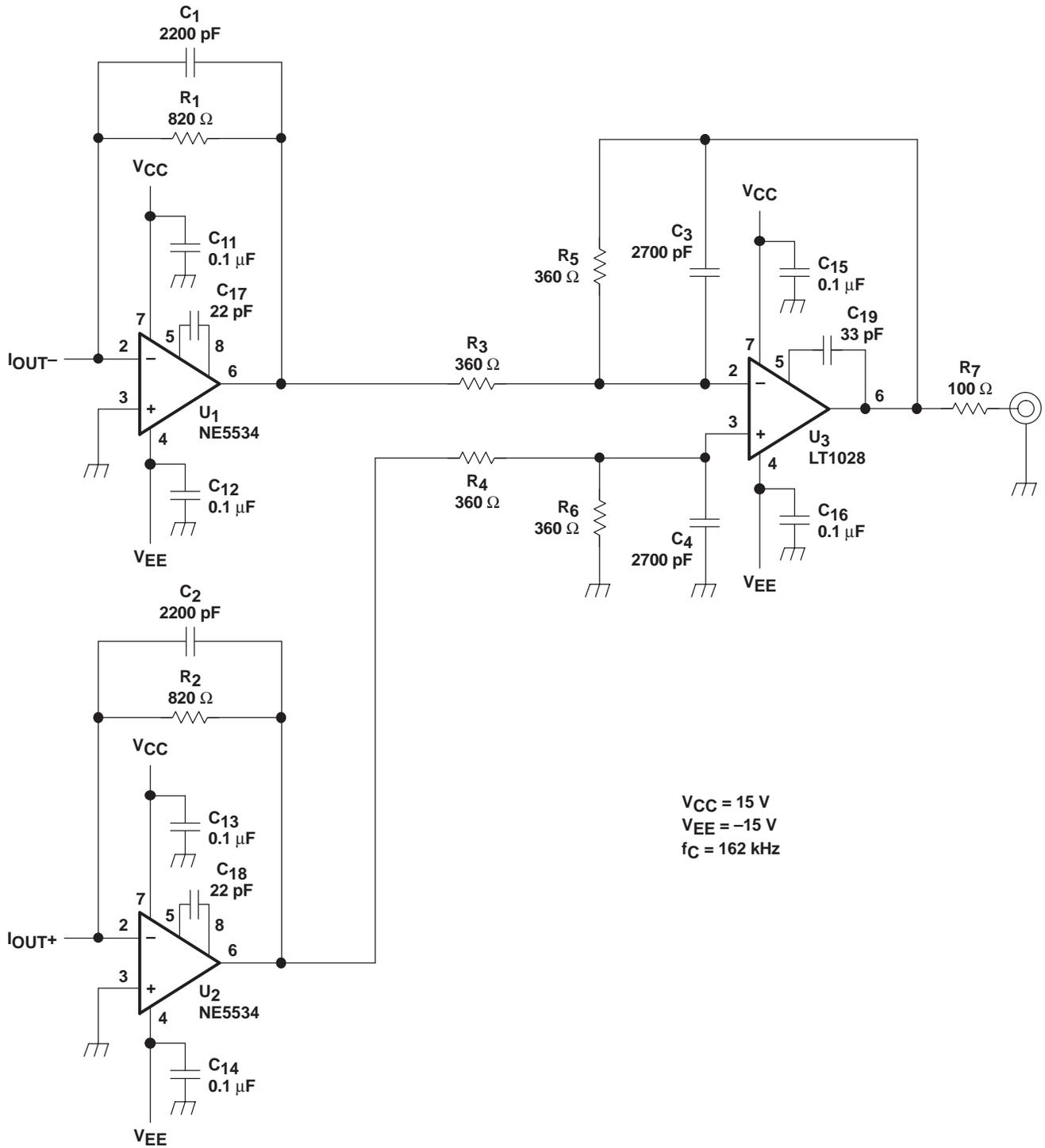
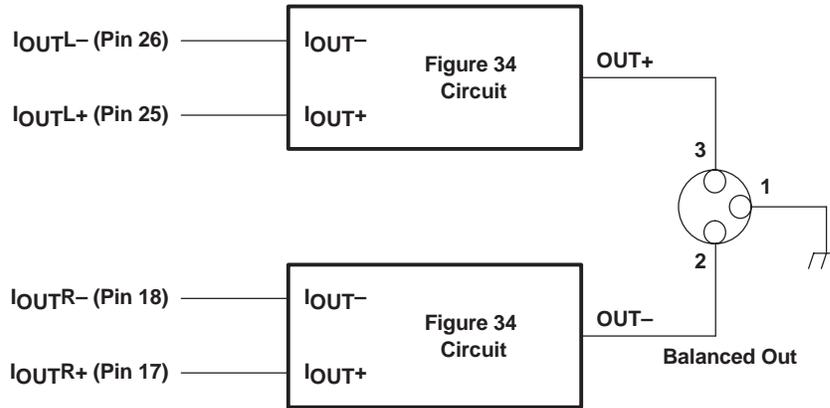**Figure 34. Measurement Circuit for PCM, V$_{OUT}$ = 4.5 V RMS**

V$_{CC}$ = 15 V
V$_{EE}$ = −15 V
f$_C$ = 162 kHz

Figure 35. Measurement Circuit for DSD

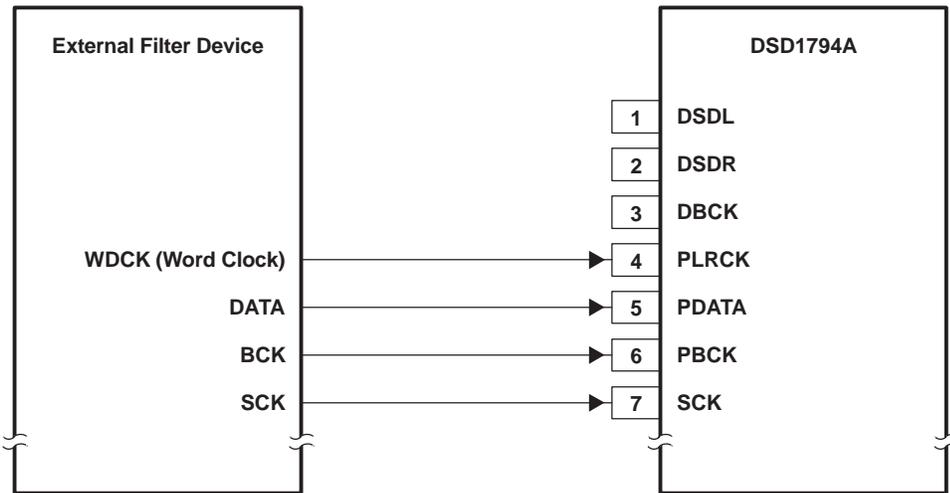**Figure 36. Measurement Circuit for Monaural Mode**
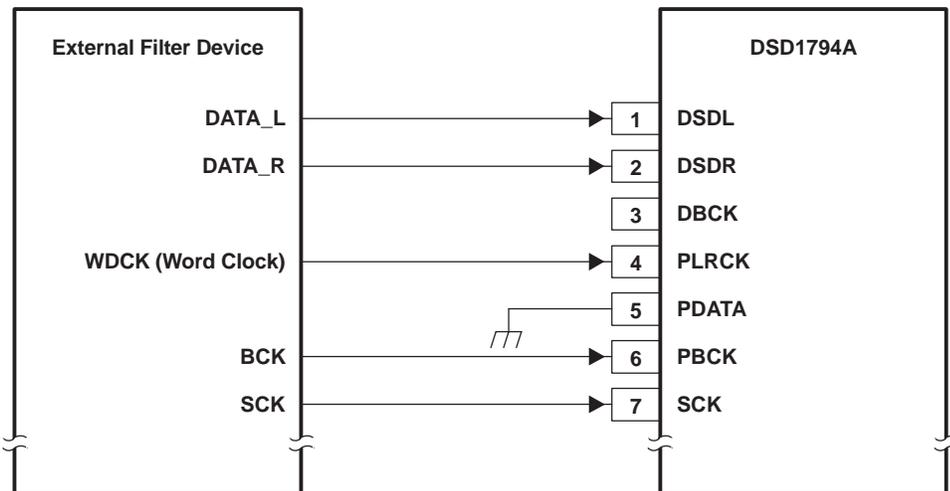
APPLICATION FOR EXTERNAL DIGITAL FILTER INTERFACE

Figure 37. Connection Diagram for External Digital Filter (Internal DF Bypass Mode) Application

Application for Interfacing With an External Digital Filter

For some applications, it may be desirable to use an external digital filter to perform the interpolation function, as it can provide improved stop-band attenuation when compared to the internal digital filter of the DSD1794A.

The DSD1794A supports several external digital filters, including:

- Texas Instruments DF1704 and DF1706
- Pacific Microsonics PMD200 HDCD filter/decoder IC
- Programmable digital signal processors

The external digital filter application mode is accessed by programming the following bit in the corresponding control register:

- DFTH = 1 (register 20)

The pins used to provide the serial interface for the external digital filter are shown in the connection diagram of Figure 37. The word clock (WDCK) signal must be operated at 8× or 4× the desired sampling frequency, f_S.

System Clock (SCK) and Interface Timing

The DSD1794A in an application using an external digital filter requires the synchronization of WDCK and the system clock. The system clock is phase-free with respect to WDCK. Interface timing among WDCK, BCK, and DATA is shown in Figure 39.

Audio Format

The DSD1794A in the external digital filter interface mode supports right-justified audio formats including 16-bit, 20-bit, and 24-bit audio data, as shown in Figure 38. The audio format is selected by the FMT[2:0] bits of control register 18.
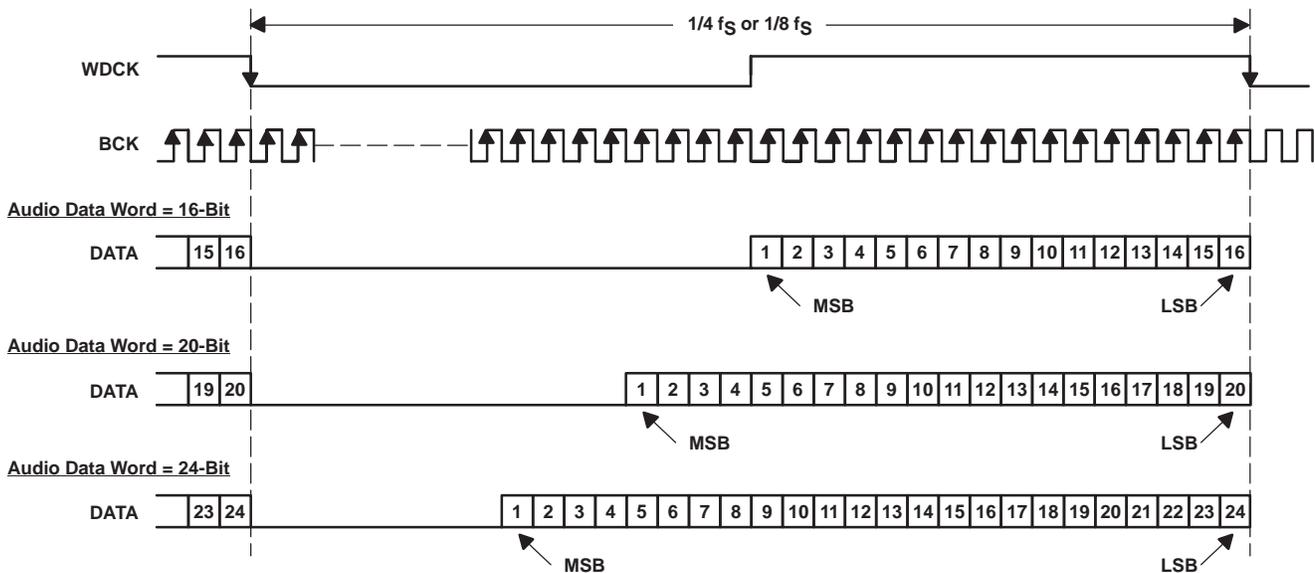

Figure 38. Audio Data Input Format for External Digital Filter (Internal DF Bypass Mode) Application
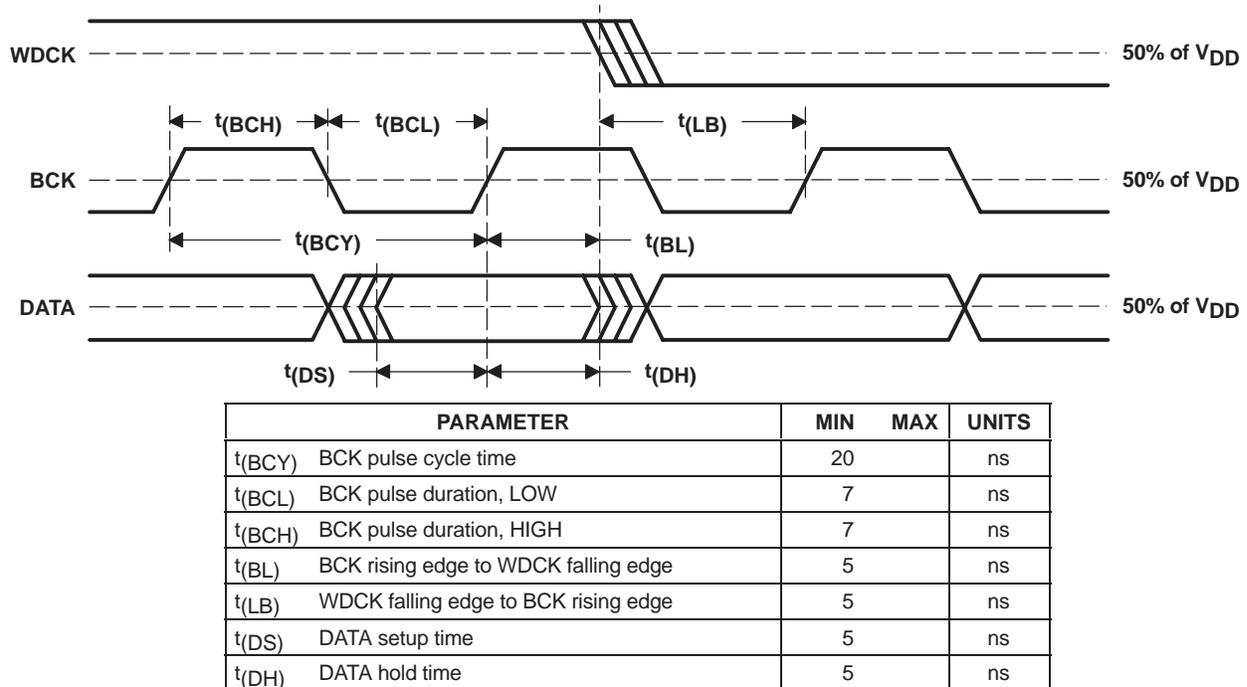
PARAMETER		MIN	MAX	UNITS
t(BCY)	BCK pulse cycle time	20		ns
t(BCL)	BCK pulse duration, LOW	7		ns
t(BCH)	BCK pulse duration, HIGH	7		ns
t(BL)	BCK rising edge to WDCK falling edge	5		ns
t(LB)	WDCK falling edge to BCK rising edge	5		ns
t(DS)	DATA setup time	5		ns
t(DH)	DATA hold time	5		ns

Figure 39. Audio Interface Timing for External Digital Filter (Internal DF Bypass Mode) Application

Functions Available in the External Digital Filter Mode

The external digital filter mode allows access to the majority of the DSD1794A mode control functions.

The following table shows the register mapping available when the external digital filter mode is selected, along with descriptions of functions which are modified when using this mode selection.

	B15	B14	B13	B12	B11	B10	B9	B8	B7	B6	B5	B4	B3	B2	B1	B0
Register 16	R/$\overline{\text{W}}$	0	0	1	0	0	0	0	–	–	–	–	–	–	–	–
Register 17	R/$\overline{\text{W}}$	0	0	1	0	0	0	1	–	–	–	–	–	–	–	–
Register 18	R/$\overline{\text{W}}$	0	0	1	0	0	1	0	–	FMT2	FMT1	FMT0	–	–	–	–
Register 19	R/$\overline{\text{W}}$	0	0	1	0	0	1	1	REV	–	–	OPE	–	DFMS	–	INZD
Register 20	R/$\overline{\text{W}}$	0	0	1	0	1	0	0	–	SRST	0	1	MONO	CHSL	OS1	OS0
Register 21	R/$\overline{\text{W}}$	0	0	1	0	1	0	1	–	–	–	–	–	–	–	PCMZ
Register 22	R	0	0	1	0	1	1	0	–	–	–	–	–	–	ZFGR	ZFGL

NOTE: 1 indicates that the bit is required for selection of external digital filter mode.
– indicates that function is disabled. No operation even if data bit is set

FMT[2:0]: Audio Data Format Selection

Default value: 000

FMT[2:0]	Audio Data Format Select
000	16-bit right-justified format (default)
001	20-bit right-justified format
010	24-bit right-justified format
Other	N/A

OS[1:0]: Delta-Sigma Modulator Oversampling Rate Selection

Default value: 00

OS[1:0]	Operation Speed Select
00	8 times WDCK (default)
01	4 times WDCK
10	16 times WDCK
11	Reserved

The effective oversampling rate is determined by the oversampling performed by both the external digital filter and the delta-sigma modulator. For example, if the external digital filter is 8× oversampling, and the user selects OS[1:0] = 00, then the delta-sigma modulator oversamples by 8×, resulting in an effective oversampling rate of 64×. The 16× WDCK oversampling rate is not available above a 100-kHz sampling rate. If the oversampling rate selected is 16× WDCK, the system clock frequency must be over 256 f_S.

APPLICATION FOR DSD FORMAT (DSD MODE) INTERFACE
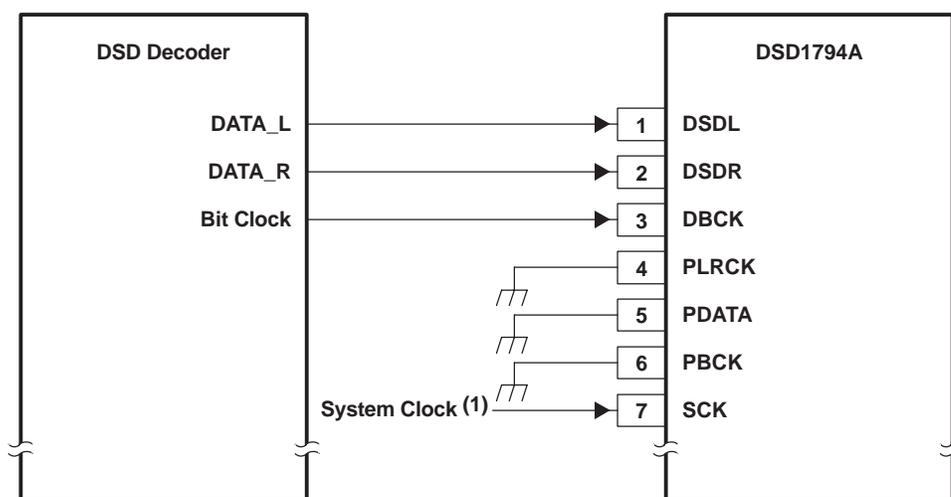
(1) The system clock is necessary for the initilaization sequence and the I^{2}C interface operation.

Figure 40. Connection Diagram in DSD Mode

Feature

This mode is used for interfacing directly to a DSD decoder, which is found in Super Audio CD™ (SACD) applications.

The DSD mode is accessed by programming the following bit in the corresponding control register.

> DSD = 1 (register 20)

The DSD mode provides a low-pass filtering function to convert the 1-bit oversampled data stream to the analog domain. The filtering is provided using an analog FIR filter structure. Four FIR responses are available, and are selected by the DMF[1:0] bits of control register 18.

The DSD bit must be set before inputting DSD data, otherwise the DSD1794A erroneously detects the TDMCA mode, and commands are not accepted through the serial control interface.

Pin Assignment When DSD Format Interface

- DSDL (pin 1): L-channel DSD data input
- DSDR (pin 2): R-channel DSD data input
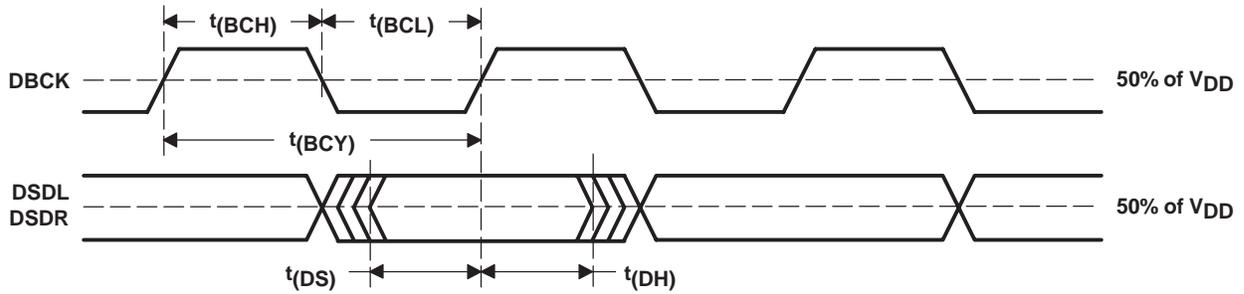- DBCK (pin 3): Bit clock (BCK) for DSD data

Super Audio CD is a trademark of Sony Kabushiki Kaisha TA Sony Corporation, Japan.
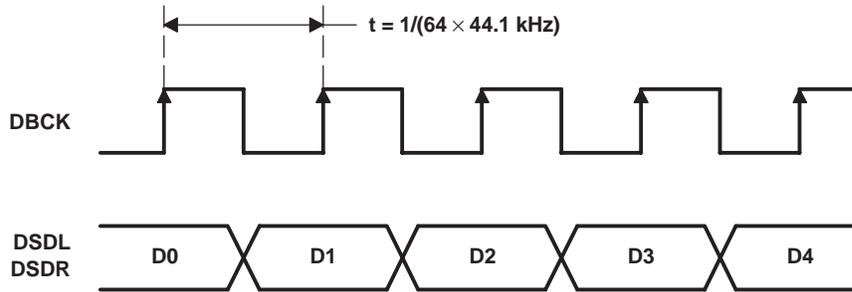
Figure 41. Normal Data Output Form From DSD Decoder

PARAMETER		MIN	MAX	UNITS
t$_{(BCY)}$	DBCK pulse cycle time	85[1]		ns
t$_{(BCH)}$	DBCK high-level time	30		ns
t$_{(BCL)}$	DBCK low-level time	30		ns
t$_{(DS)}$	DSDL, DSDR setup time	10		ns
t$_{(DH)}$	DSDL, DSDR hold time	10		ns

[1] 2.8224 MHz × 4. (2.8224 MHz = 64 × 44.1 kHz. This value is specified as a sampling rate of DSD.)

Figure 42. Timing for DSD Audio Interface

ANALOG FIR FILTER PERFORMANCE IN DSD MODE
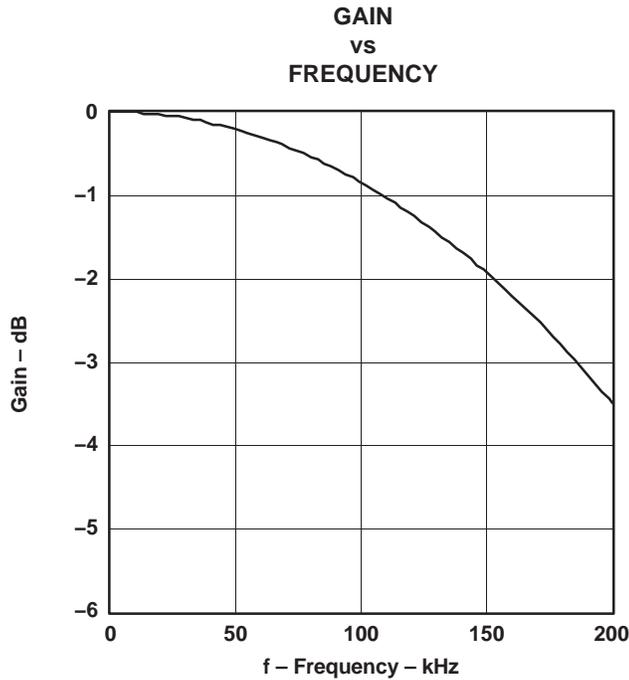
GAIN
vs
FREQUENCY

Figure 43. DSD Filter-1, Low BW

GAIN
vs
FREQUENCY

$f_C = 185 \text{ kHz}$
$\text{Gain}^{(1)} = -6.6 \text{ dB}$
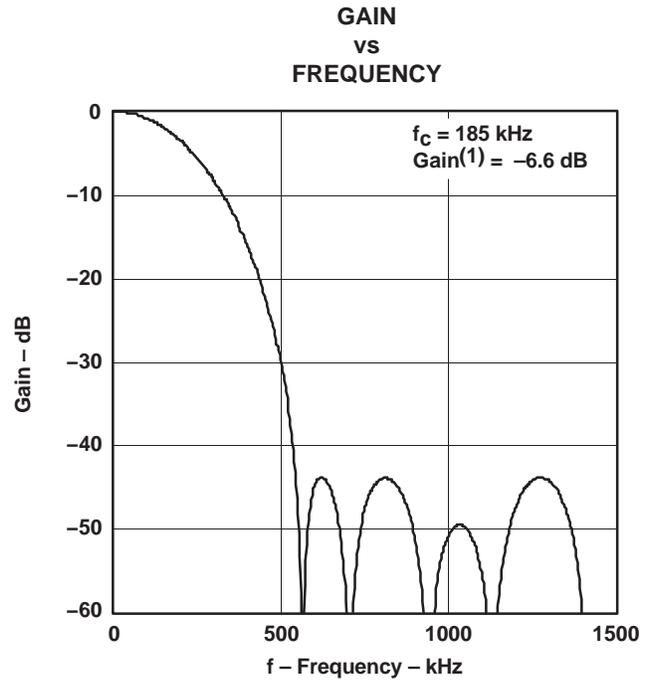
Figure 44. DSD Filter-1, High BW
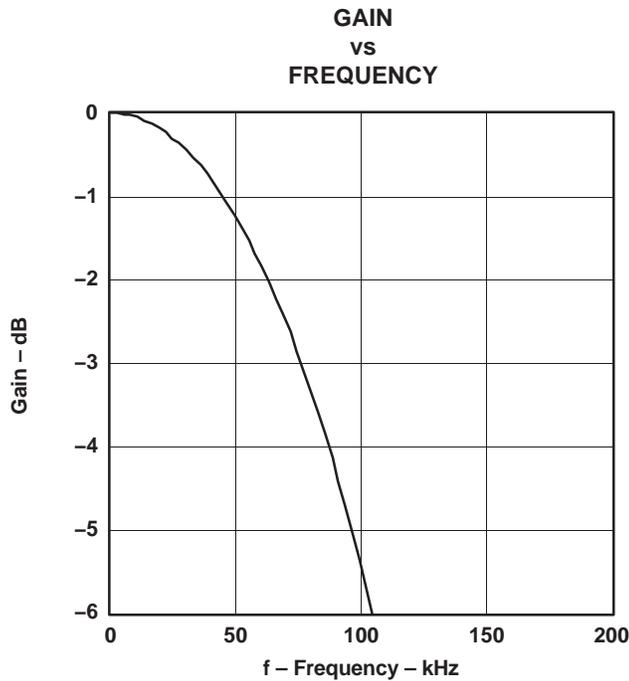
GAIN
vs
FREQUENCY

Figure 45. DSD Filter-2, Low BW
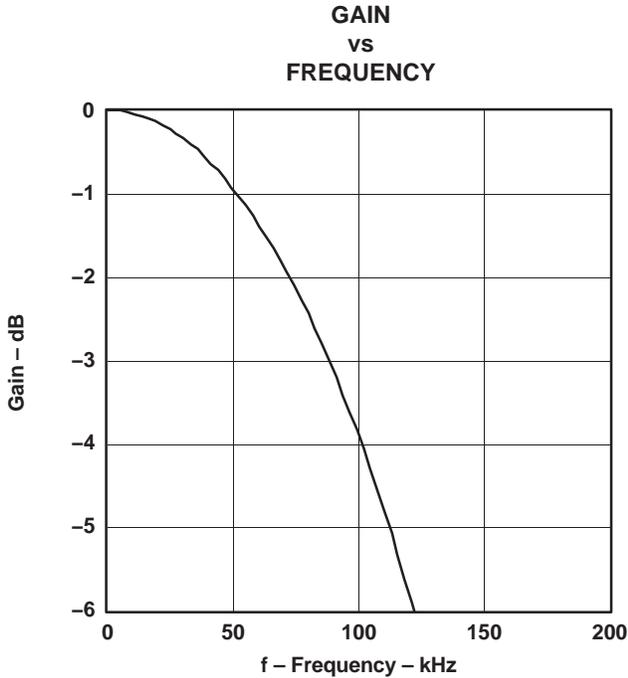
GAIN
vs
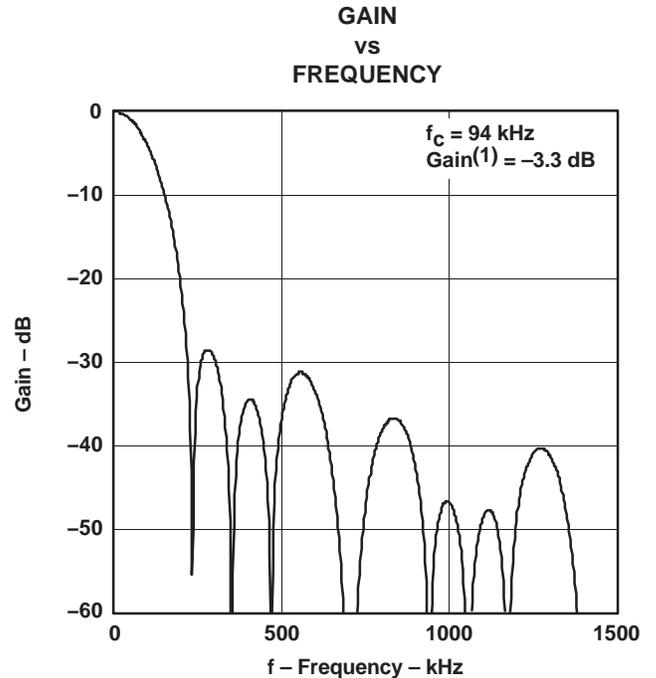FREQUENCY

$f_C = 77 \text{ kHz}$
$\text{Gain}^{(1)} = -6 \text{ dB}$
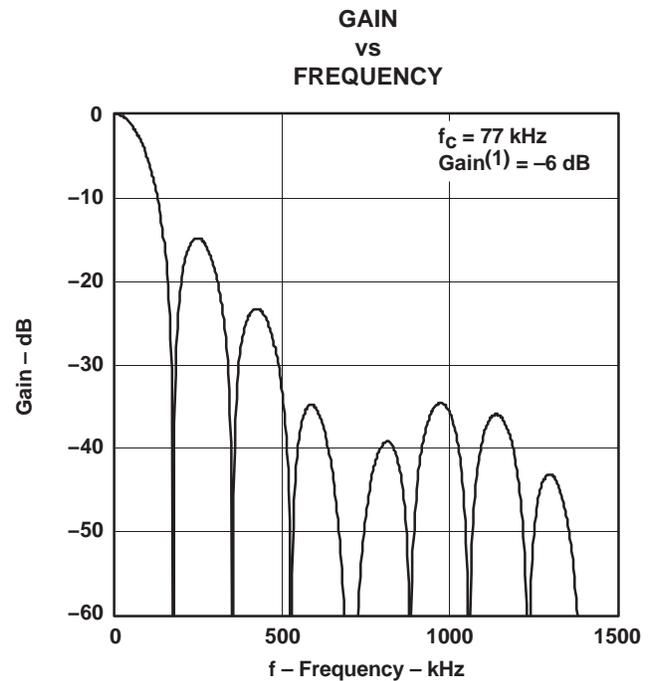
Figure 46. DSD Filter-2, High BW

(1) This gain is in comparison to PCM 0 dB, when the DSD input signal efficiency is 50%.

All specifications at DBCK = 2.8224 MHz (44.1 kHz × 64 f_S) and 50% modulation DSD data input, unless otherwise noted.

ANALOG FIR FILTER PERFORMANCE IN DSD MODE (CONTINUED)
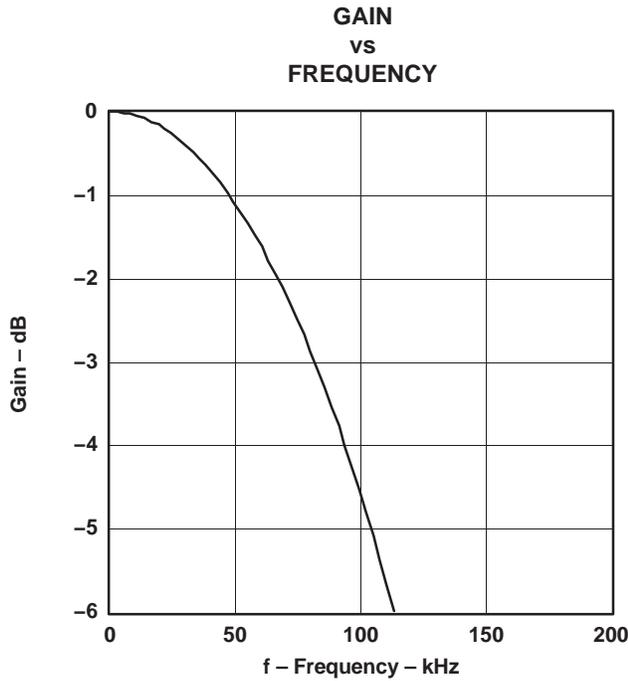
GAIN
vs
FREQUENCY

Figure 47. DSD Filter-3, Low BW

GAIN
vs
FREQUENCY

f_C = 85 kHz
Gain[1] = −1.5 dB
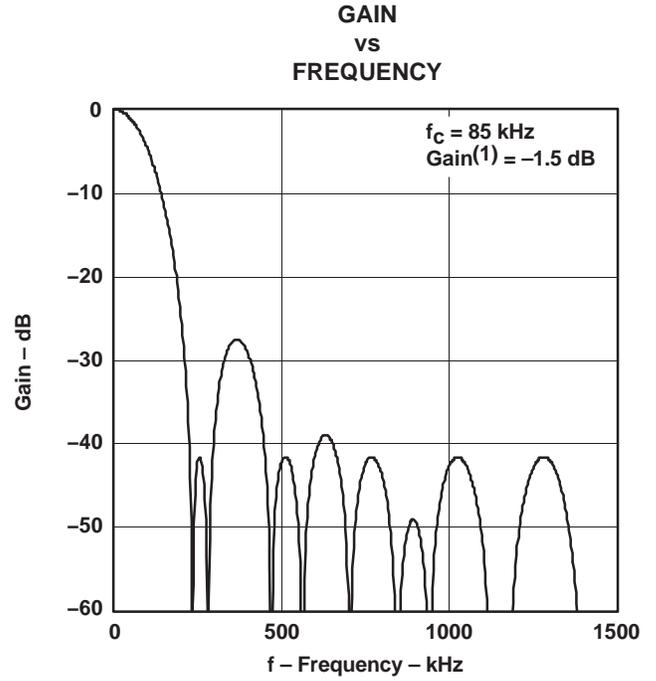
Figure 48. DSD Filter-3, High BW

GAIN
vs
FREQUENCY

Figure 49. DSD Filter-4, Low BW

GAIN
vs
FREQUENCY

f_C = 94 kHz
Gain[1] = −3.3 dB

Figure 50. DSD Filter-4, High BW

(1) This gain is in comparison to PCM 0 dB, when the DSD input signal efficiency is 50%.

All specifications at DBCK = 2.8224 MHz (44.1 kHz × 64 f_S) and 50% modulation DSD data input, unless otherwise noted.

DSD MODE CONFIGURATION AND FUNCTION CONTROLS

Configuration for the DSD Interface Mode

DSD = 1 (Register 20, B5)

	B15	B14	B13	B12	B11	B10	B9	B8	B7	B6	B5	B4	B3	B2	B1	B0
Register 16	R/$\overline{\text{W}}$	0	0	1	0	0	0	0	–	–	–	–	–	–	–	–
Register 17	R/$\overline{\text{W}}$	0	0	1	0	0	0	1	–	–	–	–	–	–	–	–
Register 18	R/$\overline{\text{W}}$	0	0	1	0	0	1	0	–	–	–	–	DMF1	DMF0	–	–
Register 19	R/$\overline{\text{W}}$	0	0	1	0	0	1	1	REV	–	–	OPE	–	–	–	–
Register 20	R/$\overline{\text{W}}$	0	0	1	0	1	0	0	–	SRST	1	–	MONO	CHSL	OS1	OS0
Register 21	R	0	0	1	0	1	0	1	–	–	–	–	–	DZ1	DZ0	–
Register 22	R	0	0	1	0	1	1	0	–	–	–	–	–	–	ZFGR	ZFGL

NOTE: – indicates that function is disabled. No operation even if data bit is set

DMF[1:0]: Analog FIR Performance Selection

Default value: 00

DMF[1:0]	Analog-FIR Performance Select
00	FIR-1 (default)
01	FIR-2
10	FIR-3
11	FIR-4

Plots for the four analog FIR filter responses are shown in the *TYPICAL PERFORMANCE CURVES* section of this data sheet.

OS[1:0]: Analog-FIR Operation-Speed Selection

Default value: 00

OS[1:0]	Operation Speed Select
00	f_{DBCK} (default)
01	$f_{DBCK}/2$
10	Reserved
11	$f_{DBCK}/4$

The OS bit in the DSD mode is used to select the operating rate of the analog FIR. The OS bits must be set before setting the DSD bit to 1.

Requirements for System Clock

The bit clock (DBCK) for the DSD mode is required at pin 3 of the DSD1794A. The frequency of the bit clock can be N times the sampling frequency. Generally, N is 64 in DSD applications.

The interface timing between the bit clock and DSDL and DSDR is required to meet the same setup-and hold-time specifications as shown in Figure 42.
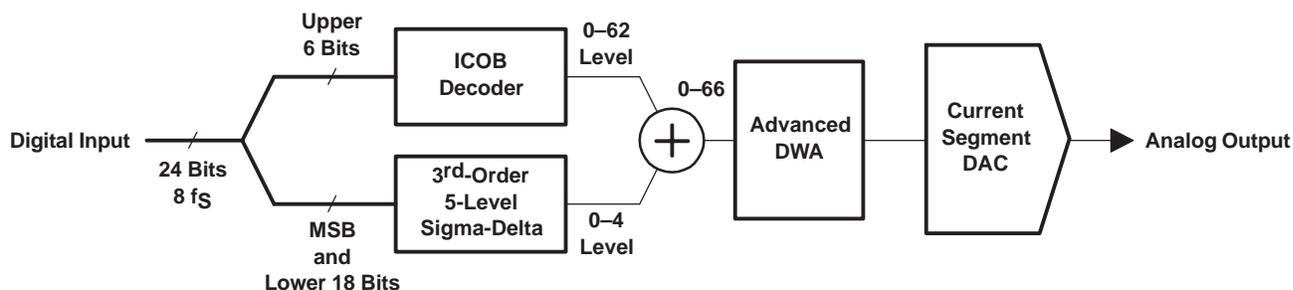
THEORY OF OPERATION

Figure 51. Advanced Segment DAC

The DSD1794A uses TI's advanced segment DAC architecture to achieve excellent dynamic performance and improved tolerance to clock jitter. The DSD1794A provides balanced current outputs.

Digital input data via the digital filter is separated into 6 upper bits and 18 lower bits. The 6 upper bits are converted to inverted complementary offset binary (ICOB) code. The lower 18 bits, associated with the MSB, are processed by a five-level third-order delta-sigma modulator operated at $64 f_S$ by default. The 1 level of the modulator is equivalent to the 1 LSB of the ICOB code converter. The data groups processed in the ICOB converter and third-order delta-sigma modulator are summed together to an up to 66-level digital code, and then processed by data-weighted averaging (DWA) to reduce the noise produced by element mismatch. The data of up to 66 levels from the DWA is converted to an analog output in the differential-current segment section.

This architecture has overcome the various drawbacks of conventional multibit processing and also achieves excellent dynamic performance.

Analog output

The following table and Figure 52 show the relationship between the digital input code and analog output.

	800000 (–FS)	000000 (BPZ)	7FFFFF (+FS)
$I_{OUT}N$ [mA]	–2.3	–6.2	–10.1
$I_{OUT}P$ [mA]	–10.1	–6.2	–2.3
$V_{OUT}N$ [V]	–1.725	–4.65	–7.575
$V_{OUT}P$ [V]	–7.575	–4.65	–1.725
V_{OUT} [V]	–2.821	0	2.821

NOTE: $V_{OUT}N$ is the output of U1, $V_{OUT}P$ is the output of U2, and V_{OUT} is the output of U3 in the application circuit of Figure 33.

**OUTPUT CURRENT
vs
INPUT CODE**
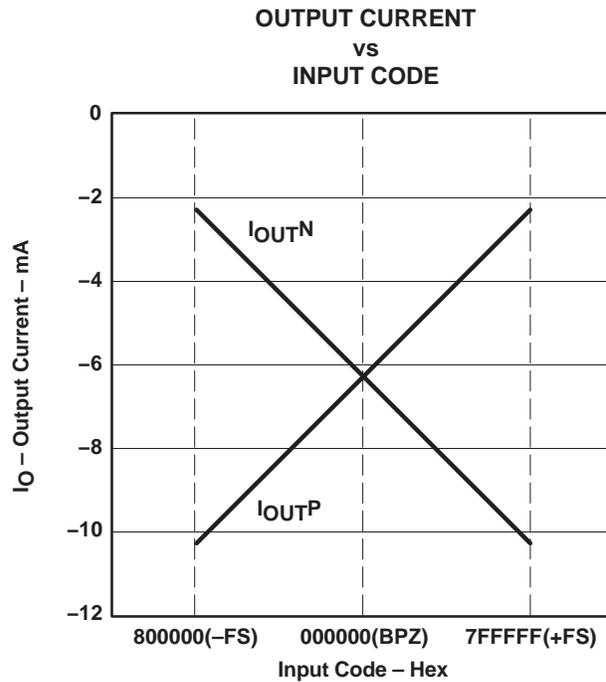
Figure 52. The Relationship Between Digital Input and Analog Output

PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
DSD1794ADB	ACTIVE	SSOP	DB	28	47	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-25 to 85	DSD1794 A	Samples
DSD1794ADBR	ACTIVE	SSOP	DB	28	2000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-25 to 85	DSD1794 A	Samples
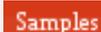
(1) The marketing status values are defined as follows:
ACTIVE: Product device recommended for new designs.
LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.
NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.
PREVIEW: Device has been announced but is not in production. Samples may or may not be available.
OBSOLETE: TI has discontinued the production of the device.

(2) **RoHS:** TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".
RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.
Green: TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.
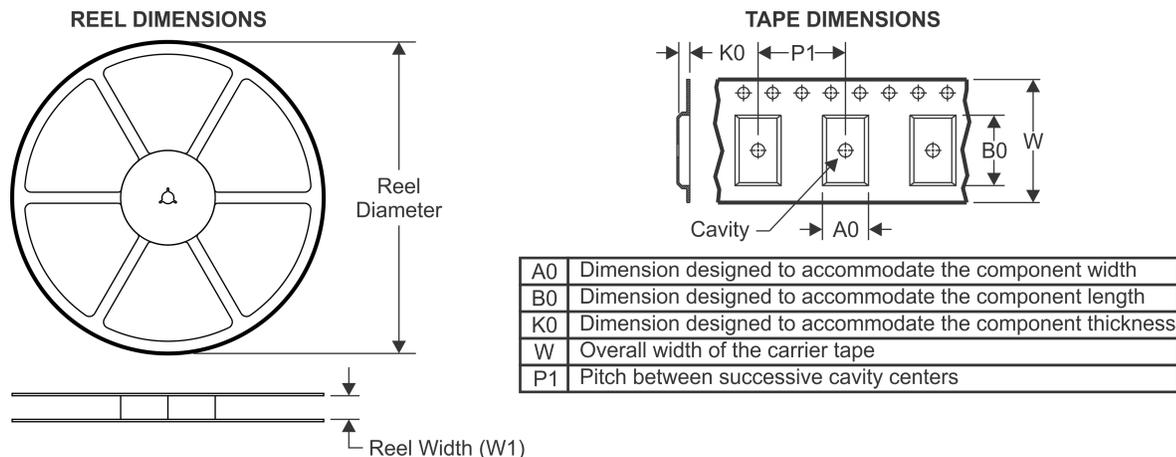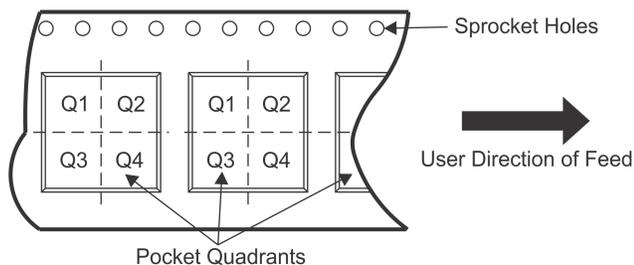
(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.

PACKAGE MATERIALS INFORMATION

TAPE AND REEL INFORMATION

REEL DIMENSIONS

TAPE DIMENSIONS

A0	Dimension designed to accommodate the component width
B0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE

*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
DSD1794ADBR	SSOP	DB	28	2000	330.0	17.4	8.5	10.8	2.4	12.0	16.0	Q1

PACKAGE MATERIALS INFORMATION

TAPE AND REEL BOX DIMENSIONS
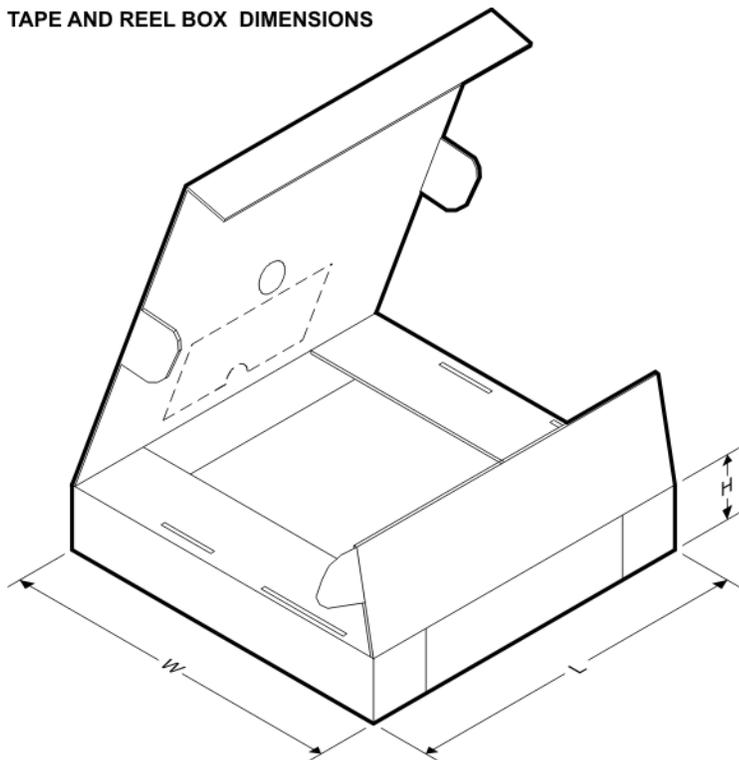
*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
DSD1794ADBR	SSOP	DB	28	2000	336.6	336.6	28.6

TUBE
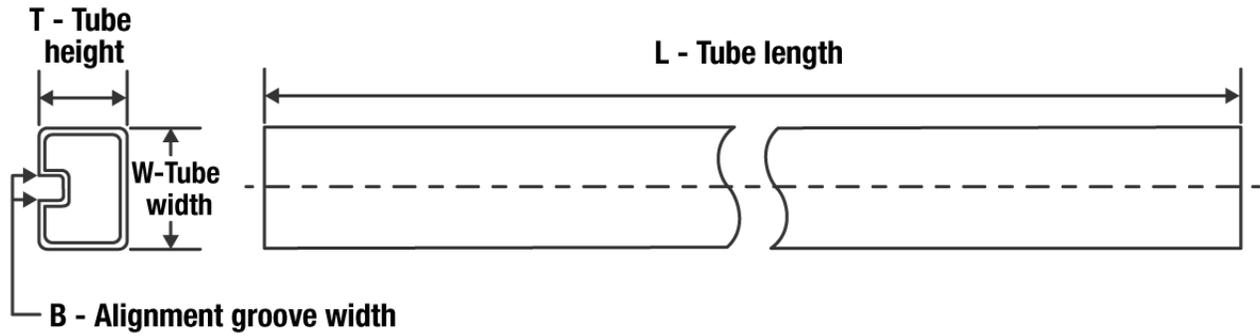
*All dimensions are nominal

Device	Package Name	Package Type	Pins	SPQ	L (mm)	W (mm)	T (µm)	B (mm)
DSD1794ADB	DB	SSOP	28	47	500	10.6	500	9.6

8.2
7.4 TYP

A

PIN 1 INDEX AREA

1

26X 0.65

28

C

0.1 C

SEATING
PLANE

2X
8.45

10.5
9.9
NOTE 3

14

15

28X 0.38
0.22

⊕ 0.15Ⓜ C A B

B

5.6
5.0
NOTE 4

SEE DETAIL A

(0.15) TYP

GAGE PLANE

0.25

2 MAX

0°-8°

0.95
0.55

0.05 MIN

DETAIL A
TYPICAL
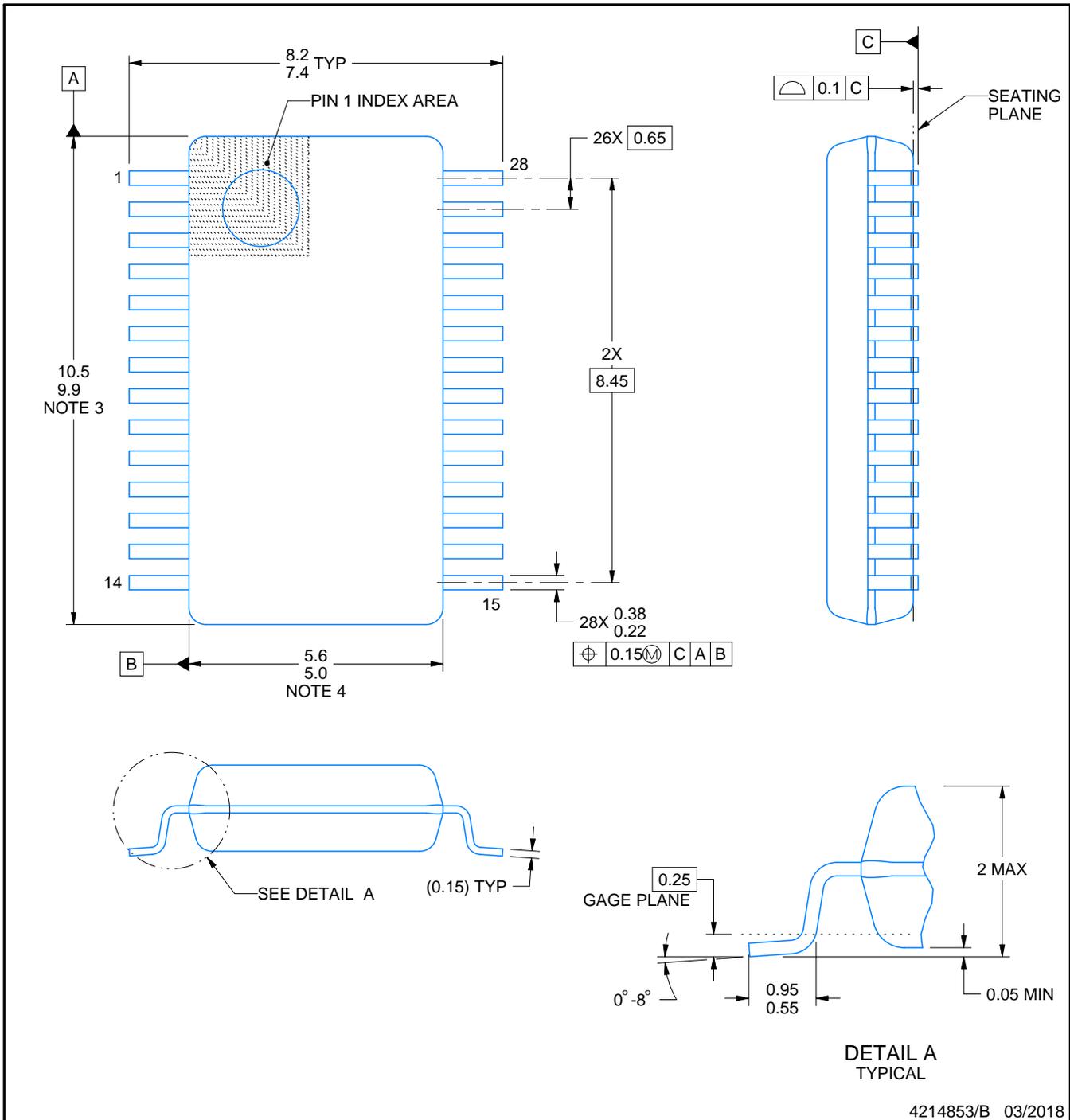
4214853/B 03/2018

NOTES:

1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 mm per side.
4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm per side.
5. Reference JEDEC registration MO-150.

TEXAS
INSTRUMENTS
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28X (1.85)

1

28X (0.45)

26X (0.65)

SYMM

(R0.05) TYP

28

SYMM

14

15

(7)

LAND PATTERN EXAMPLE
EXPOSED METAL SHOWN
SCALE: 10X

SOLDER MASK
OPENING

METAL

EXPOSED METAL

0.07 MAX
ALL AROUND

NON-SOLDER MASK
DEFINED
(PREFERRED)

METAL UNDER
SOLDER MASK

SOLDER MASK
OPENING

EXPOSED METAL

0.07 MIN
ALL AROUND

SOLDER MASK
DEFINED

SOLDER MASK DETAILS

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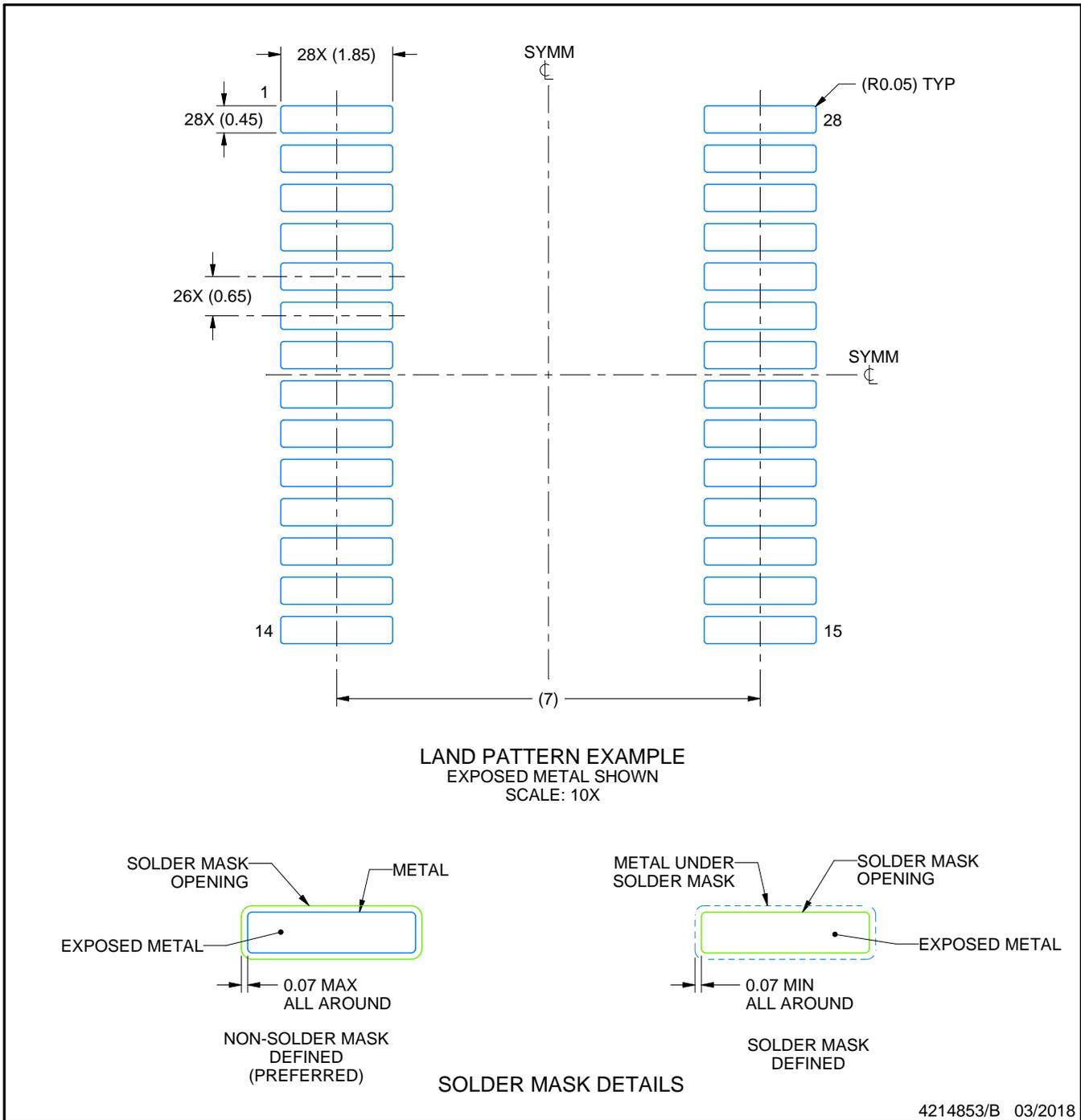
NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.
7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

28X (1.85)

SYMM

(R0.05) TYP

1

28X (0.45)

28

26X (0.65)

SYMM

14

15

(7)

SOLDER PASTE EXAMPLE
BASED ON 0.125 mm THICK STENCIL
SCALE: 10X

4214853/B   03/2018
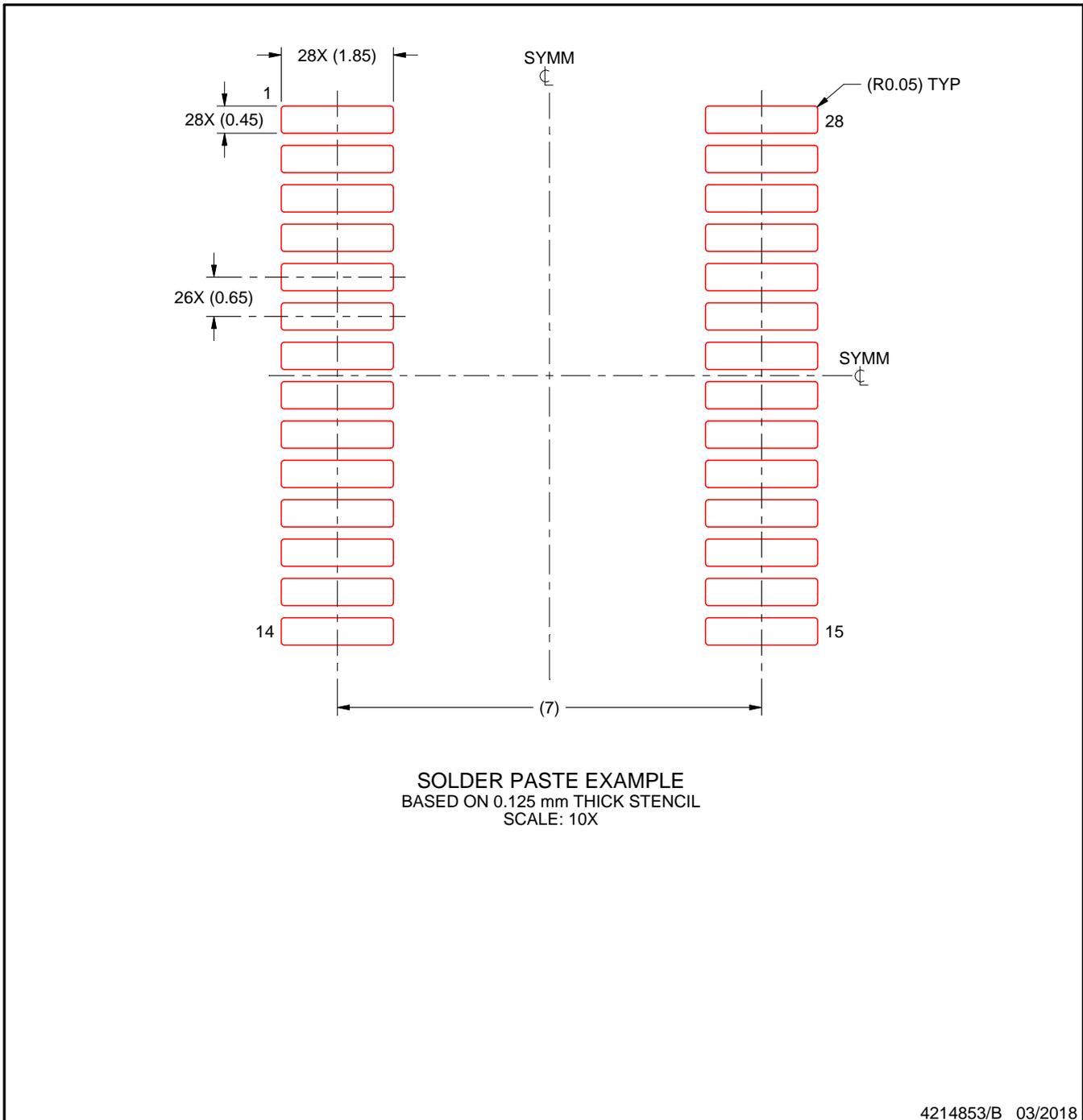
NOTES: (continued)

8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
9. Board assembly site may have different recommendations for stencil design.